MIC3003GFL

FOM Management IC
with Internal Calibration

General Description

The MIC3003GFL is a fiber optic module controller which enables the implementation of sophisticated, hot-pluggable fiber optic transceivers with intelligent laser control and an internally calibrated Digital Diagnostic Monitoring Interface per SFF-8472. It essentially integrates all non-data path functions of an SFP/SFP+ transceiver into a tiny (3mm x 3mm) MLF® package. It also works well as a microcontroller peripheral in transponders or 10Gbps transceivers. The MIC3003GFL uses the same die as the MIC3003 with all its functions, but in a smaller package and different pin out.

A highly configurable automatic power control (APC) circuit controls laser bias. Bias and modulation are temperature compensated using dual DACs, an on-chip temperature sensor, and NVRAM look-up tables. A programmable internal feedback resistor provides a wide dynamic range for the APC. Controlled laser turn-on.

An analog-to-digital converter converts the measured temperature, voltage, bias current, transmit power, and received power from analog to digital. An EEPOT provides front-end adjustment of RX power. Each parameter is compared against user-programmed warning and alarm thresholds. Analog comparators and DACs provide fast monitoring of received power and critical laser operating parameters. Data can be reported as either internally calibrated or externally calibrated.

An interrupt output, power-on hour meter, and data-ready bits add user friendliness beyond SFF-8472. The interrupt output and data-ready bits reduce overhead in the host system. The power-on hour meter logs operating hours using an internal real-time clock and stores the result in NVRAM.

In addition to the features listed above, the MIC3003 features an extended temperature range, options to mask alarms and warnings interrupt and TXFAULT, a reset signal source, and the ability to support up to four chips with the same address on the serial interface. It also supports eight-byte SMBus block writes.

Communication with the MIC3003 is via an industry standard 2-wire SMBus serial interface. Nonvolatile memory is provided for serial ID, configuration, and separate OEM and user scratchpad spaces.

Datasheets and support documentation can be found on Micrel's web site at: www.micrel.com.

Features

- Packaged in a ultra small (3mm x 3mm) 24-pin MLF® package
- Extended temperature range
- Alarms and warnings interrupt and TXFAULT masks
- Capability to support up to four devices on one SMBus
- APC or constant-current laser bias
- Turbo mode for APC loop start-up and shorter laser turn on time
- Supports multiple laser types and bias circuit topologies
- Integrated digital temperature sensor
- Temperature compensation of modulation, bias, bias fault and alarm thresholds via NVRAM look-up tables
- NVRAM to support GBIC/SFP serial ID function
- User writable EEPROM scratchpad
- Reset signal compatible with some new systems requirements
- Diagnostic monitoring interface per SFF-8472
  - Monitors and reports critical parameters: temperature, bias current, TX and RX optical power, and supply voltage
  - S/W control and monitoring of TXFAULT, RXLOS, RATESELECT, and TXDISABLE
  - Internal or external calibration
  - EEPOT for adjusting RX power measurement
- Power-on hour meter
- Interrupt capability
- Extensive test and calibration features
- 2-wire SMBus-compatible serial interface
- SFP/SFP+ MSA and SFF-8472 compliant
- 3.0V to 3.6V power supply range
- 5V-tolerant I/O

Applications

- SFP/SFP+ optical transceivers
- SONET/SDH transceivers and transponders
- Fibre Channel transceivers
- 10Gbps transceivers
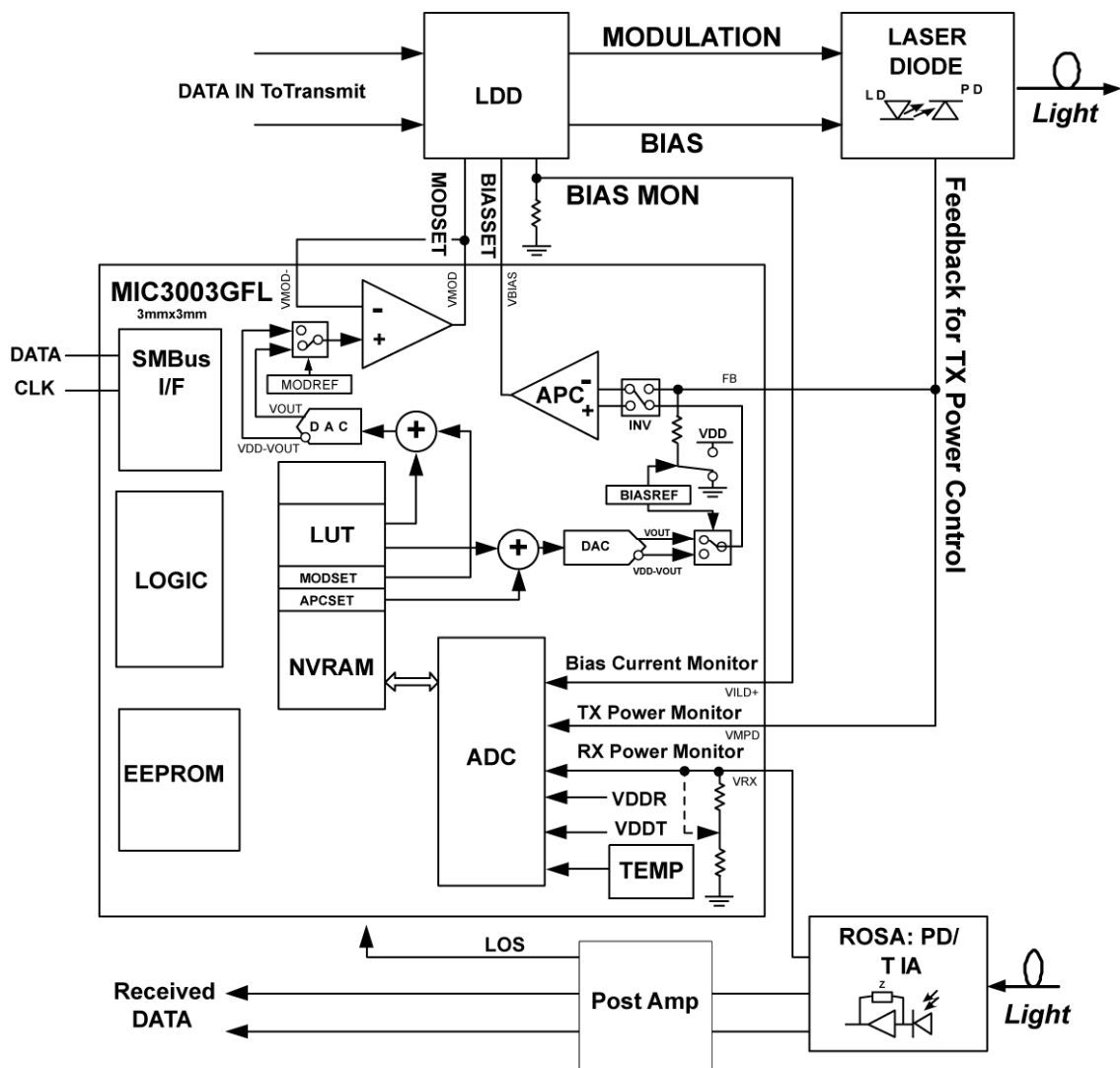- Free space optical communications
- Proprietary optical links

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Typical Application

Ordering Information

Part Number	Package Marking	Junction Temp. Range	Package Type	Lead Finish
MIC3003GFL	GFL 3003 with Pb-Free bar-line indicator	–45°C to +105°C	24-pin (3mm x 3mm) MLF®	Pb-Free, NiPdAu
MIC3003GFLTR[1]	GFL 3003 with Pb-Free bar-line indicator	–45°C to +105°C	24-pin (3mm x 3mm) MLF®	Pb-Free, NiPdAu

Note:

1.   Tape and Reel.

Contents

List of Figures

List of Tables

Pin Configuration

24-Pin MLF® (ML)

Pin Description (MIC3003GFL only)

Pin Number	Pin Name	Pin Function
1	VDDD	Power supply input for digital functions.
2	QGPO	Open-drain output. Can be selected (via OEMCFG3 bit 7) to be an open-drain GPO or an active-low, open-drain, pulsed reset signal output controlled by the status of bits [0-2] of byte A2h: FFh.
3	GNDD	Ground return for digital functions.
4	RS0	Digital input. Receiver Rate Select input. OR'ed with soft rate select bit SRS0 to determine the state of the RRSOUT pin. The state of this pin is always reflected in the RS0S bit.
5	VIN/INT	If bit 4 (IE) in the USRCTL register is set to 0 (its default value), this pin is configured as an analog input. If IE bit is set to 1, this pin is configured as an open-drain output.
		Analog input: Multiplexed A/D input for monitoring supply voltage, with a 0V to 5.5V input range. Open-drain output: outputs the internally generated active-low interrupt signal /INT.
6	CLK	Digital input. Serial clock input.
7	DATA	Digital I/O, open-drain, bi-directional serial data input/output.
8	TXDISABLE	Digital input; Active high. The transmitter is disabled when this input is high or the STXDIS bit is set to 1. The state of this input is always reflected in the TXDIS bit.
9	TXFAULT	Digital Output; Open-Drain, with programmable polarity. If OEMCFG5 bit 4 is set to 0, a high level indicates a hardware fault impeding transmitter operation. If OEMCFG5 bit 4 is set to 1, a low level indicates a hardware fault impeding transmitter operation. The state of this pin is always reflected in the TXFLT bit.
10	RS1	Digital Input; Transmitter Rate Select Input; OR'ed with soft rate select bit SRS1 to determine the state of the TRSOUT pin. The state of this pin is always reflected in the RS1S bit.
11	VRX	Analog Input. Multiplexed A/D converter input for monitoring received optical power. The input range is 0 to V_{REF}. A 5-bit programmable EEPOT on this pin provides coarse calibration and ranging of the RX power measurement.

Pin Number	Pin Name	Pin Function
12	SHDN/TXFIN	Digital output/Input; programmable polarity. When used as shutdown output (SHDN), OEMCFG3 bit 2 set to 0, SHDN is asserted at the detection of a fault condition if OEMCFG4 bit 7 is set to 0. If OEMCFG4 bit 7 is set to 1, a fault condition will not assert SHDN. When programmed as TXFIN, it is an input for external fault signals to be OR'ed with the internal fault sources to drive TXFAULT.
13	VILD+	Analog Input. Multiplexed A/D input for monitoring laser bias current via a sense resistor (signal input); accommodates inputs referenced to V_{DD} or GND (see pin 14 description).
14	VILD–	Analog Input. Reference terminal for the multiplexed pseudo-differential A/D converter inputs for monitoring laser bias current via a sense resistor (VILD+ is the sensing input). Tie to V_{DD} or GND to reference the voltage sensed on VILD+ to V_{DD} or GND, respectively.
15	VDDA	Power supply input for analog functions.
16	GNDA	Ground return for analog functions.
17	VMPD	Analog Input. Multiplexed A/D converter input for monitoring transmitted optical power via a monitor photodiode. In most applications, VMPD will be connected directly to FB. The input range is 0 - V_{REF} or 0 - V_{REF}/4 depending upon the setting of the APC configuration bits
18	FB	Analog Input. Feedback voltage for the APC loop op-amp. Polarity and scale are programmable via the APC configuration bits I OEMCFG1. Connect to V_{BIAS} if APC is not used.
19	VMOD	Analog Output. Buffered DAC output to set the modulation current on the laser driver IC. Operates with either a 0– V_{REF} or a $(V_{DD}–V_{REF})$ – V_{DD} output swing so as to generate either a ground-referenced or a V_{DD} referenced programmed voltage. A simple external circuit can be used to generate a programmable current for those drivers that require a current rather than a voltage input.
20	VMOD–	Analog input. This pin is the inverting terminal of the VMOD buffer op-amp. Connect to VMOD (gain = 1) or a feedback resistor network to set a different gain value.
21	VBIAS	Analog output. Buffered DAC output capable of sourcing or sinking up to 10mA under control of the APC function to drive an external transistor or the APCSET pin of a laser diode driver for laser diode DC bias. The output and feedback polarity are programmable to accommodate either an NPN or a PNP transistor to drive a common-anode or common-cathode laser diode.
22	COMP	Analog output. Compensation terminal for the APC loop. Connect a capacitor between this pin and GNDA or V_{DDA} with the appropriate value to tune the APC loop time constant to a desirable value.
23	RRSOUT/ GPO	Digital Output. Open-Drain or push-pull. If OEMCFG3 bit 4 is set to 0, RRSOUT is selected. It represents the receiver rate select as per SFF. This output is controlled by the SRS0 bit OR'ed with RS0 input and is open drain only. If OEMCFG3 bit 4 is set to 1, GPO is selected. General-purpose, non-volatile output, it is controlled by the GPO configuration bits in OEMCFG3.
24	RXLOS/ TRSOUT	Digital output. This programmable polarity, open-drain outputs has two purposes: If OEMCFG6 bit 2 = 0, indicates the loss of the received signal as indicated by a level of received optical power below the programmed RXLOS comparator threshold; may be wire-OR'ed with external signals. Normal operation is indicated by a low level when OEMCFG6 bit 3 is set to 0 and a high level when OEMCFG6 bit 3 is set to 1. RXLOS is de-asserted when VRX > LOSFLTn. The LOS bit reflects the state of RXLOS whether driven by the MIC3003 or an external circuit. If OEMCFG6 bit 2 = 1, TRSOUT is selected. This signal represents the transmitter rate select as per the SFF specification. This output is controlled by the SRS1 bit OR'ed with the RS1 input.

Absolute Maximum Ratings[1]

Power Supply Voltage, V_{DD} .... +3.8V
Voltage on CLK, DATA, TXFAULT, VIN, RXLOS,
   TXDISABLE, RS0, RS1 .... –0.3V to +6.0V
Voltage On Any Other Pin .... –0.3V to V_{DD}+0.3V
Power Dissipation, T_A = 85°C 1.5W
Junction Temperature (T_J) .... 150°C
Storage Temperature (T_S) .... –65°C to +150°C
Soldering (20 sec.) 260ºC
ESD Ratings[3]
   Human Body Model 2kV
   Machine Model 300V

Operating Ratings[2]

Power Supply Voltage, V_{DDA}/V_{DDD} .... +3.0V to +3.6V
Ambient Temperature Range (T_A) .... –40°C to +105°C
Package Thermal Resistance
   MLF® (θ_{JA}) .... 60°C/W

Electrical Characteristics

For typical values, T_A = 25°C, V_{DDA} = V_{DDD} = +3.3V, unless otherwise noted. **Bold** values are guaranteed for +3.0V ≤ (V_{DDA} = V_{DDD}) ≤ 3.6V, $T_{(min)}$ ≤ T_A ≤ $T_{(min)}$, [8]

Symbol	Parameter	Condition	Min	Typ	Max	Units
Power Supply						
I_{DD}	Supply Current	CLK = DATA = V_{DDD} = V_{DDA}; TXDISABLE low; all DACs at full-scale; all A/D inputs at full-scale; all other pins open.		2.3	3.5	mA
		CLK = DATA = V_{DDD} = V_{DDA}; TXDISABLE high; FLTDAC at full-scale; all A/D inputs at full-scale; all other pins open.		2.3	3.5	mA
V_{POR}	Power-on Reset Voltage	All registers reset to default values; A/D conversions initiated.		2.9	2.98	V
V_{UVLO}	Under-Voltage Lockout Threshold	Note 5	2.5	2.73		V
V_{HYST}	Power-on Reset Hysteresis Voltage			170		mV
t_{POR}	Power-on Reset Time	V_{DD} > V_{POR}, Note 4		50		µs
V_{REF}	Reference Voltage		1.210	1.225	1.240	V
$\Delta V_{REF}/\Delta V_{DDA}$	Voltage Reference Line Regulation			1.7		mV/V
Temperature-to-Digital Converter Characteristics						
	Local Temperature Measurement Error	–40°C ≤ T_A ≤ +105°C, Note 6		±1	±3	°C
t_{CONV}	Conversion Time	Note 4			60	ms
t_{SAMPLE}	Sample Period				100	ms
Voltage-to-Digital Converter Characteristics (V_{RX}, V_{AUX}, V_{BIAS}, V_{MPD}, V_{ILD}±)						
	Voltage Measurement Error	–40°C ≤ T_A ≤ +105°C, Note 6		±1	±2.0	%fs
t_{CONV}	Conversion Time	Note 4			10	ms
t_{SAMPLE}	Sample Period	Note 4			100	ms

Notes:
1. Exceeding the absolute maximum rating may damage the device.
2. The device is not guaranteed to function outside its operating rating.
3. Devices are ESD sensitive. Handling precautions recommended. Human body model, 1.5k in series with 100pF.
4. Guaranteed by design and/or testing of related parameters. Not 100% tested in production.
5. The MIC3003 will attempt to enter its shutdown state when V_{DD} falls below V_{UVLO}. This operation requires time to complete. If the supply voltage falls too rapidly, the operation may not be completed.
6. Does not include quantization error.

Voltage Input, V_{IN} (Pin 5 used as an ADC Input)

Symbol	Parameter	Condition	Min	Typ	Max	Units
V_{IN}	Input Voltage Range	$-0.3 \text{ V} \leq V_{DD} \leq 3.6\text{V}$	GNDA		5.5	V
I_{LEAK}	Input Current	$V_{IN} = V_{DD}$ or GND; $V_{AUX} = V_{IN}$		55		µA
C_{IN}	Input Capacitance			10		pF

Digital-to-Voltage Converter Characteristics (V_{MOD}, V_{BIAS})

	Parameter	Condition	Min	Typ	Max	Units
	Accuracy	$-40°C \leq T_A \leq +105°C$, Note 7		±1	2.0	%fs
t_{CONV}	Conversion Time	Note 8			20	ms
DNL	Differential Non-linearity Error	Note 8		±0.5	±1	LSB

Bias Current Sense Inputs, $V_{ILD}+$, $V_{ILD}-$

Symbol	Parameter	Condition	Min	Typ	Max	Units
V_{ILD}	Differential Input Signal Range, $\lvert V_{ILD}^+ - V_{ILD}^- \rvert$		0		$V_{REF}/4$	mV
I_{IN+}	V_{ILD}^+ input current				±1	µA
I_{IN-}	V_{ILD}^- input current	V_{ILD}^- referred to V_{DDA}		+150		µA
	$\lvert V_{ILD}^+ - V_{ILD}^- \rvert = 0.3V$	V_{ILD}^- referred to GND		-150		µA
C_{IN}	Input Capacitance			10		pF

APC Op Amp, FB, V_{BIAS}, COMP

Symbol	Parameter	Condition	Min	Typ	Max	Units
GBW	Gain Bandwidth Product	C_{COMP} = 20pF; Gain = 1		1		MHz
TC_{VOS}	Input Offset Voltage Temperature Coefficient[4]			1		µV/°C
V_{OUT}	Output Voltage Swing	I_{OUT} = 10mA, SRCE bit = 1	GNDA		1.25	V
		I_{OUT} = -10mA, SRCE bit = 0	V_{DDA} -1.25		V_{DDA}	V
I_{SC}	Output Short-Circuit Current			55		mA
t_{SC}	Short Circuit Withstand Time	$T_J \leq 150°C$, Note 8				sec
PSRR	Power Supply Rejection Ratio	C_{COMP} = 20pF; gain = 1, to GND		55		dB
		C_{COMP} = 20pF; gain = 1, to V_{DD}		40		
A_{MIN}	Minimum Stable Gain	C_{COMP} = 20pF, note 8			1	V/V
$\Delta V/\Delta t$	Slew Rate	C_{COMP} = 20pF; gain = 1		3		V/µs
ΔRFB	Internal Feedback Resistor Tolerance			±20		%
$\Delta RFB/\Delta t$	Internal Feedback Resistor Temperature Coefficient			25		ppm/C
I_{START}	Laser Start-up Current Magnitude	START = 01_h		0.375		mA
		START = 02_h		0.750		mA
		START = 04_h		1.500		mA
		START = 08_h		3.000		mA
C_{IN}	Pin Capacitance			10		pF

Notes:

7. Does not include quantization error.

8. Guaranteed by design and/or testing of related parameters. Not 100% tested in production.

Electrical Characteristics

Symbol	Parameter	Condition	Min	Typ	Max	Units
V_{MOD} Buffer Op-Amp, VMOD, V_{MOD}–						
GBW	Gain Bandwidth	C_{COMP} = 20pF; gain = 1		1		MHz
TC_{VOS}	Input Offset Voltage Temperature Coefficient			1		µV/°C
I_{BIAS}	V_{MOD}– Input Current			±0.1	±1	µA
V_{OUT}	Output Voltage Swing	I_{OUT} = ±1mA	GNDA+75		V_{DDA}-75	mV
I_{SC}	Output Short-Circuit Current			35		mA
t_{SC}	Short Circuit Withstand Time	T_J ≤ 150°C, Note 9				sec
PSRR	Power Supply Rejection Ratio	C_{COMP} = 20pF; gain = 1, to GND		65		dB
		C_{COMP} = 20pF; gain = 1, to V_{DD}		44		dB
A_{MIN}	Minimum Stable Gain	C_{COMP} = 20pF			1	V/V
ΔV/ΔT	Slew Rate	C_{COMP} = 20pF; gain = 1		1		V/µs
C_{IN}	Pin Capacitance			10		pF
Control and Status I/O, TXDISABLE, TXFAULT, RS0, RRSOUT(GPO), SHDN(TXFIN), RXLOS(TRSOUT), /INT, RS1, QGPO						
V_{IL}	Low Input Voltage				0.8	V
V_{IH}	High Input Voltage		2.0			V
V_{OL}	Low Output Voltage	I_{OL} ≤ 3mA			0.3	V
V_{OH}	High Output Voltage (applies to SHDN only)	I_{OH} ≤ 3mA			V_{DDD}–0.3	V
I_{LEAK}	Input Current				±1	µA
C_{IN}	Input Capacitance			10		pF
Transmit Optical Power Input, V_{MPD}						
V_{IN}	Input Voltage Range	Note 9	GNDA		V_{DDA}	V
V_{RX}	Input Signal Range	BIASREF=0			V_{REF}	V
		BIASREF=1	V_{DDA}–V_{REF}		V_{DDA}	V
C_{IN}	Input Capacitance	Note 9		10		pF
I_{LEAK}	Input Current				±1	µA
Received Optical Power Input, VRX, RXPOT						
	Input Voltage Range	Note 9	GNDA		V_{DDA}	V
V_{RX}	Valid Input Signal Range (ADC Input Range)		0		V_{REF}	V
$R_{RXPOT(32)}$	End-to-End Resistance	RXPOT = $1F_h$		32		KΩ
ΔRXPOT	Resistor Tolerance			±20		%
ΔRXPOT/ΔT	Resistor Temperature Coefficient			25		ppm/ºC
$ΔV_{RX}/V_{RXPOT}$	Divider Ratio Accuracy	00 ≤ RXPOT ≤ $1F_h$	-5		+5	%
I_{LEAK}	Input Current	RXPOT = 0 (disconnected)			±1	µA
C_{IN}	Input Capacitance	Note 9		10		pF
I_{LEAK}	Input Current				±1	µA

Note:
9. Guaranteed by design and/or testing of related parameters. Not 100% tested in production.

Electrical Characteristics

Symbol	Parameter	Condition	Min	Typ	Max	Units
Control and Status I/O Timing, TXFAULT, TXDISABLE, RS0, RRSOUT, and RXLOS						
t_{OFF}	TXDISABLE Assert Time	From input asserted to optical output at 10% of nominal, C_{COMP} = 10nF.			**10**	µs
t_{ON}	TXDISABLE De-assert Time	From input de-asserted to optical output at 90% of nominal, C_{COMP} = 10nF.			**1**	ms
t_{INIT}	Initialization Time	From power on or transmitter enabled to optical output at 90% of nominal and TX_FAULT de-asserted. Note 10.			**300**	ms
t_{INIT2}	Power-on Initialization Time	From power on to APC loop-enabled.			**200**	ms
t_{FAULT}	TXFAULT Assert Time	From fault condition to TXFAULT assertion. Note 10.			95	µs
t_{RESET}	Fault Reset Time	Length of time TXDISABLE must be asserted to reset fault condition.	**10**			µs
t_{LOSS_ON}	RXLOS Assert Time	From loss of signal to RXLOS asserted.			95	µs
t_{LOSS_OFF}	RXLOS De-assert Time	From signal acquisition to LOS de-asserted.			100	µs
t_{DATA}	Analog Parameter Data Ready	From power on to valid analog parameter data available. Note 10			400	ms
t_{PROP_IN}	TXFAULT, TXDISABLE, RXLOS, RS0, RS1 Input Propagation Time	Time from input change to corresponding internal register bit set or cleared. Note 10.			1	µs
t_{PROP_OUT}	TXFAULT, TRSOUT, TRRSOUT, /INT, QGPO Output Propagation Time	From an internal register bit set or cleared to corresponding output change. Note 10.			1	µs
Fault Comparators						
FLTTMR	Fault Suppression Timer Clock Period	Note 10.	**0.475**	0.5	**0.525**	ms
	Accuracy		-3		+3	%/fs
t_{REJECT}	Glitch Rejection	Maximum length pulse that will not cause output to change state. Note 10.	4.5			µs
V_{SAT}	Saturation Detection Threshold	High level		95		%VDDA
		Low level		5		%VDDA
Power-On Hour Meter						
	Timebase Accuracy	0°C ≤ T_A ≤ +70°, Note 10.	**+5**		**-5**	%
		–40°C ≤ T_A ≤ +105°C	**+10**		**-10**	%
	Resolution	Note 10.		10		hours
Non-Volatile (FLASH) Memory						
t_{WR}	Write Cycle Time, Note 11	Measured from the SMBus STOP condition of a one-byte to eight-byte write transaction. Note 10.			**13**	ms
	NVRAM Data Retention		**100**			years
Endurance	Maximum permitted number of write cycles to any single NVRAM location		**10,000**			cycles

Notes:

10. Guaranteed by design and/or testing of related parameters. Not 100% tested in production.

11. The MIC3003 will not respond to serial bus transactions during an EEPROM write cycle. The host will receive a NACK response during t_{WR}.

Serial Data I/O Pin, Data

Symbol	Parameter	Condition	Min	Typ	Max	Units
V_{OL}	Low Output Voltage	I_{OL} = 3mA			**0.4**	V
		I_{OL} = 6mA			**0.6**	V
V_{IL}	Low Input Voltage				**0.8**	V
V_{IH}	High Input Voltage		**2.1**			V
I_{LEAK}	Input Current				**±1**	µA
C_{IN}	Input Capacitance	Note 12		10		pF

Serial Clock Input, CLK

Symbol	Parameter	Condition	Min	Typ	Max	Units
V_{IL}	Low Input Voltage	2.7V ≤ V_{DD} ≤ 3.6V			**0.8**	V
V_{IH}	High Input Voltage	2.7V ≤ V_{DD} ≤ 3.6V	**2.1**			V
I_{LEAK}	Input Current				**±1**	µA
C_{IN}	Input Capacitance	Note 12		10		pF

Serial Interface Timing[4]

Symbol	Parameter	Condition	Min	Typ	Max	Units
t_1	CLK (clock) Period		**2.5**			µs
t_2	Data In Setup Time to CLK High		**100**			ns
t_3	Data Out Stable After CLK Low		**300**			ns
t_4	Data Low Setup Time to CLK Low	Start Condition	**100**			ns
t_5	Data High Hold Time After CLK High	Stop Condition	**100**			ns
t_{DATA}	Data Ready Time	From power on to completion of one set of ADC conversions; analog data available via serial interface.			**400**	ms

QGPO Reset Pulse Timing

Symbol	Parameter	Condition	Min	Typ	Max	Units
t_1	QGPO reset pulse low duration	OEMCFG3 bit 7 = 1 A2h:255 (FFh) [2-0] switch to 111	**112.5**	125	137.5	µs
t_2	QGPO reset de-assertion to the clearing of A2:FFh bits 2:0	OEMCFG3 bit 7 = 1 A2h:255 (FFh) [2-0] ≠ 111	**20.25**	22.5	24.75	ms

Note:

12. Guaranteed by design and/or testing of related parameters. Not 100% tested in production.

Serial Interface Timing Diagram
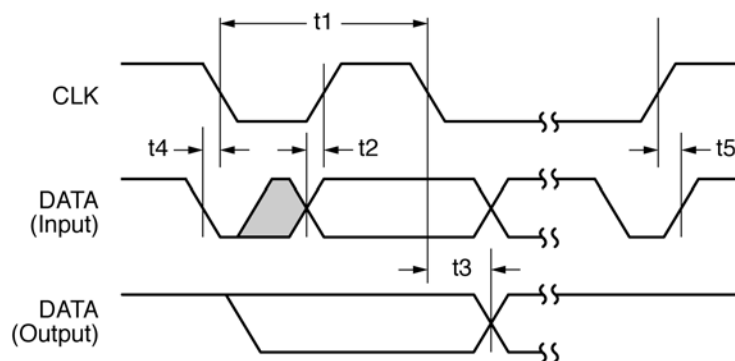
Serial Interface Timing

Serial Interface Address Maps

Address (Decimal)	Field Size (Bytes)	Name	Description
0 –95	96	Serial ID defined by SFP MSA	General-purpose NVRAM; R/W under valid OEM password.
96 – 127	32	Vendor Specific	Vendor specific EEPROM
128 – 255	128	Reserved	Reserved for future use. General-purpose NVRAM; R/W under valid OEM password.

Table 1. Serial Interface Address Map, Device Address = A0$_h$

Address(s) Hex	Dec	Field Size (Bytes)	Name	Description
00-27	0-39	40	Alarm and Warning Thresholds	High/low limits for warnings and alarms; writeable using the OEM password; read-only otherwise.
28-37	40-55	16	Reserved	Reserved – do not write; reads undefined.
38-5B	56-91	36	Calibration Constants	Numerical constants for external calibration; writeable using the OEM password; read-only otherwise.
5C-5E	92-94	3	Reserved	Reserved – do not write; reads undefined.
5F	95	1	Checksum	General-purpose NVRAM; writeable using the OEM password; read-only otherwise.
60-69	96-105	10	Analog Data	Real time analog parameter data.
6A-6D	106-109	4	Reserved	Reserved – do not write; reads undefined.
6E	110	1	Control/Status Register	Control and status bits.
6F	111	1	Rate Select Control	Bits [7-6] control the use of the RS0 and RS1 inputs and the SRS0 and SRS1 register bits.
70-71	112-113	2	Alarm Flags	Alarm status bits; read-only.
72-73	114-115	2	Reserved	Reserved – do not write; reads undefined.
74-75	116-117	2	Warning Flags	Warning status bits; read-only.
76	118	1	Extended Control/Status Register	Additional control and status bits.
77	119	1	Reserved	Reserved – do not write; reads undefined.
78-7E	120-126	7	OEMPW	OEM password entry field. The four-byte OEM password location can be selected to be 78h-7Bh (120-123) by setting OEMCFG5 bit 2 to 0 (default) or 7Bh-7Eh (123-126) by setting OEMCFG5 bit 2 to a one.
7F	127	1	Vendor-specific	Vendor specific. Reserved – do not write; reads undefined.
80-F7	128-247	120	User Scratchpad	User-writeable EEPROM. General-purpose NVRAM.
F8-F9	248-249	2	Alarms Masks	Bit = 0: Corresponding alarm not masked. Bit = 1: Corresponding alarm masked.
FA-FB	250-251	2	Warnings Masks	Bit = 0: Corresponding warning not masked. Bit = 1: Corresponding warning masked.
FC-FD	252-253	2	Reserved	Reserved – do not write; reads undefined.
FE	254	1	USRCTL	End-user control and status bits.
FF	255	1	RESETOUT	Bits [2:0] of this register control the QGPO reset output (pin 2) If OEMCFG3 bit 7 is set to 1.

Table 2. Serial Interface Address Map, Device Address = A2

Address(s)		Field Size (Bytes)		
Hex	Dec		Name	Description
00-3F	0-63	64	BIASLUT1	First 64 entries of the bias current temperature compensation LUT (Look-up Table) The additional 12 entries are located in A6: 58h – 63h.
40-7F	64-127	64	MODLUT1	First 64 entries of the modulation current temperature compensation LUT. The additional 12 entries are located in A6:.64h – 6Fh.
80-BF	128-191	64	IFTLUT1	First 64 entries of the bias current fault threshold temperature compensation LUT. The additional 12 entries are located in A6: 70h - 7Bh.
C0-FF	192-255	64	HATLUT1	First 64 entries of the bias current high alarm threshold temperature compensation LUT. The additional 12 entries are located in A6: 7C-87h.

Table 3. Serial Interface Address Map (Temperature Compensation Tables), Device Address = A4$_h$

Address(s)		Field Size (Bytes)		
Hex	Dec		Name	Description
00	0	1	OEMCFG0	OEM configuration register 0
01	1	1	OEMCFG1	OEM configuration register 1
02	2	1	OEMCFG2	OEM configuration register 2
03	3	1	APCSET0	APC setpoint register 0
04	4	1	APCSET1	APC setpoint register 1
05	5	1	APCSET2	APC setpoint register 2
06	6	1	MODSET0	Modulation setpoint register 0
07	7	1	IBFLT	Bias current fault-comparator threshold. This register is temperature compensated.
08	8	1	TXPFLT	TX power fault threshold
09	9	1	LOSFLT	RX LOS fault-comparator threshold
0A	10	1	FLTTMR	Fault comparator timer setting
0B	11	1	FLTMSK	Fault source mask bits
0C-0F	12-15	4	OEMPWSET	Password for access to OEM areas
10	16	1	OEMCAL0	OEM calibration register 0
11	17	1	OEMCAL1	OEM calibration register 1
12	18	1	LUTINDX	Look-up table index read-back
13	19	1	OEMCFG3	OEM configuration register 3
14	20	1	APCDAC	Reads back current APC DAC value (setpoint+offset)
15	21	1	MODDAC	Reads back current modulation DAC value (setpoint+offset)
16	22	1	OEMREAD	Reads back OEM calibration data
17	23	1	LOSFLTn	LOS de-assert threshold
18	24	1	RXPOT	RXPOT tap selection
19	25	1	OEMCFG4	OEM configuration register 4
1A	26	1	OEMCFG5	OEM configuration register 5
1B	27	1	OEMCFG6	OEM configuration register 6
1C-1D	28-29	2	SCRATCH	Reserved – do not write; reads undefined.
1E	30	1	MODSET 1	Modulation setpoint register 1

Address(s)		Field Size (Bytes)	Name	Description
HEX	DEC			
1F	31	1	MODSET 2	Modulation setpoint register 2
20-27	32-39	8	POHDATA	Power-on hour meter scratchpad
28-47	40-71	32	RXLUT	RX power internal calibration coefficient table. Eight sets of slope and offset coefficients provide a piecewise-linear transform for the receive power ADC result.
48-57	72-87	16	CALCOEF	Slope and offset coefficients used for temperature, voltage, bias current, and transmit power internal calibration
58-63	88-99	12	IFTLUT2	Additional 12 entries of the bias current fault threshold temperature compensation LUT.
64-6F	100-111	12	BIASLUT2	Additional 12 entries of the bias current temperature compensation LUT.
70-7B	112-123	12	MODLUT2	Additional 12 entries of the modulation current temperature compensation LUT.
7C-87	124-135	12	HATLUT2	Additional 12 entries of the bias current high alarm threshold temperature compensation LUT.
88-CF	136-207	72	SCRATCH	OEM scratchpad area
D0-DD	208-221	14	RXLUTSEG/ SCRATCH	Receive power calibration segment delimiters. Each of the eight segments can have its own slope and offset coefficient. Used to refine the shape of the piecewise-linear function used for receive power in internal calibration mode. These bytes may also be part of the OEM scratch pad if the hard coded delimiters option is selected, see the description of OEMCFG6
DE-FA	222-250	29	SCRATCH	OEM scratchpad area
FB-FC	251-252	2	POH	Power on hour meter result; read-only
FD	253	1	Data Ready Flags	Data ready bits for each measured parameter; read-only
FE	254	1	MFG_ID	Manufacturer identification (Micrel's manufacturer ID is 42, 2Ah)
FF	255	1	DEV_ID	Device ID and die revision

Table 4. Serial Interface Address Map (OEM Configuration Registers), Device Address = $A6_h$

Block Diagram
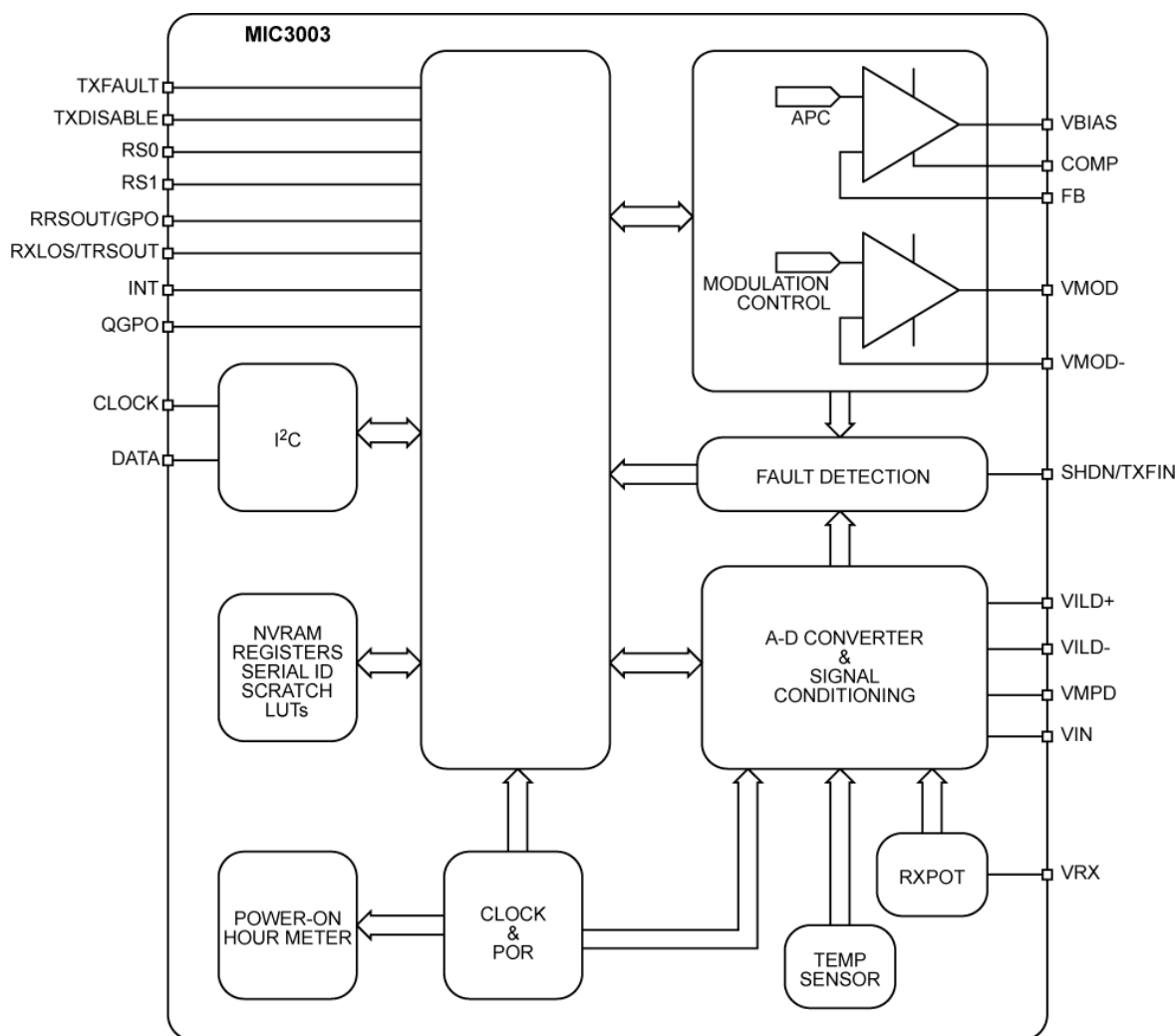
Figure 1. MIC3003 Block Diagram

Analog-to-Digital Converter/Signal Monitoring

A block diagram of the monitoring circuit is shown below. Each of the five analog parameters monitored by the MIC3003 is sampled in sequence. All five parameters are sampled and the results updated within the t_{CONV} duration given in the "Electrical Characteristics" section. In OEM mode, the channel that is normally used to measure V_{IN} may be assigned to measure the level of the V_{DDA} pin or one of five other nodes. This provides a kind of analog loopback for debug and test purposes. The V_{AUX} bits in OEMCFG0 control which voltage source is being sampled. The various V_{AUX} channels are level-shifted differently depending on the signal source, resulting in different LSB values and signal ranges. See Table 5.
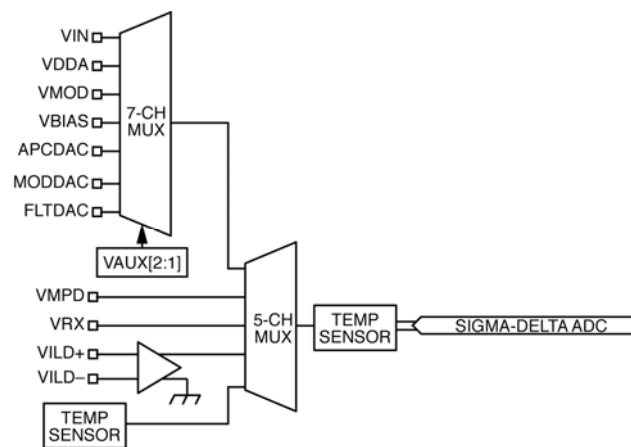
Figure 2. Analog-to-Digital Converter Block Diagram

Channel	ADC Resolution (bits)	Conditions	Input Range (V)	LSB[1]
TEMP	8 or 9		N/A	1°C or 0.5°C
VAUX	8	See Table 6		
VMPD	8	GAIN = 0; BIASREF = 0	$GNDA - V_{REF}$	4.77mV
		GAIN = 0; BIASREF = 1	$V_{DDA} - (V_{DDA} - V_{REF})$	
		GAIN = 1; BIASREF = 0	$GNDA - V_{REF}/4$	1.17mV
		GAIN = 1; BIASREF = 1	$V_{DDA} - (V_{DDA} - V_{REF}/4)$	
VILD	8	VILD- = VDDA	$V_{DDA} - (V_{DDA} - V_{REF})$	4.77mV
		VILD- = GNDA	$GNDA - V_{REF}$	
VRX	12	RXPOT = 00	$0 - V_{REF}$	0.298mV

Table 5. A/D Input Signal Ranges and Resolutions

Note:

1. Assumes typical VREF value of 1.22V.

Channel	VAUX [2:0]	Input Range (V)	LSB[1] (mV)
V_{IN}	$000 = 00_h$	0.5V to 5.5V	25.6mV
V_{DDA}	$001 = 01_h$	0.5V to 5.5V	25.6mV
V_{BIAS}	$010 = 02_h$	0.5V to 5.5V	25.6mV
V_{MOD}	$011 = 03_h$	0.5V to 5.5V	25.6mV
APCDAC	$100 = 04_h$	0V to V_{REF}	4.77mV
MODDAC	$101 = 05_h$	0V to V_{REF}	4.77mV
FLTDAC	$110 = 06_h$	0V to V_{REF}	4.77mV

Table 6. V_{AUX} Input Signal Ranges and Resolutions

Note:

1. Assumes typical V_{REF} value of 1.22V.

Alarms and Warnings Interrupt Source Masking

Alarm and warning violations set the flags in the Alarm and Warning Status Registers, and also assert the interrupt output if they are not masked. If an alarm or warning is masked, it will not set the interrupt. Table 8 shows the locations of the masking bits. The warning or alarm is masked if the corresponding bit is set to 1.

Serial Address A2h			
Byte	Bit	Default Value	Description
248	7	0	Masking bit for Temperature High Alarm interrupt source
	6	0	Masking bit for Temperature Low Alarm interrupt source
	5	0	Masking bit for Voltage High Alarm interrupt source
	4	0	Masking bit for Voltage Low Alarm interrupt source
	3	0	Masking bit for Bias High Alarm interrupt source
	2	0	Masking bit for Bias Low Alarm interrupt source
	1	0	Masking bit for TX Power High Alarm interrupt source
	0	0	Masking bit for TX Power Low Alarm interrupt source
249	7	0	Masking bit for RX Power High Alarm interrupt source
	6	1	Masking bit for RX Power Low Alarm interrupt source
	[5-0]	Reserved	

Table 7. Alarms Interrupt Sources Masking Bits

Serial Address A2h			
Byte	Bit	Default Value	Description
250	7	0	Masking bit for Temperature High Warning interrupt source
	6	0	Masking bit for Temperature Low Warning interrupt source
	5	0	Masking bit for Voltage High Warning interrupt source
	4	0	Masking bit for Voltage Low Warning interrupt source
	3	0	Masking bit for Bias High Warning interrupt source
	2	0	Masking bit for Bias Low Warning interrupt source
	1	0	Masking bit for TX Power High Warning interrupt source
	0	0	Masking bit for TX Power Low Warning interrupt source
251	7	0	Masking bit for RX Power High Warning interrupt source
	6	1	Masking bit for RX Power Low Warning interrupt source
	[5-0]	Reserved	

Table 8. Warnings Interrupt Sources Masking Bits

Alarms and Warnings as TXFAULT Source

Alarms and warnings are not sources for TXFAULT with the default setting. To set alarms as a TXFAULT source set OEMCFG4 bit 6 to 1. To set warnings as a TXFAULT, source set OEMCFG4 bit 7 to 1. The alarms and warnings TXFAULT sources can be masked individually in the same way shown in Tables 7 and 8.

Latching of Alarms and Warnings

Alarms and warnings are latched by default, i.e., once asserted the flags remain ON until the register is read or TXDSABLE is toggled. If OEMCFG4 bit 5 is set to 1, the warnings are not latched and will be set and reset with the warning condition. Reading the register or toggling TXDISABLE will clear the flag. If OEMCFG4 bit 4 is set to 1, the alarms are not latched and will be set and reset with the alarm condition. Reading the register or toggling TXDISABLE will clear the flag.

SMBus Multipart Support

If more than one MIC3003GFL device shares the same serial interface and multipart mode is selected on them (OEMCFG5 bit 3 = 1), then pin 12 and pin 23 become SMBus address bits 3 and 4 respectively. Therefore, the parts should have a different setting on those pins to create four address combinations based upon the state of pin 12 and pin 23 state, (00, 01, 10, 11) where 0 is a pull down to GND and 1 is a pull up to VCC. The parts come from the factory with the same address (A0) and multipart mode off (OEMCFG5 bit 3 is 0). After power up, write 1 to OEMCFG5 bit 3 to turn ON multipart mode, which is done to all parts at the same time since they all respond to serial address A0 at this point. With multipart mode on, the parts have now different addresses based on the states of pins 12 and 23. Another option is to access each part individually, set their single mode address in OEMCFG2 bits [4-7] to different values and then turn off multipart mode to return to normal mode where the parts have new different addresses.

QGOP Pin Function

QGOP can be used in GOP mode as a general purpose output by setting OEMCFG3 bit 7 to 0, or as in RESET mode as a reset signal output by setting OEMCFG3 bit 7 to 1.

If RESET mode is selected, the reset signal state is controlled by RSETOUT (A2:FFh bits [2-0]). By default, these three bits are 000, and the QGPO output is undriven (state: High). When the three bits are written to 111, QGPO's open-drain output will be driven low for 125 µs (typical), after which QGPO reenters the undriven state. The RESETOUT field is cleared from 111 to 000 22.5 ms (typical) after the de-assertion edge of QGPO. Other values of this delay may be selected by setting TRSTCLR (OEMCFG2 bits [2-0]) to different values as shown on table.

If Reset mode in OEMCFG3 is not selected, these three bits have no function.

TRSTCLR [2-0]	Delay from QGPO Switching high to RESETOUT clear
000	Zero delay
001	17.5 ms typical
010	22.5 ms typical (default)
011	27 ms typical
100	45 ms typical

Table 9. RESETOUT Clear Delay

Calibration Modes

The default mode of calibration in the MIC3003 is external calibration, for which the INTCAL bit (bit 0 in OEMCFG3 register) is set to 0. The internal calibration mode is selected by setting INTCAL to 1.

A/ External Calibration

The voltage and temperature values returned by the MIC3003's A/D converter are internally calibrated. The binary values of TEMPh:TEMPl and VOLTh:VOLTl are in the format called for by SFF-8472 under Internal Calibration.

SFF-8472 calls for a set of calibration constants to be stored by the transceiver OEM at specific non-volatile memory locations; refer to the SFF-8472 specifications for the memory map of the calibration coefficients. The MIC3003 provides the non-volatile memory required for the storage of these constants. The Digital Diagnostic Monitoring Interface specification should be consulted for full details. Slopes and offsets are stored for use with voltage, temperature, bias current, and transmitted power measurements. Coefficients for a fourth-order polynomial are provided for use with received power measurements. The host system can retrieve these constants and use them to process the measured data.

Voltage

The voltage values returned by the MIC3003's A/D converter are internally calibrated. The binary values of VOLTh:VOLTl are in the format called for by SFF-8472 under Internal Calibration. Since VINh:VINl requires no processing, the corresponding slope should be set to one and the offset to zero.

Temperature

The temperature values returned by the MIC3003's A/D converter are internally calibrated. The binary values of TEMPh:TEMPl are in the format called for by SFF-8472 under Internal Calibration.

The temperature value may be offset by storing a value in A6:74(4Ah). The temperature offset is a six-bit signed quantity with .5 degrees C resolution.

The temperature offset coefficient at A6:74(4Ah) is used in the same way in both internal and external calibration modes.

Bias Current

Bias current is sensed via an external sense resistor as a voltage appearing between VILD+ and VILD-. The value returned by the A/D is therefore a voltage analogous to bias current. Bias current, IBIAS, is simply V_{VILD}/R_{SENSE}. The binary value in IBIASh (IBIASl is always zero) is related to bias current by:

$$I_{BIAS} = \frac{(0.3V)\dfrac{IBIASh}{255}}{R_{SENSE}} \qquad \textbf{(1)}$$

The value of the least significant bit (LSB) of IBIASh is given by:

$$LSB(IBIASh) = \frac{0.3v}{255 \times R_{SENSE}} \text{ Amps} = \frac{300mV}{255 \times R_{SENSE}}$$

$$mA = \frac{1176.9}{R_{SENSE}} \mu A \qquad \textbf{(2)}$$

Per SFF-8472, the value of the bias current LSB is 2µA. The necessary conversion factor, "slope", is therefore:

$$Slope = \frac{1176.5\mu A}{512\mu A \times R_{SENSE}} = 2.298 + R_{SENSE}$$

The tolerance of the sense resistor directly impacts the accuracy of the bias current measurement. It is recommended that the sense resistor chosen be 1% accurate or better. The offset correction, if needed, can be determined by shutting down the laser, i.e., asserting TXDISABLE, and measuring the bias current. Any non-zero result gives the offset required. The offset will be equal and opposite to the result of the "zero current" measurement.

TX Power

Transmit power is sensed via a resistor carrying the monitor photodiode current. In most applications, the signal at VMPD will be feedback voltage on FB. The VMPD voltage may be measured relative to GND or V_{DDA} depending on the setting of the BIASREF bit in OEMCFG1. The value returned by the A/D is therefore a voltage analogous to transmit power. The binary value in TXOPh (TXOPl is always zero) is related to transmit power by:

$$PTX(mW) = \frac{K \times VREF\dfrac{TXOPh}{255}}{R_{SENSE}} = \frac{K \times (1220mV)\dfrac{TXOPh}{255}}{R_{SENSE}}$$

$$= \frac{K \times 4.7843 \times TXOPh}{R_{SENSE}} mW \qquad \textbf{(3)}$$

For a given implementation, the value of R_{SENSE} is known. It is either the value of the external resistor or the selected internal value of R_{FB}. The constant, K, will likely have to be determined through experimentation or closed-loop calibration, as it depends on the monitoring photodiode responsively and coupling efficiency.

It should be noted that the APC circuit acts to hold the transmitted power constant. The value of transmit power reported by the circuit should only vary by a small amount as long as the APC is functioning correctly.

RX Power

Received power is sensed as a voltage appearing at VRX. It is assumed that this voltage is generated by a sense resistor carrying the receiver photodiode current or by the RSSI circuit of the receiver. The value returned by the A/D is therefore a voltage analogous to received power. The binary values in RXOPh and RXOPl are related to receive power by:

$$RX(mW) = K \times VREF \times (256 \times RXOPh + RXOPl / 16) \bullet 65536$$

(4)

For a given implementation, the constant, K, will likely have to be determined through experimentation or closed-loop calibration, as it depends upon the gain and efficiencies of the receiver. In SFF-8472 implementations, the external calibration constants can describe up to a fourth-order polynomial in case K is nonlinear.

B/ Internal Calibration

If the INTCAL bit in OEMCFG3 is set to 1 (internal calibration selected), the MIC3003 will process each piece of data coming out of the A/D converter before storing the result in result register. Linear slope/offset correction will be applied on a per-channel basis to the measured values for voltage, bias current, TX power, and RX power. Only offset is applied to temperature.

The user must store the appropriate slope/offset coefficients in memory at the time of transceiver calibration. In the case of RX power, a look-up table is provided that implements eight-segment piecewise-linear correction. This correction may be performed as a compensation of the receiver non-linearity over temperature or receive power level. If static slope/offset correction for RX power is desired, the eight coefficient sets can simply be made the same. The user has the option to select between using preset hard-coded delimiters values or programmable delimiters where delimiters corresponding to the best linear approximation intervals of a specific receiver can be entered. The latter option will use an additional fourteen (14) bytes from the OEM scratch pad A6h:208-221(DO$_h$-DD$_h$). OEMCFG6 bits [6:5] are used to select between these options. The memory maps for the calibration coefficients are shown in Tables 11 and 12. If the programmable delimiters option is selected, the user must enter the seven delimiters of the intervals that best fit the receiver response. The diagram in Figure 3 shows the link between the delimiters and the sets of slopes and offsets.

Slopes Coefficients

The slopes allow for the correction of gain errors. Each slope coefficient is an unsigned, sixteen-bit, fixed-point binary number in the format:

[mmmmmmmm.llllllll]   **(5)**

where m is a data bit in the most-significant byte and l is a data bit in the least significant byte

Slopes are always positive. The decimal point is in between the two bytes, i.e., between bits 7 and 8. This provides a numerical range of 1/256 (0.00391) to 255.997 in steps of 1/256. The most significant byte is always stored in memory at the lower numerical address.

Offset coefficients

The offsets correct for constant errors in the measured data. Each offset, apart from temperature, is a signed, sixteen-bit, fixed-point binary number. The bit-weights of the offsets are the same as that of the final results. The sixteen-bit offsets provide a numerical range of –32768 to +32767 for voltage, bias current, transmit power, and receive power.

The numerical range for the six-bit temperature offset is –32 (–16 °C) to +31 (+15.5 °C) in increments of .5 °C. The two most significant bits of the temperature offset coefficient are ignored by the MIC3003.

Computing Internal Calibration Results

Calibration of voltage, bias current, and TX power are performed using the following calculation:

$$RESULTn = ADC\_RESULTn \times SLOPEn + OFFSETn$$

(6)

Calibration of RX power is performed using the following calculation:

$$RESULT = ADC\_RESULT \times SLOPE(m) + OFFSET(m)$$

(7)

where m represents one of the eight linearization intervals corresponding to the RX power level.

The results of these calculations are rounded to sixteen bits. If the seventeenth bit is a one, the result is rounded up to the next higher value. If the seventeenth bit is zero, the upper sixteen bits remain unchanged. The bit-weights of the offsets are the same as that of the final results. For SFF-8472 compatible applications, these bit-weights are given in Table 10.

Parameter	Magnitude of LSB
Voltage	100µV
Bias Current	2µA
TX Power	0.1µW
RX Power	0.1µW

Table 10. LSB Values of Offset Coefficients

Address(s)		Field Size	Name	Description
HEX	**DEC**			
48-49	72-73	2	RESERVED	Reserved. There is no slope for temperature. Do not write; reads undefined.
4A-4B	74-75	2	TOFFh:TOFFl	Temperature offset; signed six-bit integer offset with an LSB resolution of .5 degrees C per bit. The two most significant bits of TOFFh are ignored. TOFFl is not used.
				Note that TOFFh is also used in external calibration mode.
4C-4D	76-77	2	VSLPh:VSLPl	Voltage slope; unsigned fixed-point; MSB is at lower physical address.
4E-4F	78-79	2	VOFFh:VOFFl	Voltage offset; signed integer; MSB is at lower physical address.
50-51	80-81	2	ISLPh:ISLPl	Bias current slope; unsigned fixed-point; MSB is at lower physical address.
52-53	82-83	2	IOFFh:IOFFl	Bias current offset; signed integer; MSB is at lower physical address.
54-55	84-85	2	TXSLPh: XSLPl	TX power slope; unsigned fixed-point; MSB is at lower physical address.
56-57	86-87	2	TXOFFh: TXOFFl	TX power slope; unsigned fixed-point; MSB is at lower physical address.

Table 11. Internal Calibration Coefficient Memory Map – Part I

Address(s)		Field Size	Name	Description
HEX	**DEC**			
28-29	40-41	2	RXSLP0h: RXSLP0l	RX power slope 0; unsigned fixed-point; MSB is at lower physical address.
2A-2B	42-43	2	RXOFF0h: RXOFF0l	RX power offset 0; signed integer; MSB is at lower physical address.
2C-2D	44-45	2	RXSLP1h: RXSLP1l	RX power slope 1; unsigned fixed-point; MSB is at lower physical address.
2E-2F	46-47	2	RXOFF1h: RXOFF1l	RX power offset 1; signed integer; MSB is at lower physical address.
30-31	48-49	2	RXSLP2h: RXSLP2l	RX power slope 2; unsigned fixed-point; MSB is at lower physical address.
32-33	50-51	2	RXOFF2h: RXOFF2l	RX power offset 2; signed integer; MSB is at lower physical address.
34-35	52-53	2	RXSLP3h: RXSLP3l	RX power slope 3; unsigned fixed-point; MSB is at lower physical address.
36-37	54-55	2	RXOFF3h: RXOFF3l	RX power offset 3; signed integer; MSB is at lower physical address.
38-39	56-57	2	RXSLP4h: RXSLP4l	RX power slope 4; unsigned fixed-point; MSB is at lower physical address.
3A-3B	58-59	2	RXOFF4h: RXOFF4l	RX power offset 4; signed integer; MSB is at lower physical address.
3C-3D	60-61	2	RXSLP5h: RXSLP5l	RX power slope 5; unsigned fixed-point; MSB is at lower physical address.
3E-3F	62-63	2	RXOFF5h: RXOFF5l	RX power offset 5; signed integer; MSB is at lower physical address.
40-41	64-65	2	RXSLP6h: RXSLP6l	RX power slope 6; unsigned fixed-point; MSB is at lower physical address.
42-43	66-67	2	RXOFF6h: RXOFF6l	RX power offset 6; signed integer; MSB is at lower physical address.
44-45	68-69	2	RXSLP7h: RXSLP7l	RX power slope 7; signed integer; MSB is at lower physical address.
46-47	70-71	2	RXOFF7h: RXOFF7l	RX power offset 7; signed fixed-point; MSB is at lower physical address.

Table 12. Internal Calibration Coefficient Memory Map – Part II

Address(s) HEX	Address(s) DEC	Name
28-29	40-41	RXSLP0h: RXSLP0l
2A-2B	42-43	RXOFF0h: RXOFF0l
2C-2D	44-45	RXSLP1h: RXSLP1l
2E-2F	46-47	RXOFF1h: RXOFF1l
30-31	48-49	RXSLP2h: RXSLP2l
32-33	50-51	RXOFF2h: RXOFF2l
34-35	52-53	RXSLP3h: RXSLP3l
36-37	54-55	RXOFF3h: RXOFF3l
38-39	56-57	RXSLP4h: RXSLP4l
3A-3B	58-59	RXOFF4h: RXOFF4l
3C-3D	60-61	RXSLP5h: RXSLP5l
3E-3F	62-63	RXOFF5h: RXOFF5l
40-41	64-65	RXSLP6h: RXSLP6l
42-43	66-67	RXOFF6h: RXOFF6l
44-45	68-69	RXOFF7h: RXOFF7l
46-47	70-71	RXSLP7h: RXSLP7l

Conditions:
- RXPWR <= Delimiter #1
- Delimiter #1 < RXPWR <= Delimiter #2
- Delimiter #2 < RXPWR <= Delimiter # 3
- Delimiter #3 < RXPWR <= Delimiter #4
- Delimiter #4 < RXPWR <= Delimiter #5
- Delimiter #5 < RXPWR <= Delimiter #6
- Delimiter #6 < RXPWR <= Delimiter #7
- RXPWR > Delimiter #7

Delimiter	Address(s) HEX	Address(s) DEC
Delimiter #1	D0-D1	208-209
Delimiter #2	D2-D3	210-211
Delimiter #3	D4-D5	212-213
Delimiter #4	D6-D7	214-215
Delimiter #5	D8-D9	216-217
Delimiter #6	DA-DB	218-219
Delimiter #7	DC-DD	220-221

Figure 3. Internal Calibration RX Power Linear Approximation

C/ Reading the ADC Result Registers

The ADC result registers should be read as 16-bit registers under internal calibration while under external calibration they should be read as 8-bit or 16-bit registers at the MSB address. For example, TX power should be read under internal calibration as 16 bits at address A2h: 66h–67h and under external calibration as 8 bits at address A2h: 66h. 9-bit temperature results and 12-bit receive power results should always be read as 16-bit quantities.

Reading the result registers using two-byte burst reads on the SMBus guarantees that the two bytes are coherent with each other—that is, they form a matched result pair. If the two bytes were read separately, it is possible that the internal result could be updated between the reads, leading to an incorrect ADC result.

RXPOT

A programmable, non-volatile digitally controlled potentiometer is provided for adjusting the gain of the receive power measurement signal chain in the analog domain. Five bits in the RXPOT register are used to set and adjust the position of potentiometer. RXPOT functions as a programmable divider or attenuator. It is adjustable in steps from 1:1 (no divider action) down to 1/32 in steps of 1/32. If RXPOT is set to zero, then the divider is bypassed completely. There will be no scaling of the input signal, and the resistor network will be disconnected from the VRX pin. At all other settings of RXPOT, there will be a 32kΩ (typical) load seen on VRX.
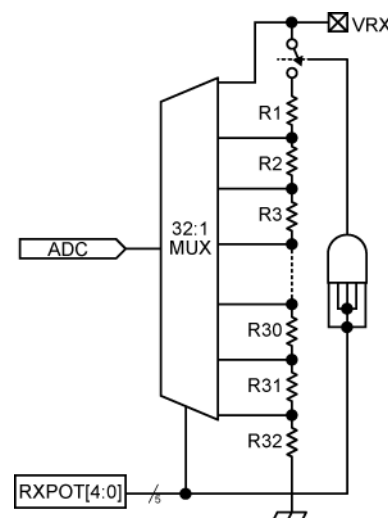
Figure 4. RXPOT Block Diagram

Laser Diode Bias Control

The MIC3003 can be configured to generate a constant bias current using electrical feedback, or regulate average transmitted optical power using a feedback signal from a monitor photodiode, as shown in Figure 5. An operational amplifier is used to control laser bias current via the V$_{BIAS}$ output. The VBIAS pin can drive a maximum of ±10mA. An external bipolar transistor provides current gain. The polarity of the op amp's output is programmable with BIASREF (bit-5 in OEMCFG1) in order to accommodate either NPN or PNP transistors that drive common anode and common cathode laser, respectively. Additionally, the polarity of the feedback signal is programmable for use with either common-emitter or emitter-follower transistor circuits.

Furthermore, the reference level for the APC circuit is selectable to accommodate electrical, i.e., current feedback, or optical feedback via a monitor photodiode. Finally, any one of seven different internal feedback resistors can be selected. This internal resistor can be used alone or in parallel with an external resistor. This wide range of adjustability (50:1) accommodates a wide range of photodiode current, i.e., wide range of transmitter output power. The APC operating point can be kept near the mid-scale value of the APC DAC, insuring maximum SNR, maximum effective resolution for digital diagnostics, and the widest possible DAC adjustment range for temperature compensation, etc. See Figure 6.

The APCCAL bit in OEMCAL0 is used to turn the APC function on and off. It will be turned on in the MIC3003's default state as shipped from the factory. When the APC is on, the value in the selected APCSETx register is added to the signed compensation value taken from the APC look-up table and loaded into the V_{BIAS} DAC. When the APC is off, the V_{BIAS} DAC may be written directly via the VBIAS register, bypassing the look-up table entirely. This provides direct control of the laser diode bias during setup and calibration. In either case, the V_{BIAS} DAC setting is reported in the APCDAC register.
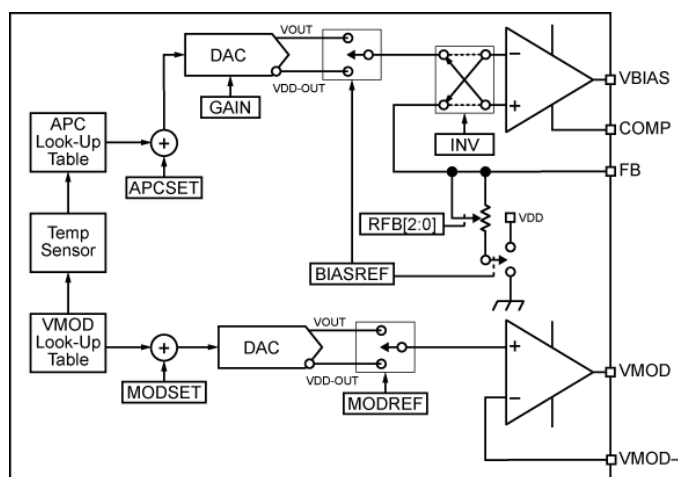
Figure 5. APC and Modulation Control Block Diagram
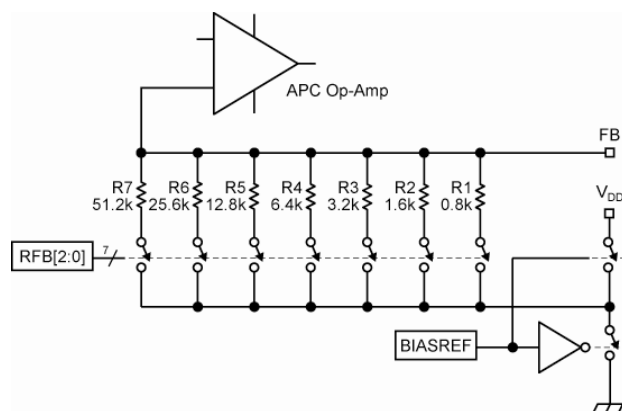
Figure 6. Programmable Feedback Resistor

Laser Modulation Control

As shown in Figure 5, a temperature-compensated DAC is provided to set and control the laser modulation current via an external laser driver circuit. The MODREF bit in OEMCFG0 selects whether the V_{MOD} DAC output swings up from ground or down from V_{DD}. If the laser driver requires a voltage input to set the modulation current, V_{MOD} output can drive it directly. If a current input is required, a fixed resistor can be used between the driver and the V_{MOD} output. Several different configurations are possible as shown in Figure 8.

When the APC is on, i.e., the APCCAL bit in OEMCAL0 is set to 0, the value corresponding to the current temperature is taken from the MODLUT look-up table, added to the selected MODSETx register, and loaded into the V_{MOD} DAC. When the APC is off, the compensation value in VMOD is loaded directly into the V_{MOD} DAC, bypassing the look-up table entirely. This provides for direct modulation control for setup and calibration.
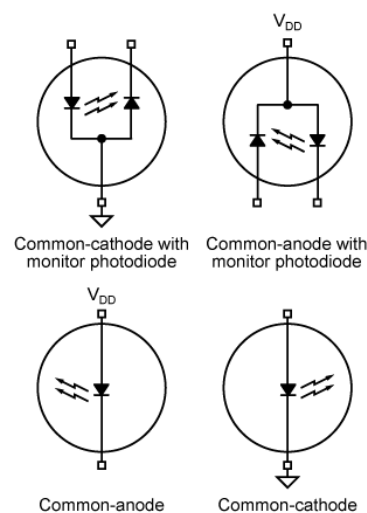
Figure 7. Transmitter Configurations Supported by MIC3003
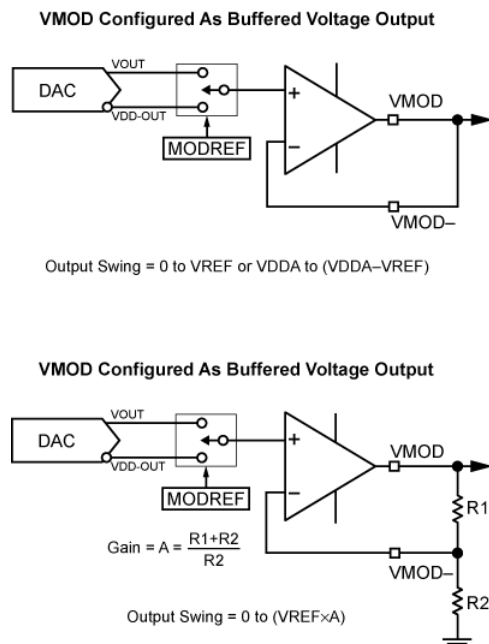
Figure 8. V$_{MOD}$ Configured as Voltage Output
with Gain

Power On and Laser Start-Up

When power is applied, the MIC3003 initializes its internal registers and state machine. This process takes t$_{POR}$, about 50ms. Following t$_{POR}$, analog-to-digital conversions begin, serial communication is possible, and the POR bit and data ready bits may be polled. The first set of analog data will be available t$_{CONV}$ after t$_{POR}$. MIC3003s are shipped from the factory with the output enable bit, OE, set to zero, off. The power-up default state, therefore, is APC off, V$_{BIAS}$, V$_{MOD}$, and SHDN outputs disabled. V$_{BIAS}$, V$_{MOD}$, and SHDN will be floating (high impedance) and the laser diode, if connected, will be off. Once the device is incorporated into a transceiver and properly configured, then the shutdown states of SHDN, V$_{BIAS}$, and V$_{MOD}$ will be determined by the state of the APC configuration and OE bits. Tables 13, 14, and 15 illustrate the shutdown states of the various laser control outputs versus the control bits.

Configuration Bits		Shutdown State
OE	SPOL	SHDN
0	Don't Care	Hi-Z
1	0	GND
1	1	V$_{DD}$

Table 13. Shutdown State of SHDN vs.
Configuration Bits

Configuration Bits			V$_{BIAS}$ Shutdown State
OE	INV	BIASREF	V$_{BIAS}$
0	Don't Care	Don't Care	Hi-Z
1	Don't Care	0	GND
1	Don't Care	1	V$_{DD}$

Table 14. Shutdown State of V$_{BIAS}$ vs.
Configuration Bits

Configuration Bits		V$_{MOD}$ Shutdown State
OE	MODREF	V$_{MOD}$
0	Don't Care	Hi-Z
1	0	GND
1	1	V$_{DD}$

Table 15. Shutdown State of V$_{MOD}$ vs.
Configuration Bits

In order to facilitate hot-plugging, the laser diode is not turned on until t$_{INIT2}$ after Power-On. Following t$_{INIT2}$, and assuming TXDISABLE is not asserted, the DACs will be loaded with their initial values. Since tCONV is much less than t$_{INIT2}$, the first set of analog data, including temperature, is available at t$_{INIT2}$. Temperature compensation will be applied to the DAC values if enabled. APC will begin if OE is asserted. (If the output enable bit, OE, is not set, the V$_{MOD}$, V$_{BIAS}$, and SHDN outputs will float indefinitely.) Figure 9 shows the power-up timing of the MIC3003. If TXDISABLE is asserted at power-up, the V$_{MOD}$ and V$_{BIAS}$ outputs will stay in their shutdown states following MIC3003 initialization. A/D conversions will begin, but the laser will remain off.
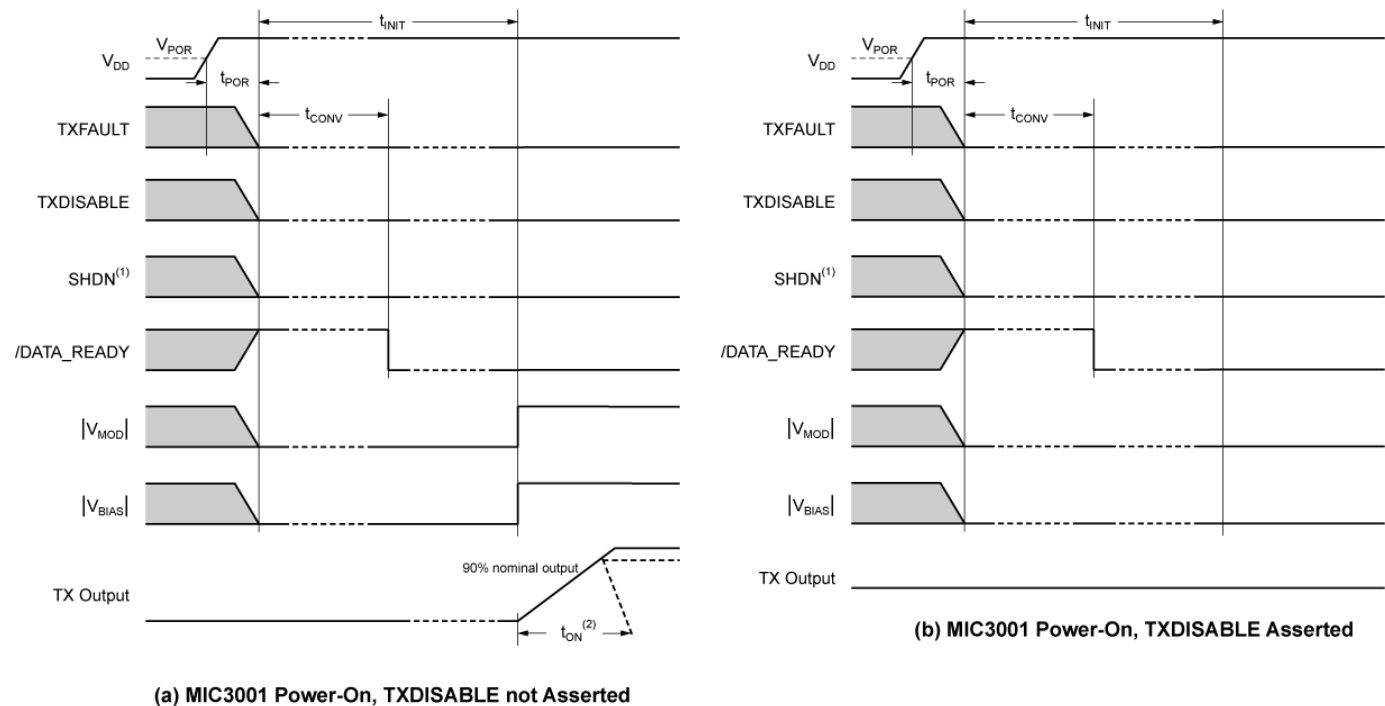
Figure 9. MIC3003 Power-On Timing (OE = 1)

Fault Comparators

In addition to detecting and reporting the events specified in SFF-8472, the MIC3003 also monitors five fault conditions: inadequate supply voltage, thermal diode faults, excessive bias current, excessive transmit power, and APC op-amp saturation. Comparators monitor these parameters in order to respond quickly to fault conditions that could indicate link failure or safety issues, see Figure 10. When a fault is detected, the laser is shut down and TXFAULT is asserted. Each fault source may be independently disabled using the FLTMSK register. FLTMSK is non-volatile, allowing faults to be masked only during calibration and testing or permanently.
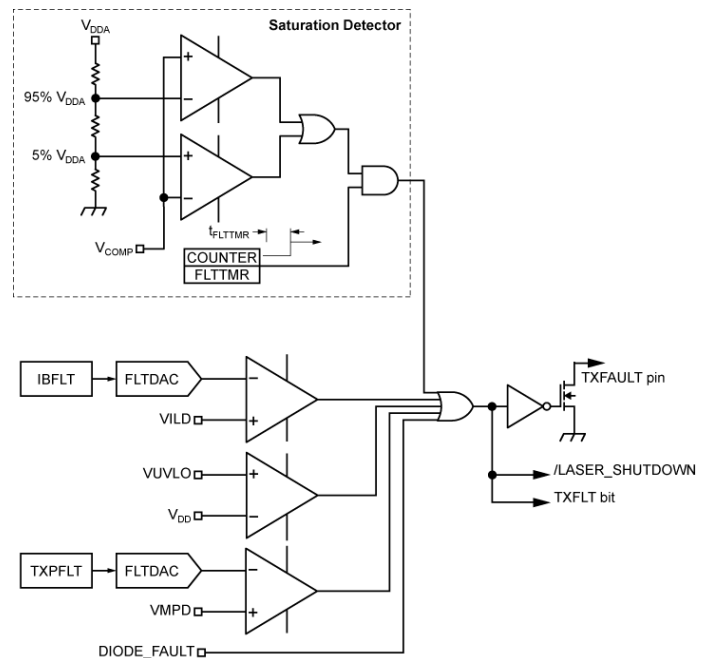
Figure 10. Fault Comparator Logic

Thermal diode faults are detected within the temperature measurement subsystem when an out-of-range signal is detected. A window comparator circuit monitors the voltage on the compensation capacitor to detect APC op-amp saturation (Figure 11). Op-amp saturation indicates that some fault has occurred in the control loop such as loss of feedback. The saturation detector is blanked for a time, tFLTTMR, following laser turn-on since the compensation voltage will essentially be zero at turn-on. The FLTTMR interval is programmable from 0.5ms to 127.5ms (typical) in increments of 0.5ms (tFLTTMR). Note that a saturation comparator cannot be relied upon to meet certain eye-safety standards that require 100ms response times. This is because the operation of a saturation detector is limited by the loop bandwidth, i.e., the choice of C_{COMP}. Even if the comparator itself was very fast, it would be subject to the limited slew-rate of the APC op-amp. Only the other fault comparator channels will meet <100ms timing requirements.

The MIC3003 can also except and respond to fault inputs from external devices. See the "SHDN and TXFIN" section.

A similar comparator circuit monitors received signal strength and asserts RXLOS when loss-of-signal is detected (Figure 12). RXLOS will be asserted if VRX drops below the level programmed in LOSFLT. Hysteresis is implemented such that RXLOS will be de-asserted when VRX subsequently rises above the level programmed in LOSFLTn. The loss-of-signal comparator may be disabled completely by setting the LOSDIS bit in OEMCFG3. Once the LOS comparator is disabled, an external device may drive RXLOS. The state of the RXLOS pin is reported in the CNTRL register regardless of whether it is driven by the internal comparator or by an external device. A programmable digital-to-analog converter provides the comparator reference voltages for monitoring received signal strength, transmit power, and bias current. Since laser bias current varies greatly with temperature, there is a temperature compensation look-up table for the bias current fault DAC value.

When a fault condition is detected, the laser will be shutdown immediately and TXFAULT will be asserted. The V_{MOD}, V_{BIAS}, and SHDN (if enabled by setting OEMCFG5 bit 7 to 1) outputs will be driven to their shutdown state according to the state of the configuration bits. The shutdown states of V_{MOD}, V_{BIAS}, and SHDN versus the configuration bit settings are shown in Table 12, Table 13, and Table 14.

SHDN and TXFIN

SHDN and TXFIN are optional functions of pin 12. SHDN is an output function and is designed to drive a redundant safety switch in the laser current path. TXFIN is an input function and serves as an input for fault signals from external devices that must be reported to the host via TXFAULT. The SHDN function is designed for

applications in which the MIC3003 is performing all APC and laser management tasks. The TXFIN function is for situations in which an external device such as a laser diode driver IC is performing laser management tasks, including fault detection.

If the TXFIN bit in OEMCFG3 is zero (the default mode), SHDN will be activated anytime the laser is off. Thus, it will be active if 1) TXDISABLE is asserted, 2) STXDIS in the CNTRL register, is set, or 3) a fault is detected. SHDN is a push-pull logic output. Its polarity is programmable via the SPOL bit in OEMCFG1.

If TXFIN bit is set to one, pin 12 serves as an input that accepts fault signals from external devices such as laser diode driver ICs. Multiple TXFAULT signals cannot simply be wire-OR'ed together as they are open-drain and active high. The input polarity is programmable via the TXFPOL bit in OEMCFG3. TXFIN is logically OR'ed with the MIC3003's internal fault sources to produce TXFAULT and determine the value of the transmit fault bit in CNTRL. See Figure 10.
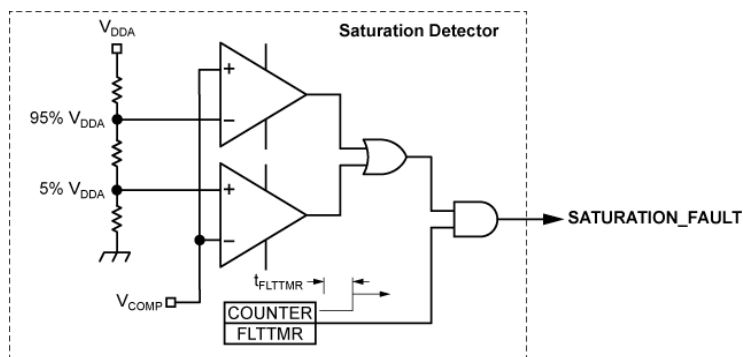
Figure 11. Saturation Detector
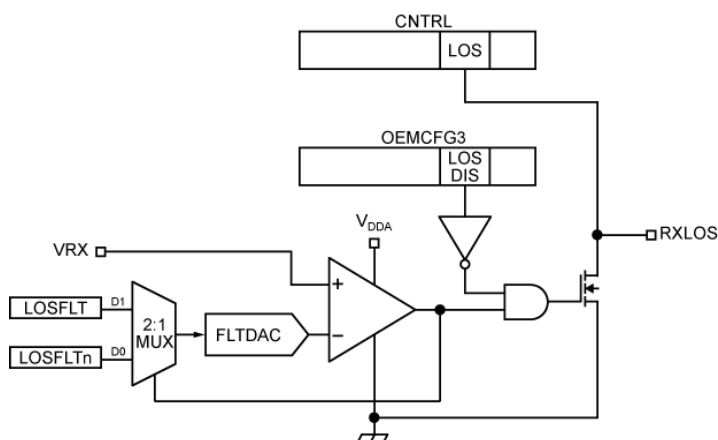
Figure 12. RXLOS Comparator Logic

Temperature Measurement

The temperature-to-digital converter is built around a switched current source and an eight-bit/nine-bit analog-to-digital converter. The temperature is calculated by measuring the forward voltage of a diode junction at two different bias current levels. An internal multiplexer directs the current source's output to a diode junction. This data is also used as the input to the temperature compensation look-up tables. Each time temperature is sampled and an updated value acquired, new corrective values for modulation current and the APC setpoint are read from the corresponding tables, added to the set values, and transferred to the DACs.

Diode Faults

The MIC3003 is designed to respond in a failsafe manner to hardware faults in the temperature sensing circuitry. If there is a fault with the on-chip sensing diode, the temperature data reported by the A/D converter will be forced to its full-scale value (+127 °C). The diode fault flag, DFLT, will be set in OEMCFG0, TXFAULT will be asserted, and the high temperature alarm and warning flags will be set. The reported temperature will remain at +127 °C until the fault condition is cleared. Diode faults may be reset by toggling TXDISABLE, as with any other fault. Diode faults will not be detected at power up until the first A/D conversion cycle is completed.

Temperature Compensation

Since the performance characteristics of laser diodes and photodiodes change with operating temperature, the MIC3003 provides a facility for temperature compensation of the APC. loop set-point, laser modulation current, bias current fault comparator threshold, and bias current high alarm flag threshold. Temperature compensation is performed using a look-up table (LUT) that stores values corresponding to each measured temperature over a 150°C span. Four identical tables reside at serial address A4h and A6h as summarized in Table 16. Each table entry is a signed twos complement integer that is used as an offset to the parameter being compensated. The default value of all table entries is zero, giving a flat response.

The A/D converter reports a new temperature sample each t_{CONV}. This occurs at roughly 10 Hz when 8-bit temperature resolution is selected. To prevent temperature oscillation due to thermal or electrical noise, sixteen successive temperature samples are averaged together and used to index the LUT.s. Temperature compensation results are therefore updated at $16 \times t_{CONV}$ intervals, or about 1.6 seconds. This can be expressed as shown in Equation 8:

$$T_{COMPm} = \frac{T_n + T_{n+1} + T_{n+2} + \cdots T_{n+15}}{16} \qquad \textbf{(8)}$$

Each time an updated average value is acquired, a new offset value for the APC setpoint is read from the corresponding look-up table (see Table 17) and transferred to the APC circuitry. This is illustrated in Equation 11. In a same way, new offset values are taken from similar look-up tables (see Table 18 and Table 19), added to the nominal values and transferred into the modulation and fault comparator DACs. The bias current high alarm threshold is compensated using a fourth look-up table (see Table 20). This compensation happens internally and does not affect any host-accessible registers.

$$APCSETm = APCSETx + APCLUT(T_{COMPm})$$

$$Table\_min \le T_{COMPm} \le Table\_max$$

$$APCSETm = APCSETx + APCLUT(max)$$

$$T_{COMP} > Table\_max$$

$$APCSETm = APCSETx + APCLUT(min)$$

$$T_{COMP} < Table\_min$$

$$\textbf{(9)}$$

If the measured temperature is greater than the maximum table value, the highest value in each table is used. If the measured temperature is less than the minimum, the minimum value is used. Hysteresis is employed to further enhance noise immunity and prevent oscillation. Each table entry spans two degrees C. The table index will not change unless the new temperature average results in a table index beyond the midpoint of the next entry in either direction. There is therefore 2 to 3°C of hysteresis on temperature compensation changes. The table index will never oscillate due to quantization noise as the hysteresis is much larger than ±1⁄2 LSB.

Serial Address	Byte Addresses	Function
Base address +4h	00h–3Fh	APC Look-up Table
	40h–7Fh	IMOD Look-up Table
	80h–BFh	IFLT Look-up Table
	C0h–FFh	Bias High Alarm Look-up Table
Base address +6h	58h–63h	APC Look-up Table (cont.)
	64h–6Fh	IMOD Look-up Table (cont.)
	70h–7Bh	IFLT Look-up Table (cont.)
	7Ch–87h	Bias High Alarm Look-up Table (cont.)

Table 16. Temperature Compensation Look-up Tables

Serial Address	Register Address	Table Offset	Temperature Offset (°C)
Base address +4h	00h	0	≤ -45
	01h	1	-44
			-43
	•	•	•
	•	•	•
	•	•	•
	3Fh	63	80
			81
Base address +6h	64h	64	82
			83
	•	•	•
	•	•	•
	6E	74	102
			103
	6F	75	≥ 104

Table 17. APC Temperature Compensation Look-Up Table

Serial Address	Register Address	Table Offset	Temperature Offset (°C)
Base address +4h	80h		≤ -45
	81h		-44
			-43
	82h		
	•	•	•
	•	•	•
	•	•	•
	BEh		
	BFh	63	80
			81
Base address +6h	58	64	82
			83
	•	•	•
	•	•	•
	62	74	102
			103
	63	75	≥ 104

Table 19. I$_{BIAS}$ Comparator Temperature Compensation Look-Up Table

Serial Address	Register Address	Table Offset	Temperature Offset (°C)
Base address +4h	40h	0	≤ -45
	41h	1	-44
			-43
	•	•	•
	•	•	•
	•	•	•
	7Fh	63	80
			81
Base address +6h	70	64	82
			83
	•	•	•
	•	•	•
	7A	74	102
			103
	7B	75	≥ 104

Table 18. I$_{MOD}$ Temperature Compensation Look-Up Table

Serial Address	Register Address	Table Offset	Temperature Offset (°C)
Base address +4h	C0h		≤ -45
	C1h		-44
			-43
	C2h		
	•	•	•
	•	•	•
	•	•	•
	FEh		
	FFh	63	80
			81
Base address +6h	7C	64	82
			83
	•	•	•
	•	•	•
	87	74	102
			103

Table 20. BIAS Current High Alarm Temperature Compensation Table

The internal state machine calculates a new table index each time a new average temperature value becomes available. This table index is derived from the average temperature value. The table index is then converted into a table address for each of the four look-up tables. These operations can be expressed as:

$$INDEX = \left| \frac{T_{AVG(n)}}{2} \right| \tag{10}$$

where TAVG(n) is the current average temperature; and

$$TABLE\_ADDRESS = INDEX + BASE\_ADDRESS$$

where BASE_ADDRESS is the physical base address of each table, i.e., 00h, 20h, 40h, 80h, or 60h (tables reside in the Base address + 4h and Base address + 6h pages of memory).

At any given time, the current table index can be read in the LUTINDX register.

Alarms and Warning Flags

There are 20 different conditions that will cause the MIC3003 to set one of the bits in the WARNx or ALARMx registers. These conditions are listed in Table 21. The less critical of these events generate warning flags by setting a bit in WARN0 or WARN1. The more critical events cause bits to be set in ALARM0 or ALARM1.

An event occurs when any alarm or warning condition becomes true. Each event causes its corresponding status bit in ALARM0, ALARM1, WARN0, or WARN1 to be set. This action cannot be masked by the host. IF OEMCFG-4 bits [7-4] are set to 0 (default value), the status bit will remain set until the host reads that particular status register, a power on-off cycle occurs, or the host toggles TXDISABLE.

If TXDISABLE is asserted at any time during normal operation, A/D conversions continue. The A/D results for all parameters will continue to be reported. All events will be reported in the normal way. If they have not already been individually cleared by read operations, when TXDISABLE is de-asserted, all status registers will be cleared.

Control and Status I/O

The logic for the transceiver control and status I/O is shown schematically in Figure 13. Note that the internal drivers on RXLOS/TRSOUT, RRSOUT/GPO, QGPO, and TXFAULT are all open-drain. These signals may be driven either by the internal logic or external drivers connected to the corresponding MIC3003 pins. In any case, the signal level appearing at the pins of the MIC3003 will be reported in the control register status bits.

Note that the control bits for TX_DISABLE and RRSOUT, TRSOUT, and the status bits for TXFAULT and RXLOS do not meet the timing requirements as specified in the SFP MSA or the GBIC Specification, revision 5.5 (SFF-8053) for the hardware signals. The speed of the SMBus serial interface limits the rate at which these functions can be manipulated and/or reported. The response time for the control and status bits is given in the "Electrical Characteristics" subsection.

Event	Condition	MIC3003 Response
Temperature high alarm	TEMP > TMAX	Set ALARM0[7]
Temperature low alarm	TEMP < TMIN	Set ALARM0[6]
Voltage high alarm	VIN > VMAX	Set ALARM0[5]
Voltage low alarm	VIN < VMIN	Set ALARM0[4]
TX bias high alarm	IBIAS > IBMAX	Set ALARM0[3]
TX bias low alarm	IBIAS < IBMIN	Set ALARM0[2]
TX power high alarm	TXOP > TXMAX	Set ALARM0[1]
TX power low alarm	TXOP < TXMIN	Set ALARM0[0]
RX power high alarm	RXOP > RXMAX	Set ALARM1[7]
RX power low alarm	RXOP < RXMIN	Set ALARM1[6]
Temperature high warning	TEMP > THIGH	Set WARN0[7]
Temperature low warning	TEMP < TLOW	Set WARN0[6]
Voltage high warning	VIN > VHIGH	Set WARN0[5]
Voltage low warning	VIN < VLOW	Set WARN0[4]
TX bias high warning	IBIAS > IBHIGH	Set WARN0[3]
TX bias low warning	IBIAS < IBLOW	Set WARN0[2]
TX power high warning	TXOP > TXHIGH	Set WARN0[1]
TX power low warning	TXOP < TXLOW	Set WARN0[0]
RX power high warning	RXOP > RXHIGH	Set WARN1[7]
RX power low warning	RXOP < RXLOW	Set WARN1[6]

Table 21. MIC3003 Alarm and Warning Events
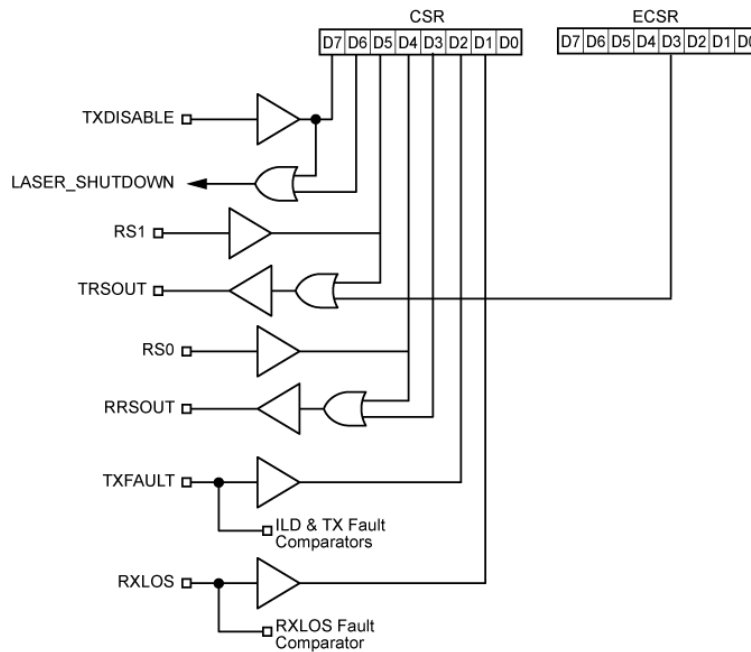
Figure 13. Control and Status I/O Logi

System Timing

The timing specifications for MIC3003 control and status I/O are given in the "Electrical Characteristics" subsection.
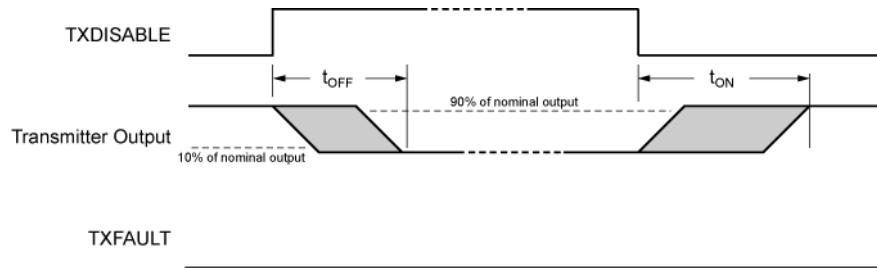
Figure 14. Transmitter On-Off Timing
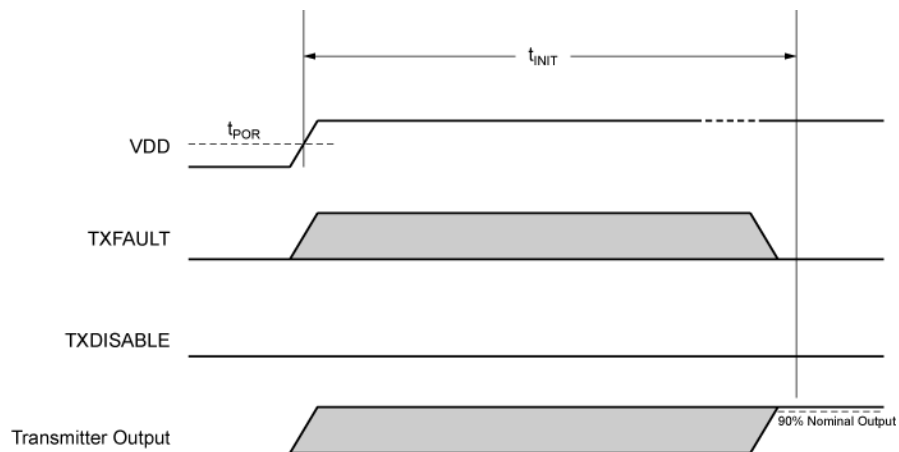
Figure 15. Initialization Timing with TXDISABLE Asserted
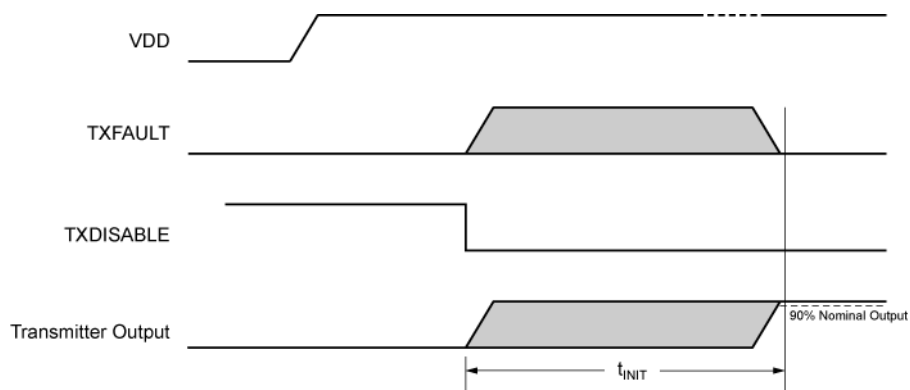
Figure 16.   Initialization Timing with TXDISABLE Not Asserted
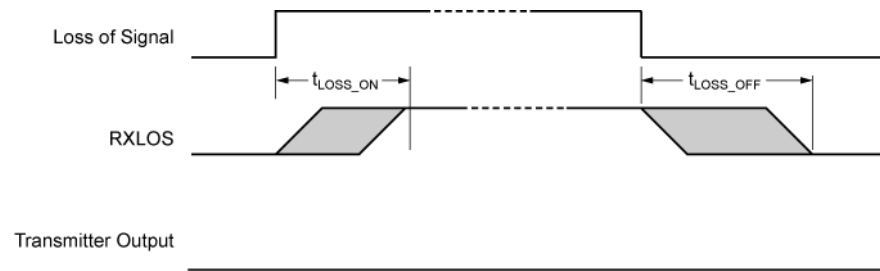
Figure 17. Loss-of-Signal (LOS) Timing
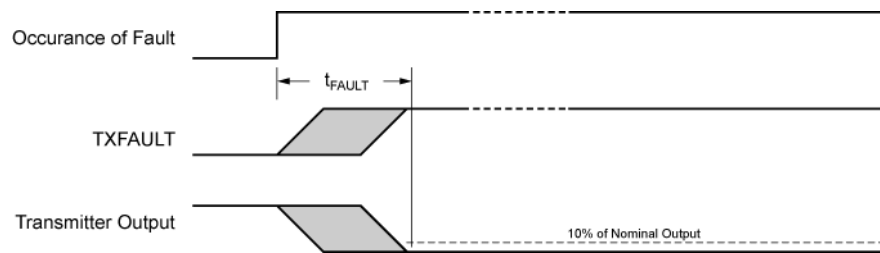
Figure 18. Transmit Fault Timing
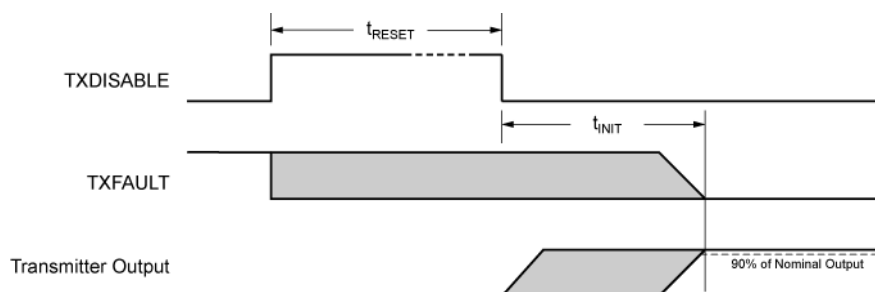
Figure 19. Successfully Clearing a Fault Condition
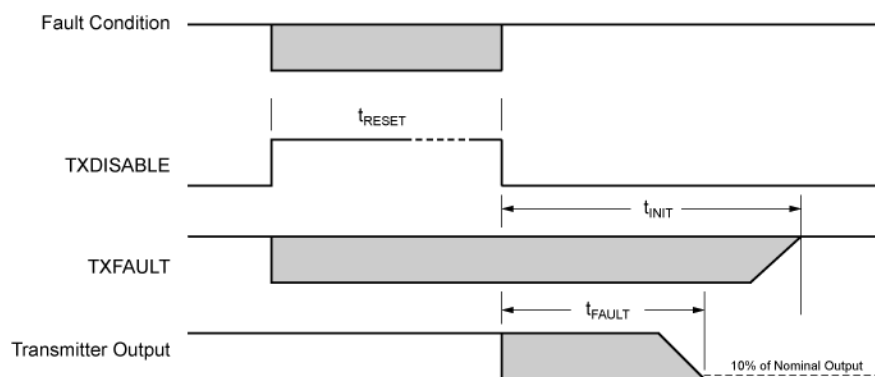
Figure 20. Unsuccessful Attempt to Clear a Fault

Warm Resets

The MIC3003 can be reset to its power-on default state during operation by setting the RST bit in OEMCFG0. When this bit is set, TXFAULT and RXLOS will be de-asserted, all registers will be restored to their normal power-on default values, and any A/D conversion in progress will be halted and the results discarded. The state of the MIC3003 following this operation is indistinguishable from a power-on reset.

Power-On Hour Meter

The Power-On Hour meter logs operating hours using an internal real-time clock and stores the result in NVRAM. The hour count is incremented at ten-hour intervals in the middle of each interval. The first increment therefore takes place five hours after power-on. Time is accumulated whenever the MIC3003 is powered. The hour meter's time base is accurate to +/-10% over all MIC3003 operating conditions, and is accurate to +/-5% in the range 0 to 70 degrees C. The counter is capable of storing counts of more than thirty years, but is ultimately limited by the write-cycle endurance of the non-volatile memory. This implies a range of at least twenty years.

Actual results will depend upon the operating conditions and write-cycle endurance of the part in question.

Two registers, POHh and POHl, contain a 15-bit power-on hour measurement and an error flag, POHFLT. Great care has been taken to make the MIC3003's hour meter immune to data corruption and to insure that valid data is maintained across power cycles. The hour meter employs multiple data copies and error correction codes to maintain data validity. This data is stored in the POHDATA registers. If POHFLT is set, however, the power-on hour meter data has been corrupted and should be ignored.

It is recommended that a two-byte sequential read operation be performed on POHh and POHl to insure coherency between the two registers. These registers are accessible by the OEM using a valid OEM password. The only operation that should be performed on these registers is to clear the hour meters initial value, if necessary, at the time of product shipment. The hour meter result may be cleared by setting all eight POHDATA bytes to 00_h.

Test and Calibration Features

Numerous features are included in the MIC3003 to facilitate development, testing, and diagnostics. These features are available via registers in the OEM area. As shown in Table 22, these features include:

Function	Description	Control Register(s)
Analog loop-back	Provides analog visibility of op-amp and DAC outputs via the ADC	OEMCFG0
Fault comparator disable control	Disables the fault comparator	OEMCAL0
Fault comparator spin-on-channel mode	Selects a single fault comparator channel	OEMCAL0
Fault comparator output read-back	Allows host to read individual fault comparator outputs	OEMRD
TRSOUT, /INT read-back	Allows host to read the state of these pins	OEMRD
Inhibit EEPROM write cycles	Speeds repetitive writes to registers backed up by NVRAM	OEMCAL0
APC calibration mode	Allows direct writes to MODDAC and APCDAC (temperature compensation not used)	OEMCAL0
Continuity checking	Forcing of RXLOS, TXFAULT, /INT	OEMCAL0
Halt A/D	Stops A/D conversions; ADC in one-shot mode	OEMCAL1
ADC idle flag	Indicates ADC status	OEMCAL1
A/D one-shot mode	Performs a single A/D conversion on the selected input channel	OEMCAL1
A/D spin-on-channel mode	Selects a single input channel	OEMCAL1
Channel selection	Selects ADC or fault comparator channel for spin-on-channel modes	OEMCAL1
LUT index read-back	Permits visibility of the LUT index calculated by the state-machine	LUTINDX
Manufacturer and device ID registers	Facilitates presence detection and version control	MFG_ID, DEV_ID

Table 22. Test and Diagnostic Features

Serial Port Operation

The MIC3003 uses standard write byte, read byte, and read word operations for communication with its host. It also supports block write and block read transactions. The write byte operation involves sending the devices address (with the R/W bit low to signal a write operation), followed by the address of the register to be operated upon and the data byte. The read byte operation is a composite write and read operation: the host first sends the devices address followed by the register address, as in a write operation. A new start bit must then be sent to the MIC3003, followed by a repeat of the device address with the R/W bit (LSB) set to the high (read) state. The data to be read from the part may then be clocked out. A read word is similar, but two successive data bytes are clocked out rather than one. These protocols are shown in Figures 21 to 24.

The MIC3003 will respond to up to four sequential device addresses depending upon whether it is in OEM or User mode. A match between one of the MIC3003's addresses and the address specified in the serial bit stream must be made to initiate communication. The MIC3003 responds to device addresses A0h and A2h in User Mode; it also responds to A4h and A6h in OEM Mode (assuming the base address is A0h).

Block Writes

To increase the speed of block writes, the MIC3003 allows up to eight consecutive bytes to be written before the internal memory update begins.

The block write sequence begins just like a write byte operation with the host sending the device address, R/W bit low, register address, etc. After the first data byte is sent the host will receive an acknowledge. Up to seven more bytes can be sent in sequence. The MIC3003 will acknowledge each one and increment its internal address register in anticipation of the next byte. After the last byte is sent, the host issues a STOP. The MIC3003's internal write process then begins.

Block writes of up to eight bytes can begin and end at any byte address without restriction. Block writes that increment over register address FFh will simply "wrap around" and continue at address 00h within the same device address space.

To accelerate calibration and testing, NVRAM write cycles can be disabled completely by setting the WRINH bit in OEMCAL0. Writes to registers that do not have NVRAM backup, will not incur write-cycle delays when writes are inhibited. Write operations on registers that exist only in NVRAM will still incur write cycle delays.
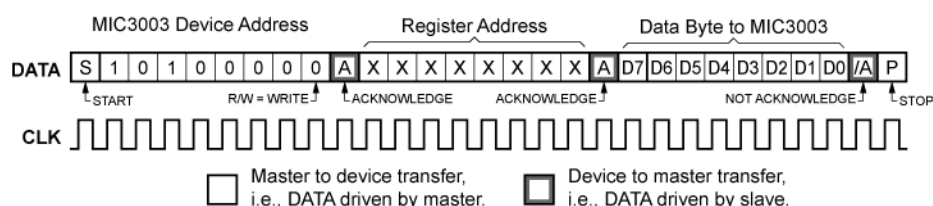
Figure 21. Write Byte Protocol
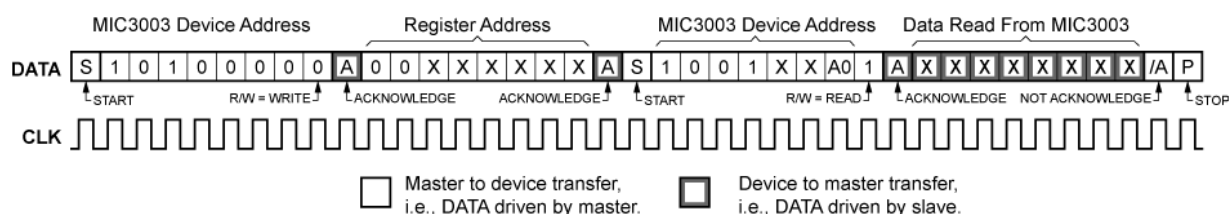
Figure 22. Read Byte Protocol
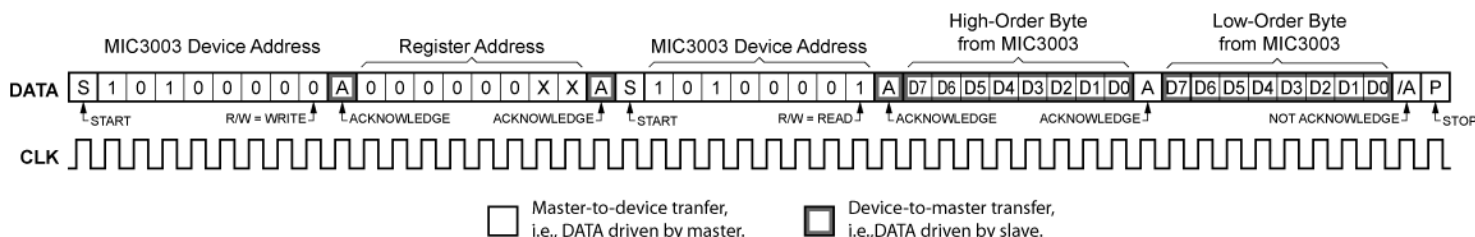
Figure 23. Read_Word Protocol
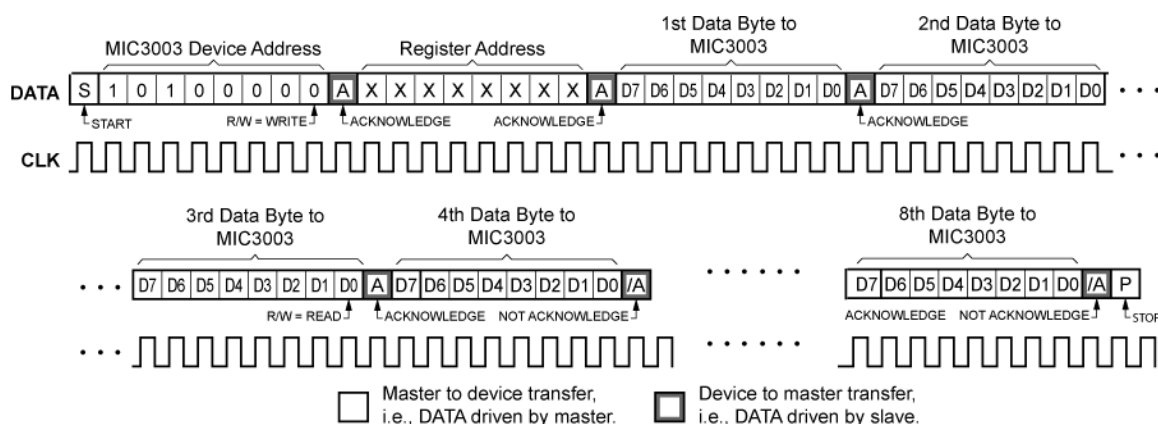
Figure 24. Eight-Byte Block Write Protocol

Acknowledge Polling

The MIC3003's non-volatile memory cannot be accessed during the internal memory update. To allow for maximum speed bulk writes, the MIC3003 supports acknowledge polling. The MIC3003 will not acknowledge serial bus transactions while internal writes are in progress. The host may therefore monitor for the end of the write process by periodically checking for an acknowledgement. The longest duration for the internal memory update to complete for a block write is approximately 26 ms.

Write Protection and Data Security

OEM Password

A password is required to access the OEM areas of the MIC3003, specifically the non-volatile memory, look-up tables, and registers at serial addresses A4$_h$ and A6$_h$. A four-byte field, OEMPWSET, at serial address A6$_h$ is used for setting the OEM password. The OEM password is set by writing OEMPWSET with the new value. The password comparison is performed following the write to the MSB of the OEMPW, address 7B$_h$ (or 7E$_h$ if OEMCFG5 bit 2 is set to 1) at serial address A2$_h$. Therefore, this byte must be written last. A four-byte burst-write sequence to address 78$_h$ (or 7B$_h$ if OEMCFG5 bit 2 is set to 1) may be used as this will result in the MSbyte being written last. New passwords written to the OEMPWSET registers will not take effect until after a power-on reset occurs or a warm reset is performed using the RST bit in OEMCFG0. This allows the new password to be verified before it takes effect.

The corresponding four-byte field for password entry, OEMPW, is located at serial address A2$_h$. This field is therefore always visible to the host system. OEMPW is compared to the four-byte OEMPWSET field at serial address A6$_h$. If the two fields match, access is allowed to the OEM areas of the MIC3003 non-volatile memory at serial addresses A4$_h$ and A6$_h$. If OEMPWSET is all zeroes, no password security will exist. The value in OEMPW will be ignored. This helps prevent a deliberately unsecured MIC3003 from being inadvertently locked. Once a valid password is entered, the MIC3003 OEM areas will be accessible. The OEM areas may be re-secured by writing an incorrect password value at OEMPW, e.g., all zeroes. In all cases, OEMPW must be written LSB first through MSB last. The OEM areas will be inaccessible following the final write operation to OEMPW's LSB. The OEMPW field is reset to all zeros at power on. Any values written to these locations will be readable by the host regardless of the locked/unlocked status of the device. If OEMPWSET is set to zero (00000000$_h$), the MIC3003 will remain unlocked regardless of the contents of the OEMPW field. This is the factory default security setting.

Note that a valid OEM password allows access to the OEM and user areas of the chip, i.e., the entire memory map.

OEM Mode and User Mode

When the OEM password is unlocked (either by matching the set password or if the password is all zeros), the MIC3003 is in OEM Mode. If the part is locked, the part is in User Mode.

Detailed Register Descriptions

Note: Serial bus addresses shown assume that the base device address is AO_h.

Alarm Threshold Registers

Temperature High Alarm Threshold							
D[7] read/write	D[6] read/write	D[5] read/write	D[4] read/write	D[3] read/write	D[2] read/write	D[1] read/write	D[0] read/write
Default value for both bytes			$0000\ 0000_b = 00_h$ (0 °C)				
Serial address			$A2_h$				
Byte addresses			MS byte (TMAXh): $00 = 00_h$				
			LS byte (TMAXl): $01 = 01_h$				
Each LS bit of TMAXh represents one degree Celsius. TMAXl is not used, since all limit comparisons for temperature use eight-bit values. The eight bits of the high alarm threshold value (TMAXh) are compared to the temperature result (TEMPh). ALARM0 bit 7 is set if Result > Threshold.							

Temperature Low Alarm Threshold							
D[7] read/write	D[6] read/write	D[5] read/write	D[4] read/write	D[3] read/write	D[2] read/write	D[1] read/write	D[0] read/write
Default value for both bytes			$0000\ 0000_b = 00_h$ (0 °C)				
Serial address			$A2_h$				
Byte addresses			MS byte (TMINh): $02 = 02_h$				
			LS byte (TMINl): $03 = 03_h$				
Each LS bit of TMINh represents one degree Celsius. TMINl is not used, since all limit comparisons for temperature use eight-bit values.. The eight MS bits of the low alarm threshold value (TMINh) are compared to the temperature result (TEMPh). ALARM0 bit 6 is set if Result < Threshold.							

Voltage High Alarm Threshold							
D[7] read/write	D[6] read/write	D[5] read/write	D[4] read/write	D[3] read/write	D[2] read/write	D[1] read/write	D[0] read/write
Default value for both bytes			$0000\ 0000_b = 00_h$ (0 V)				
Serial address			$A2_h$				
Byte addresses			MS byte (VMAXh): $08 = 08_h$				
			LS byte (VMAXl): $09 = 09_h$				
Each LS bit of VMAXh represents 25.6 mV and each LS bit of VMAXl represents 0.1 mV. The sixteen-bit threshold value (VMAXh:VMAXl) is compared to the sixteen bits value of the voltage result (VINh:VINl). ALARM0 bit 5 is set if Result > Threshold.							

Voltage Low Alarm Threshold							
D[7] read/write	D[6] read/write	D[5] read/write	D[4] read/write	D[3] read/write	D[2] read/write	D[1] read/write	D[0] read/write
Default value for both bytes			$0000\ 0000_b = 00_h$ (0 V)				
Serial address			$A2_h$				
Byte addresses			MS byte (VMINh): 10 = $0A_h$				
			LS byte (VMINl): 11 = $0B_h$				
Each LS bit of VMINh represents 25.6 mV and each LS bit of VMINl represents 0.1 mV. The sixteen-bit threshold value (VMINh:VMINl) is compared to the sixteen-bit value of the voltage result (VINh:VINl). ALARM0 bit 4 is set if Result < Threshold.							

Bias Current High Alarm Threshold							
D[7] read/write	D[6] read/write	D[5] read/write	D[4] read/write	D[3] read/write	D[2] read/write	D[1] read/write	D[0] read/write
Default value for both bytes			$0000\ 0000_b = 00_h$ (0 mA)				
Serial address			$A2_h$				
Byte addresses			MS byte (IMAXh): 16 = 10_h				
			LS byte (IMAXl): 17 = 11_h				
Each LS bit of IMAXh represents 512 μA and each LS bit of IMAXl represents 2 μA. The sixteen-bit threshold value (IMAXh:IMAXl) is compared to the sixteen-bit value of the bias current result (ILDh:ILDl). ALARM0 bit 3 is set if Result > Threshold.							

Bias Current Low Alarm Threshold							
D[7] read/write	D[6] read/write	D[5] read/write	D[4] read/write	D[3] read/write	D[2] read/write	D[1] read/write	D[0] read/write
Default value for both bytes			$0000\ 0000_b = 00_h$ (0 mA)				
Serial address			$A2_h$				
Byte addresses			MS byte (IMINh): 18 = 12_h				
			LS byte (IMINl): 19 = 13_h				
Each LS bit of IMINh represents 512 μA and each LS bit of IMINl represents 2 μA. The sixteen-bit threshold value (IMINh:IMINl) is compared to the sixteen-bit value of the bias current result (ILDh:ILDl). ALARM0 bit 2 is set if Result < Threshold.							

TX Optical Power High Alarm Threshold							
D[7] read/write	D[6] read/write	D[5] read/write	D[4] read/write	D[3] read/write	D[2] read/write	D[1] read/write	D[0] read/write
Default value for both bytes			$0000\ 0000_b = 00_h$ (0 mW)				
Serial address			$A2_h$				
Byte addresses			MS byte (TXMAXh): 24 = 18_h				
			LS byte (TXMAXl): 25 = 19_h				
Each LS bit of TXMAXh represents 25.6 μW, and each LS bit of TXMAXl represents 0.1 μW. The sixteen-bit threshold value (TXMAXh:TXMAXl) is compared to the sixteen-bit value of the TX power result (TXOPh:TXOPl). ALARM0 bit 1 is set if Result > Threshold.							

TX Optical Power Low Alarm Threshold							
D[7] read/write	D[6] read/write	D[5] read/write	D[4] read/write	D[3] read/write	D[2] read/write	D[1] read/write	D[0] read/write
Default value for both bytes			$0000\ 0000_b = 00_h$ (0 mW)				
Serial address			$A2_h$				
Byte addresses			MS byte (TXMINh): 26 = $1A_h$				
			LS byte (TXMINl): 27 = $1B_h$				
Each LS bit of TXMINh represents 25.6 μW, and each LS bit of TXMINl represents 0.1 μW. The sixteen-bit threshold value (TXMINh:TXMINl) is compared to the sixteen-bit value of the TX power reading (TXOPh:TXOPl). ALARM0 bit 0 is set if Result < Threshold.							

RX Optical Power High Alarm Threshold							
D[7] read/write	D[6] read/write	D[5] read/write	D[4] read/write	D[3] read/write	D[2] read/write	D[1] read/write	D[0] read/write
Default value for both bytes			$0000\ 0000_b = 00_h$ (0 mW)				
Serial address			$A2_h$				
Bytes addresses			MS byte (RXMAXh): 32 = 20_h				
			LS byte (RXMAXl): 33 = 21_h				
Each LS bit of RXMAXh represents 25.6 μW, and each LS bit of RXMAXl represents 0.1 μW. The sixteen-bit threshold value (RXMAXh:RXMAXl) is compared to the sixteen-bit value of the RX power result (RXOPh:RXOPl). ALARM1 bit 7 is set if Result > Threshold.							

RX Optical Power Low Alarm Threshold							
D[7] read/write	D[6] read/write	D[5] read/write	D[4] read/write	D[3] read/write	D[2] read/write	D[1] read/write	D[0] read/write
Default value for both bytes			$0000\ 0000_b = 00_h$ (0mW)				
Serial Address			$A2_h$				
Byte Address			MS byte (RXMINh): 34 = 22_h				
			LS byte (RXMINl): 35 = 23_h				
Each LSB of RXMINh represents 25.6 μW, and each LS bit of RXMINl represents 0.1 μW. The sixteen-bit threshold value (RXMINh:RXMINl) is compared to the sixteen-bit value of the RX power result (RXOPh:RXOPl). ALARM1 bit 6 is set if Result < Threshold.							

Warning Threshold Registers

Temperature High Warning Threshold							
D[7] read/write	D[6] read/write	D[5] read/write	D[4] read/write	D[3] read/write	D[2] read/write	D[1] read/write	D[0] read/write
Default value for both bytes			0000 0000$_b$ = 00$_h$ (0 °C)				
Serial address			A2$_h$				
Bytes addresses			MS byte (THIGHh): 04 = 04$_h$				
			LS byte (THIGHl): 05 = 05$_h$				
Each LS bit of THIGHh represents one degree Celsius. THIGHl is not used, since all limit comparisons for temperature use eight-bit values. The eight bits of the high warning threshold value (THIGHh) are compared to the temperature result (TEMPh). WARN0 bit 7 is set if Result > Threshold.							

Temperature Low Warning Threshold							
D[7] read/write	D[6] read/write	D[5] read/write	D[4] read/write	D[3] read/write	D[2] read/write	D[1] read/write	D[0] read/write
Default value for both bytes			0000 0000$_b$ = 00$_h$ (0 °C)				
Serial address			A2$_h$				
Bytes addresses			MS byte (TLOWh): 06 = 06$_h$				
			LS byte (TLOWl): 07 = 07$_h$				
Each LS bit of TLOWh represents one degree Celsius. TLOWl is not used, since all limit comparisons for temperature use eight-bit values. The eight bits of the high warning threshold value (TLOWh) are compared to the temperature result (TEMPh). WARN0 bit 6 is set if Result < Threshold.							

Voltage High Warning Threshold							
D[7] read/write	D[6] read/write	D[5] read/write	D[4] read/write	D[3] read/write	D[2] read/write	D[1] read/write	D[0] read/write
Default value for both bytes			0000 0000$_b$ = 00$_h$ (0 V)				
Serial address			A2$_h$				
Bytes addresses			MS byte (VHIGHh): 12 = 0C$_h$				
			LS byte (VHIGHl): 13 = 0D$_h$				
Each LS bit of VHIGHh represents 25.6 mV and each LS bit of VHIGHl represents 0.1 mV. The sixteen-bit threshold value (VHIGHh:VHIGHl) is compared to the sixteen bits value of the voltage result (VINh:VINl). WARN0 bit 5 is set if Result > Threshold.							

Voltage Low Warning Threshold							
D[7] read/write	D[6] read/write	D[5] read/write	D[4] read/write	D[3] read/write	D[2] read/write	D[1] read/write	D[0] read/write
Default value for both bytes			$0000\ 0000_b = 00_h$ (0 V)				
Serial address			$A2_h$				
Byte addresses			MS byte (VLOWh): 14 = $0E_h$				
			LS byte (VLOWl): 15 = $0F_h$				
Each LS bit of VLOWh represents 25.6 mV and each LS bit of VLOWl represents 0.1 mV. The sixteen-bit threshold value (VLOWh:VLOWl) is compared to the sixteen-bit value of the voltage result (VINh:VINl). WARN0 bit 4 is set if Result < Threshold.							

Bias Current High Warning Threshold							
D[7] read/write	D[6] read/write	D[5] read/write	D[4] read/write	D[3] read/write	D[2] read/write	D[1] read/write	D[0] read/write
Default value for both bytes			$0000\ 0000_b = 00_h$ (0 mA)				
Serial address			$A2_h$				
Bytes addresses			MS byte (IHIGHh): 20 = 14_h				
			LS byte (IHIGHl): 21 = 15_h				
Each LS bit of IHIGHh represents 512 μA and each LS bit of IHIGHl represents 2 μA. The sixteen-bit threshold value (IHIGHh:IHIGHl) is compared to the sixteen-bit value of the bias current result (ILDh:ILDl). WARN0 bit 3 is set if Result > Threshold.							

Bias Current Low Warning Threshold							
D[7] read/write	D[6] read/write	D[5] read/write	D[4] read/write	D[3] read/write	D[2] read/write	D[1] read/write	D[0] read/write
Default value for both bytes			$0000\ 0000_b = 00_h$ (0 mA)				
Serial address			$A2_h$				
Bytes addresses			MS byte (ILOWh): 22 = 16_h				
			LS byte (ILOWl): 23 = 17_h				
Each LS bit of ILOWh represents 512 μA and each LS bit of ILOWl represents 2 μA. The sixteen-bit threshold value (ILOWh:ILOWl) is compared to the sixteen-bit value of the bias current result (ILDh:ILDl). WARN0 bit 2 is set if Result < Threshold.							

TX Optical Power High Warning							
D[7] read/write	D[6] read/write	D[5] read/write	D[4] read/write	D[3] read/write	D[2] read/write	D[1] read/write	D[0] read/write
Default value for both bytes			$0000\ 0000_b = 00_h$ (0 mW)				
Serial address			$A2_h$				
Bytes addresses			MS byte (TXHIGHh): 28 = $1C_h$				
			LS byte (TXHIGHl): 29 = $1D_h$				
Each LS bit of TXHIGHh represents 25.6 μW, and each LS bit of TXHIGHl represents 0.1 μW. The sixteen-bit threshold value (TXHIGHh:TXHIGHl) is compared to the sixteen-bit value of the TX power result (TXOPh:TXOPl).WARN0 bit 1 is set if Result > Threshold.							

TX Optical Power Low Warning							
D[7] read/write	D[6] read/write	D[5] read/write	D[4] read/write	D[3] read/write	D[2] read/write	D[1] read/write	D[0] read/write
Default value for both bytes			$0000\ 0000_b$ = 00_h (0 mW)				
Serial address			$A2_h$				
Byte addresses			MS byte (TXLOWh): 30 = $1E_h$				
			LS byte (TXLOWl): 31 = $1F_h$				
Each LS bit of TXLOWh represents 25.6 μW, and each LS bit of TXLOWl represents 0.1 μW. The sixteen-bit threshold value (TXLOWh:TXLOWl) is compared to the sixteen-bit value of the TX power reading (TXOPh:TXOPl). ALARM0 bit 0 is set if Result < Threshold.							

RX Optical Power High Warning Threshold							
D[7] read/write	D[6] read/write	D[5] read/write	D[4] read/write	D[3] read/write	D[2] read/write	D[1] read/write	D[0] read/write
Default value for both bytes			$0000\ 0000_b$ = 00_h (0 mW)				
Serial address			$A2_h$				
Byte addresses			MS byte (RXHIGHh): 36 = 24_h				
			LS byte (RXHIGHl): 37 = 25_h				
Each LS bit of RXHIGHh represents 25.6 μW, and each LS bit of RXHIGHl represents 0.1 μW. The sixteen-bit threshold value (RXHIGHh:RXHIGHl) is compared to the sixteen-bit value of the RX power result (RXOPh:RXOPl). WARN1 bit 7 is set if Result > Threshold.							

RX Optical Power Low Warning Threshold							
D[7] read/write	D[6] read/write	D[5] read/write	D[4] read/write	D[3] read/write	D[2] read/write	D[1] read/write	D[0] read/write
Default value for both bytes			$0000\ 0000_b$ = 00_h (0 mW)				
Serial address			$A2_h$				
Byte addresses			MS byte (RXLOWh): 38 = 26_h				
			LS byte (RXLOWl): 39 = 27_h				
Each LSB of RXLOWh represents 25.6 μW, and each LS bit of RXLOWl represents 0.1 μW. The sixteen-bit threshold value (RXLOWh:RXLOWl) is compared to the sixteen-bit value of the RX power result (RXOPh:RXOPl). WARN1 bit 6 is set if Result < Threshold.							

Checksum (CHKSUM) Checksum of bytes 0 - 94 at serial address A2h							
D[7] read/write	D[6] read/write	D[5] read/write	D[4] read/write	D[3] read/write	D[2] read/write	D[1] read/write	D[0] read/write
Default value			$0000\ 0000_b$ = 00_h				
Serial address			$A2_h$				
Byte address			95 = $5F_h$				
This register is provided for compliance with SFF-8472. It is implemented as general-purpose non-volatile memory. Read/write access is possible whenever a valid OEM password has been entered. CHKSUM is read-only in User Mode.							

ADC Result Registers

Temperature Result							
D[7] read-only	D[6] read-only	D[5] read-only	D[4] read-only	D[3] read-only	D[2] read-only	D[1] read-only	D[0] read-only
Serial address			$A2_h$				
Byte addresses			MS byte (TEMPh): 96 = 60_h				
			LS byte (TEMPl): 97 = 61_h				
Each LS bit of TEMPh represents one degree Celsius. The TEMPh register is to be used in conjunction with the most significant bit of TEMPl to yield an eight-bit or nine-bit signed (two's complement) temperature value. If OEMCFG6 bit 1 is set to zero, temperature is read to 1 °C resolution in TEMPh only, and TEMPl is zero. If OEMCFG6 bit 1 is set to one, then temperature is read to 0.5 °C resolution as a nine-bit value consisting of TEMPh and the MS bit of TEMPl. The lower seven bits of TEMPl are zero. TEMPh will contain measured temperature data after the completion of one conversion.							

Voltage							
D[7] read-only	D[6] read-only	D[5] read-only	D[4] read-only	D[3] read-only	D[2] read-only	D[1] read-only	D[0] read-only
Default Value			$0000\ 0000_b$ = 00_h (0V)[2]				
Serial address			$A2_h$				
Byte addresses			MS byte (VINh): 98 = 62_h				
			LS byte (VINl): 99 = 63_h				
Each LSB of VINh represents 25.6 mV, and each LS bit of VINl represents 0.1 mV. VINh is used in conjunction with VINl to yield an unsigned sixteen-bit value. In external calibration mode, the host should process the results using the appropriate slope and offset coefficients. VINh contains the eight-bit ADC result and VINl is zero. In internal calibration mode, the MIC3003's ALU applies the coefficients stored in (VSLPh:VSLPl) and (VOFFh:VOFFl). The VIN registers will contain valid data after one ADC conversion cycle.							

Notes:

1. TEMPh will contain measured temperature data after the completion of one conversion.
2. VINh will contain measured data after one A/D conversion cycle.

Laser Diode Bias Current							
D[7] read-only	D[6] read-only	D[5] read-only	D[4] read-only	D[3] read-only	D[2] read-only	D[1] read-only	D[0] read-only
Default Value			$0000\ 0000_b$ = 00_h (0mA)[3]				
Serial address			$A2_h$				
Byte addresses			MS byte (ILDh):100 = 64_h				
			LS byte (ILDl):101 = 65_h				
Each LSB of ILDh represents 512 μA, and each LS bit of ILDl represents 2 μA. ILDh is used in conjunction with ILDl to yield an unsigned sixteen-bit value. In external calibration mode, the host should process the results using the appropriate slope and offset coefficients. ILDh contains the eight-bit ADC result and ILDl is zero. In internal calibration mode, the MIC3003's ALU applies the coefficients stored in (ISLPh:ISLPl) and (IOFFh:IOFFl). The ILD registers will contain valid data after one ADC conversion cycle.							

Transmitted Optical Power							
D[7] read-only	D[6] read-only	D[5] read-only	D[4] read-only	D[3] read-only	D[2] read-only	D[1] read-only	D[0] read-only
Default Value			$0000\ 0000_b = 00_h$ (0mW)[4]				
Serial address			$A2_h$				
Byte address			MS byte (TXOPh): 102 = 66_h				
			LS byte (TXOPI): 103 = 67_h				
Each LSB of TXOPh represents 25.6 μW, and each LS bit of TXOPI represents 0.1 μW. TXOPh is used in conjunction with TXOPI to yield an unsigned sixteen-bit value. In external calibration mode, the host should process the results using the appropriate slope and offset coefficients. TXOPh contains the eight-bit ADC result and TXOPI is zero. In internal calibration mode, the MIC3003's ALU applies the coefficients stored in (TXSLPh:TXSLPI) and (TXOFFh:TXOFFI). The TXOP registers will contain valid data after one ADC conversion cycle.							

Notes:

3. ILDh will contain measured data after one A/D conversion cycle.
4. TXOPh will contain measured data after one A/D conversion cycle.

Received Optical Power							
D[7] read-only	D[6] read-only	D[5] read-only	D[4] read-only	D[3] read-only	D[2] read-only	D[1] read-only	D[0] read-only
Default Value			$0000\ 0000_b = 00_h$ (0mW)[6]				
Serial address			$A2_h$				
Byte address			MS byte (RXOPh): 104 = 68_h				
			LS byte (RXOPI): 105 = 69_h				
Each LSB of RXOPh represents 25.6 μW, and each LS bit of RXOPI represents 0.1 μW.RXOPh is used in conjunction with RXOPI to yield an unsigned sixteen-bit value. In external calibration mode, the host should process the results using the appropriate slope and offset coefficients. RXOPh contains the twelve-bit ADC result and the lower four bits of RXOPI are zero. In internal calibration mode, the MIC3003's ALU applies the coefficients stored in (RXSLP[0-7]h:RXSLP[0-7]l) and (RXOFF[0-7]h:RXOFFI[0-7]). The RXOP registers will contain valid data after one ADC conversion cycle.							

Control and Status (CNTRL)							
D[7] TXDIS read-only	D[6] STXDIS read/write	D[5] RS1S read-only	D[4] RS0S read-only	D[3] SRS0 read/write	D[2] TXFLT read-only	D[1] LOS read-only	D[0] POR read-only
Default value			$0000\ 0000_b = 00_h$				
Serial address			$A2_h$				
Byte address			110 = $6E_h$				

Bit(s)		Function	Operation
D[7]	TXDIS	Reflects the state of the TXDISABLE pin	1 = disabled, 0 = enabled
D[6]	STXDIS	Soft transmit disable STXDIS is Ored with TXDIS to control the laser which will be turned off if one of these two signals is set to 1	1 = disabled 0 = enabled
D[5]	RS1S	Reflects the state of RS1 (pin 10)	1 = RS1 is high (>4.25 Gbps); 0 = RS1 is low (≤4.25 Gbps)
D[4]	RS0S	Reflects the state of RS0 (pin 4)	1 = RS0 is high (>4.25 Gbps) 0 = RS0 is low (≤4.25 Gbps)
D[3]	SRS0	Soft rate select (sets the state of the RS0 pin)	1 = Set RS0 high 0 = Set RS0 low
D[2]	TXFLT	Reflects the state of the TXFAULT pin	1 = TXFAULT is high (fault) 0 = TXFAULT low (no fault)
D[1]	LOS	Loss Of Signal. Reflects the state of the RXLOS pin	1 = RXLOS is high (loss of signal) 0 = RXLOS is low (no loss of signal)
D[0]	POR	MIC3003 power-on status	0 = POR complete, all analog data results have been converted at least once 1 = POR and first ADC sample cycle in progress

Application Select Control Mode (ASCM)		
D[7-6] Control bits read/write	D[5-0] Table select read/write	
Default value	0000 0000$_b$ = 00$_h$	
Serial address	A2$_h$	
Byte address	111 = 6F$_h$	
Bit(s)	**Function**	**Operation**
D[7-6]	Application Select Control Bits	
D[5-0]	Table Select	

Bit 7	Bit 6	
0	0	Rate Select/Extended Rate Select Emulation Mode: • RS0 (pin 4) **controls** RRSOUT(pin 23) • RS1 (pin 10) **does not control** TRSOUT (pin24) • Byte 110 bit 3 **controls** RRSOUT (pin 23) • Byte 118 bit 3 **controls** TRSOUT (pin 24)
0	1	Hardware Application Select Mode: • RS0 (pin 4) **controls** RRSOUT(pin 23) • RS1 (pin 10) **controls** TRSOUT (pin 24) • Byte 110 bit 3 **does not control** RRSOUT (pin 23) • Byte 118 bit 3 **does not control** TRSOUT (pin 24)
1	X	Software Mode: • RS0 (pin4) **does not control** RRSOUT(pin 23) • RS1 (pin10) **does not control** TRSOUT (pin 24) • Byte 110 bit 3 **does not control** RRSOUT (pin 23) • Byte 118 bit 3 **does not control** TRSOUT (pin 24)

Alarm Flags

Alarm Status Register 0 (ALARM0)							
D[7] A7 read-only	D[6] A6 read-only	D[5] A5 read-only	D[4] A4 read-only	D[3] A3 read-only	D[2] A2 read-only	D[1] A1 read-only	D[0] A0 read-only
Default value	$0000\ 0000_b = 00_h$ (no events pending)						
Serial address	$A2_h$						
Byte address	$112 = 70_h$						
The power-up default value is 00_h. Following the first complete A/D conversion cycle, however, any of the bits may be set depending upon the results.							

Bit(s)		Function	Operation
D[7]	A7	High temperature alarm, TEMP > TMAX	1 = condition exists, 0 = normal/OK
D[6]	A6	Low temperature alarm, TEMP < TMIN	1 = condition exists, 0 = normal/OK
D[5]	A5	High voltage alarm, VIN > VMAX	1 = condition exists, 0 = normal/OK
D[4]	A4	Low voltage alarm, VIN < VMIN	1 = condition exists, 0 = normal/OK
D[3]	A3	High laser diode bias alarm, IBIAS > IMAX	1 = condition exists, 0 = normal/OK
D[2]	A2	Low laser diode bias alarm, IBIAS < IMIN	1 = condition exists, 0 = normal/OK
D[1]	A1	High transmit optical power alarm, TXOP > TXMAX	1 = condition exists, 0 = normal/OK
D[0]	A0	Low transmit optical power alarm, TXOP < TXMIN	1 = condition exists, 0 = normal/OK

Alarm Status Register 1 (ALARM1)							
D[7] A15 read-only	D[6] A14 read-only	D[5] reserved	D[4] reserved	D[3] reserved	D[2] reserved	D[1] reserved	D[0] reserved
Default value	$0000\ 0000_b = 00_h$ (no events pending)						
Serial address	$A2_h$						
Byte address	$113 = 71_h$						
The power-up default value is 00_h. Following the first complete A/D conversion cycle, however, either of the bits may be set depending upon the results.							

Bit(s)		Function	Operation
D[7]	A15	High received power (overload) alarm, RXOP > RXMAX	1 = condition exists, 0 = normal/OK
D[6]	A14	Low received power (LOS) alarm, RXOP < RXMIN	1 = condition exists, 0 = normal/OK
D[5:0]		Reserved	Reserved, returns zero on reads

Warning Flags

Warning Status Register 0 (WARN0)							
D[7] W7 read-only	D[6] W6 read-only	D[5] W5 read-only	D[4] W4 read-only	D[3] W3 read-only	D[2] W2 read-only	D[1] W1 read-only	D[0] W0 read-only
Default value	0000 0000$_b$ = 00$_h$ (no events pending)						
Serial address	A2$_h$						
Byte address	116 = 74$_h$						
The power-up default value is 00$_h$. Following the first complete A/D conversion cycle, however, any of the bits may be set depending upon the results.							

Bit(s)		Function	Operation
D[7]	W7	High temperature warning, TEMP > THIGH	1 = condition exists, 0 = normal/OK
D[6]	W6	Low temperature warning, TEMP < TLOW	1 = condition exists, 0 = normal/OK
D[5]	W5	High voltage warning, VIN > VHIGH	1 = condition exists, 0 = normal/OK
D[4]	W4	Low voltage warning, VIN < VLOW	1 = condition exists, 0 = normal/OK
D[3]	W3	High laser diode bias warning, IBIAS > IHIGH	1 = condition exists, 0 = normal/OK
D[2]	W2	Low laser diode bias warning, IBIAS < ILOW	1 = condition exists, 0 = normal/OK
D[1]	W1	High transmit optical power warning, TXOP > TXHIGH	1 = condition exists, 0 = normal/OK
D[0]	W0	Low transmit optical power warning, TXOP < TXLOW	1 = condition exists, 0 = normal/OK

Warning Status Register 1 (WARN1)							
D[7] W15 read-only	D[6] W14 read-only	D[5] read-only	D[4] read-only	D[3] read-only	D[2] read-only	D[1] read-only	D[0] read-only
Default Value	0000 0000$_b$ = 00$_h$ (no events pending)						
Serial Address	A2$_h$						
Byte Address	117 = 75$_h$						
The power-up default value is 00$_h$. Following the first complete A/D conversion cycle, however, either of the bits may be set depending upon the results.							

Bit(s)		Function	Operation
D[7]	W15	Received power high warning, RXOP > RXHIGH	1 = condition exists, 0 = normal/OK
D[6]	W14	Received power low warning, RXOP < RXMIN	1 = condition exists, 0 = normal/OK
D[5:0]		Reserved	Reserved, returns zero on reads

Extended Control and Status (ECNTRL)							
D[7] reserved	D[6] reserved	D[5] reserved	D[4] reserved	D[3] SRS1 read/write	D[2] reserved	D[1] PLOS read/write	D[0] PLS read/write
Default Value				$0000\ 0000_b = 00_h$			
Serial Address				$A2_h$			
Byte Address				$118 = 76_h$			

Bit(s)		Function	Operation
D[7-4]		Reserved	Reserved—always read as zeros
D[3]	SRS1	Soft rate select (RS1)	Assert the TRSOUT pin in Rate Select/Extended Rate Select Emulation Mode
D[2]		Reserved	Reserved—always reads as zero
D[1]	PLOS	Power Level Operation State	These two bits are read/write but change no functionality of the MIC3003
D[0]	PLS	Power Level Select	

OEM Password Entry (OEMPW)							
D[7] read/write	D[6] read/write	D[5] read/write	D[4] read/write	D[3] read/write	D[2] read/write	D[1] read/write	D[0] read/write
Default value for all four bytes			$0000\ 0000_b = 00_h$ (reset to zero at power-on)				
Serial address			$A2_h$				
Byte addresses			If OEMCFG5-2 = 0: 120 – 123 = $78_h – 7B_h$ If OEMCFG5-2 = 1: 123 – 126 = $7B_h – 7E_h$				

This four-byte field is for entry of the password required to access the OEM area of the MIC3003's memory and registers. A valid OEM password will also permit access to the user areas of memory. This field is compared to the four-byte OEMPWSET field at serial address A6h, bytes 12 - 15 ($0C_h – 0F_h$). If the two fields match, access is allowed to the OEM areas of the MIC3003 non-volatile memory at serial addresses $A4_h$ and $A6_h$. The OEM password is set by writing the new value into OEMPWSET. The password comparison is performed following the write to the highest address byte of OEMPW, address $7B_h$ if OEMCFG5 bit 2 is low, or $7E_h$ if OEMCFG5 bit 2 is high. This byte must be written last.

A four-byte burst-write sequence to OEMPW may be used as this will result in the highest address byte being written last.

OEMPW is reset to zero at power on. Any values written to OEMPW will be readable by the host regardless of the locked/unlocked status of the device. If OEMPWSET is set to zero (00000000_h), the MIC3003 will remain unlocked regardless of the contents of the OEMPW field. This is the factory default security setting.

Byte	Weight
3	OEM Password Entry, Most Significant Byte (Address = 7Bh resp. 7Eh)
2	OEM Password Entry, 2nd Most Significant Byte (Address = 7Ah resp. 7Dh)
1	OEM Password Entry, 2nd Least Significant Byte (Address = 79h resp. 7Ch)
0	OEM Password Entry, Least Significant Byte (Address = 78h resp. 7Bh)

Power-On Hours (POHh and POHl)							
D[7] read/write	D[6] read/write	D[5] read/write	D[4] read/write	D[3] read/write	D[2] read/write	D[1] read/write	D[0] read/write
Default value of both bytes when the MIC3003 is shipped from the factory			0000 0000$_b$ = 00$_h$				
Serial address			A6$_h$				
Bytes addresses			MS byte (POHh): 251 = FB$_h$				
			LS byte (POHl): 252 = FC$_h$				
The lower seven bits of POHh contain the most significant bits of the 15-bit power-on hours measurement. The value in POHh should be combined with the eight-bit power-on hours low byte, POHl, to yield the complete result.							
The most significant bit of POHh, POHFLT, is an error flag. If POHFLT is set, the power-on hour meter data has been corrupted and should be ignored.							
It is recommended that a two-byte sequential (block) SMBus read operation be performed on POHh and POHl to insure coherency between the two registers.							
This register is non-volatile and will be maintained through power and reset cycles, including unanticipated power failures.							

POHh Bit(s)	Function	Operation
D[7]	Power-on hours fault flag, POHFLT	1 = fault: the power-on hours value is corrupted and cannot be relied upon
		0 = no fault: the power-on hours value is correct
D[6:0]	Power-on hours, most significant seven bits	

Data Ready Flags (DATARDY)							
D[7] TRDY read/write	D[6] VRDY read/write	D[5] IRDY read/write	D[4] TXRDY read/write	D[3] RXDY read/write	D[2] reserved	D[1] reserved	D[0] reserved
Default value			0000 0000$_b$ = 00$_h$				
Serial address			A6$_h$				
Byte address			253 = FD$_h$				
When the A/D conversion for a given parameter is completed and the results available to the host, the appropriate data ready flag will be set. The flag will be cleared when the host reads the corresponding result register.							

Bit(s)		Function	Operation
D[7]	TRDY	Temperature data ready flag	0 = Result register contains old data; 1 = new data ready
D[6]	VRDY	Voltage data ready flag	0 = Result register contains old data; 1 = new data ready
D[5]	IRDY	Bias current data ready flag	0 = Result register contains old data; 1 = new data ready
D[4]	TXRDY	Transmit power data ready flag	0 = Result register contains old data; 1 = new data ready
D[3]	RXRDY	Receive power data ready flag	0 = Result register contains old data; 1 = new data ready
D[2:0]		Reserved	Reserved

User Control Register (USRCTL)							
D[7] Reserved read/write	D[6] PORM read/write	D[5] PORS read/write	D[4] IE read/write	D[3] APCSEL read/write	D[2] APCSEL read/write	D[1] APCSEL read/write	D[0] MODSEL read/write
Default value			$0010\ 0000_b = 20_h$				
Serial address			$A2_h$				
Byte address			$254 = FE_h$				
This register provides for control of the nominal APC setpoint and management of interrupts by the end-user. APCSEL[1:0] select which of the APC setpoint registers, APCSET0, APCSET1, or APCSET2 are used as the nominal automatic power control setpoint. Similarly, MODSEL[1:0] select which of MODSET0, MODSET1, or MODSET2 are used to select the modulation level of the laser.							
IE must be set for any host interrupts to occur via the /INT pin. If IE is set while /INT is asserted, /INT will be de-asserted immediately.							
PORS is always set high by any power-on reset event. If PORM is high, the power-on event will also generate a host interrupt. PORS will be cleared to zero and the interrupt output de-asserted when USRCTL is read by the host.							
If PORM is set following the setting of PORS, PORS will remain set, and /INT will be asserted immediately. /INT will not be de-asserted until USRCTL is read by the host.							
PORM, IE, APCSEL, and MODSEL are non-volatile and will be maintained through power and reset cycles.							

Bit(s)		Function	Operation
D[7]		Reserved	Always write as zero; reads undefined.
D[6]	PORM	Power-on interrupt mask	1 = POR interrupts mask enabled 0 = POR interrupts mask disabled
D[5]	PORS	Power-on interrupt flag	1 = POR interrupt occurred 0 = No POR interrupt
D[4]	IE	Global interrupt enable	1 = Host interrupts are enabled 0 = Host interrupts are disabled
D[3:2]	APCSEL	Selects APC setpoint register	00 = APCSET0 01 = APCSET1 10 = APCSET2 11 = Reserved
D[1:0]	MODSEL	Selects Modulation setpoint register	00 = MODSET0 01 = MODSET1 10 = MODSET2 11 = Reserved

RESETOUT							
D[7] reserved read-only	D[6] reserved read-only	D[5] reserved read-only	D[4] reserved read-only	D[3] reserved read-only	D[2] RESETOUT read/write	D[1] RESETOUT read/write	D[0] RESETOUT read/write
Default Value			$0000\ 0000_b = 00_h$				
Serial address			$A2_h$				
Byte address			$255 = FF_h$				

Bit(s)		Function	Operation
D[7-3]		Reserved	Read-only; these bits always return 00000.
D[2:0]	RESETOUT	Controls the reset output at pin 2 (QGPO) when Reset mode is selected (OEMCFG3-7 set to 1)	By default, these three bits are 000, and the QGPO output is undriven.
			If RESET mode is selected in OEMCFG3:
			When the three bits are written to 111, QGPO's open-drain output will be driven low for 125 µs (typical), after which QGPO reenters the undriven state.
			The RESETOUT field is cleared from 111 to 000 22.5 ms (typical) after the de-assertion edge of QGPO. Other values of this delay may be selected in the TRSTCLR[2:0] field in OEMCFG2.
			If Reset mode in OEMCFG3 is not selected, these three bits have no function.

OEM Configuration Register 0 (OEMCFG0)							
D[7] RST write only	D[6] QGPOS read/write	D[5] DFLT read-only	D[4] OE reserved	D[3] MODREF reserved	D[2] VAUX[2] read/write	D[1] VAUX[1] read/write	D[0] VAUX[0] read/write
Default value			$0000\ 0000_b = 00_h$				
Serial address			$A6_h$				
Byte address			$00 = 00_h$				
A write to OEMCFG0 will result in any A/D conversion in progress being aborted and the result discarded. The A/D will begin a new conversion sequence once the write operation is complete.							
All bits in OEMCFG0 are non-volatile except DFLT and RST. A valid OEM password is required for access to this register.							

Bit(s)		Function	Operation
D[7]	RST		0 = no action; 1 = reset
			Issuing a software reset by setting RST high is equivalent to a full power cycle of the MIC3003.
D[6]	QGPOS	Determines the state of QGPO in GPO mode	If OEMCFG3 bit 7 (QGPOM) is low, this bit determines whether the QGPO output is high (undriven) or low (driven-open-drain).
			If QGPOM is high (Reset mode), this bit has no function
D[5]	DFLT	Diode fault flag.	1 = diode fault; 0 = OK.
D[4]	OE	Output enable for SHDN, V_{MOD}, and V_{BIAS}.	1 = enabled; 0 = hi-Z
D[3]	MODREF	Selects whether V_{MOD} is referenced to ground or V_{DD}.	1 = V_{DD}; 0 = GND
D[2:0]	VAUX[2:0]	Selects the voltage reported in VINh:VINl.	000 = V_{IN}
			001 = V_{DDA}
			010 = V_{BIAS}
			011 = V_{MOD}
			100 = APCDAC
			101 = MODDAC
			110 = FLTDAC

OEM Configuration Register 1 (OEMCFG1)							
D[7] INV read/write	D[6] GAIN read/write	D[5] BIASREF read/write	D[4] RFB[2] read/write	D[3] RFB[1] read/write	D[2] RFB[0] read/write	D[1] SRCE read/write	D[0] SPOL read/write
Default value			$0000\ 0010_b = 02_h$				
Serial address			$A6_h$				
Byte address			$1 = 01_h$				
A write to OEMCFG1 will result in any A/D conversion in progress being aborted and the result discarded. The A/D will begin a new conversion sequence once the write operation is complete. All bits in OEMCFG1 are non-volatile and will be maintained through power and reset cycles. A valid OEM password is required for access to this register.							

Bit(s)		Function	Operation
D[7]	INV	Inverts the APC op-amp inputs. When low, the BIAS DAC output is connected to the "+"input and FB is connected to the "−" input of the op amp. Set low to use the APC feedback loop.	0 = emitter follower (no inversion); 1 = common emitter (inverted); read/write; non-volatile.
D[6]	GAIN	Sets the feedback voltage range by changing the APCDAC output swing; $0-V_{REF}$ for optical feedback, $0-V_{REF}/4$ for electrical feedback.	$1 = V_{REF}/4$ full scale; $0 = V_{REF}$ full scale
D[5]	BIASREF	Selects whether FB and VMPD are referenced to ground or V_{DD} and selects feedback resistor termination voltage (V_{DDA} or GNDA).	$1 = V_{DD}$; 0 = GNDA If this bit is set to 0, bit 1 should be set to 1 If this bit is set to 1, bit 1 should be set to 0
D[4:2]	RFB[2:0]	Selects internal feedback resistance. Resistors will be terminated to V_{DDA} or GNDA according to BIASREF.	000 = ∞ 001 = 800 Ω 010 = 1.6kΩ 011 = 3.2kΩ 100 = 6.4kΩ 101 = 12.8kΩ 110 = 25.6kΩ 111 = 51.2kΩ
D[1]	SRCE	V_{BIAS} source or sink drive.	1 = source (NPN): bit 5 should be set to 0. 0 = sink (PNP): bit 5 should be set to 1.
D[0]	SPOL	Polarity of the shutdown output, SHDN, when active.	1 = SHDN is active-high 0 = SHDN is active-low

OEM Configuration Register 2 (OEMCFG2)							
D[7] SMBADR[3] read/write	D[6] SMBADR[2] read/write	D[5] SMBADR[1] read/write	D[4] SMBADR[0] read/write	D[3] read/write	D[2] TRSTCLR[2] read/write	D[1] TRSTCLR[1] read/write	D[0] TRSTCLR[0] read/write
Default value			$1010\ 0010_b = xx_h$ (device address $= 1010\ xxxx_b$) This value is the basis for using A0$_h$, A2$_h$, A4$_h$, and A6$_h$ as the names of the different device address spaces of the MIC3003.				
Serial address			A6$_h$				
Byte address			$2 = 02_h$				
Caution: Changes to SMBADR take effect immediately. Any accesses following a write to SMBADR must be to the newly programmed serial bus address. A valid OEM password is required for access to this register. This register is non-volatile and will be maintained through power and reset cycles.							

	Bit(s)	Function	Operation
D[7:4]	SMBADR[3:0]	Most significant four bits of the serial bus device address	Writes take effect immediately.
D[3:0]	TRSTCLR[2:0]	Set the delay between QGPO and the clearing of RESETOUT	These three bits set the delay between the de-assertion edge of the QGPO output in Reset mode and the subsequent clearing of the three RESETOUT bits in the RESETOUT Register: 000: Zero delay 001: 17.5 ms typical 010: 22.5 ms typical (default) 011: 27.0 ms typical 100: 45 ms typical Minimum and maximum values may be found by adding tolerances of -10% and +10% to the above values. If Reset mode is not selected, these bits have no function.

APC Setpoint 0, 1, and 2 (APCSET0, APCSET1, APCSET2) Automatic Power Control Setpoint							
D[7] read/write	D[6] read/write	D[5] read/write	D[4] read/write	D[3] read/write	D[2] read/write	D[1] read/write	D[0] read/write
Default value for all three bytes			$0000\ 0000_b = 00_h$				
Serial address			$A6_h$				
Byte addresses			APCSET0: 3 = 03_h				
			APCSET1: 4 = 04_h				
			APCSET2: 5 = 05_h				
When the APC is on, the eight-bit signed integer compensation value corresponding to the current temperature is taken from the BIASLUT look-up table, added to the selected APCSET (0, 1, or 2) register and loaded into the V_{BIAS} DAC. If DAC Calibration mode is selected in OEMCAL0, a write to any one of the three APCSETn registers will cause the V_{BIAS} DAC to be updated immediately. DAC Calibration mode disables the output of the BIASLUT lookup table, so the unmodified APCSETn register value propagates directly to the DAC. The eight-bit value presented to the V_{BIAS} DAC is always available for read back in the APCDAC register. A valid OEM password is required for access to these registers. These registers are non-volatile and will be maintained through power and reset cycles.							

Modulation Setpoint 0, 1, and 2 (MODSET0, MODSET1, and MODSET2) Nominal V_{MOD} Setpoint							
D[7] read/write	D[6] read/write	D[5] read/write	D[4] read/write	D[3] read/write	D[2] read/write	D[1] read/write	D[0] read/write
Default value for all three bytes			$0000\ 0000_b = 00_h$				
Serial address			$A6_h$				
Byte address			MODSET0: 6 = 06_h				
			MODSET1: 30 = $1E_h$				
			MODSET2: 31 = $1F_h$				
When the APC is on, the eight-bit signed integer compensation value corresponding to the current temperature is taken from the MODLUT look-up table, added to the selected MODSET (0, 1, or 2) register and loaded into the V_{MOD} DAC. If DAC Calibration mode is selected in OEMCAL0, a write to any one of the three MODSETn registers will cause the V_{MOD} DAC to be updated immediately. DAC Calibration mode disables the output of the MODLUT lookup table, so the unmodified MODSETn register value propagates directly to the DAC. The eight-bit value presented to the V_{MOD} DAC is always available for read back in the MODDAC register. A valid OEM password is required for access to these registers. These registers are non-volatile and will be maintained through power and reset cycles.							

I_{BIAS} Fault Threshold (IBFLT) Bias Current Fault Threshold							
D[7] read/write	D[6] read/write	D[5] read/write	D[4] read/write	D[3] read/write	D[2] read/write	D[1] read/write	D[0] read/write
Default value			$0000\ 0000_b = 00_h$				
Serial address			$A6_h$				
Byte address			$7 = 07_h$				
When the Fault Comparator is running, the eight-bit signed integer compensation value corresponding to the current temperature is taken from the IFTLUT look-up table, added to IBFLT, and used for comparison with the laser bias current value for fault generation. Faults are generated if the bias current value exceeds the compensated (LUT offset) IBFLT register contents.							
If DAC Calibration mode is selected in OEMCAL0, the output of the IFTLUT lookup table is disabled, so the unmodified IBFLT register value propagates directly to the Fault Comparator DAC.							
A valid OEM password is required for access to these registers. These registers are non-volatile and will be maintained through power and reset cycles.							

Transmit Power Fault Threshold (TXFLT)							
D[7] read/write	D[6] read/write	D[5] read/write	D[4] read/write	D[3] read/write	D[2] read/write	D[1] read/write	D[0] read/write
Default value			$0000\ 0000_b = 00_h$				
Serial address			$A6_h$				
Byte address			$8 = 08_h$				
When the Fault Comparator is running the TXFLT register value is used for comparison with the transmit power value for fault generation. Faults are generated if the transmit power exceeds the TXFLT register contents.							
A valid OEM password is required for access to this register. This register is non-volatile and will be maintained through power and reset cycles.							

Loss-Of-Signal Threshold (LOSFLT)							
D[7] read/write	D[6] read/write	D[5] read/write	D[4] read/write	D[3] read/write	D[2] read/write	D[1] read/write	D[0] read/write
Default value			$0000\ 0000_b = 00_h$				
Serial address			$A6_h$				
Byte address			$9 = 09_h$				
When the Fault Comparator is running, a fault is generated if the received power is lower than the LOSFLT value set in this register.							
A valid OEM password is required for access to this register. This register is non-volatile and will be maintained through power and reset cycles.							

Fault Suppression Timer (FLTTMR) Fault Suppression Interval in Increments of 0.5 ms							
D[7] read/write	D[6] read/write	D[5] read/write	D[4] read/write	D[3] read/write	D[2] read/write	D[1] read/write	D[0] read/write
Default value			$0000\ 0000_b = 00_h$				
Serial address			$A6_h$				
Byte address			$10 = 0A_h$				
Saturation faults are suppressed for a time, t_{FLTTMR}, following laser turn-on. This avoids nuisance tripping while the APC loop starts up. The length of this interval is (FLTTMR x 0.5 ms), typical. A value of zero will result in no fault suppression. A valid OEM password is required for access to this register. This register is non-volatile and will be maintained through power and reset cycles.							

Fault Mask (FLTMSK)							
D[7] OEMIM read/write	D[6] POHE read/write	D[5] reserved	D[4] reserved	D[3] SATMSK read/write	D[2] TXMSK read/write	D[1] IAMSK read/write	D[0] DFMSK read/write
Default value			$0000\ 0000_b = 00_h$				
Serial address			$A6_h$				
Byte address			$11 = 0B_h$				
A valid OEM password is required for access to this register. This register is non-volatile and will be maintained through power and reset cycles.							

Bit		Function	Operation
D[7]	OEMIM	OEM interrupt mask	1 = Interrupts are masked 0 = Interrupts are enabled This bit is similar to the IE (Global Interrupt Enable) bit in the User Control Register. The /INT output can only be asserted if IE is high and OEMIM is low.
D[6]	POHE	OEM Power-on Hour Meter enable	1 = Power-on Hour Meter enabled 0 = Power-on Hour Meter disabled
D[5:4]	D[5:4]	Reserved	Always write as zero; reads undefined
D[3]	SATMSK	APC saturation fault mask	1 = masked; 0 = enabled
D[2]	TXMSK	High TX optical power fault mask	1 = masked; 0 = enabled
D[1]	IAMSK	High bias current high fault mask	1 = masked; 0 = enabled
D[0]	DFMSK	Diode fault mask	1 = masked; 0 = enabled

OEM Password Setting (OEMPWSET)							
D[7] read/write	D[6] read/write	D[5] read/write	D[4] read/write	D[3] read/write	D[2] read/write	D[1] read/write	D[0] read/write
Default value for all bytes			0000 0000$_b$ = 00$_h$				
Serial address			A6$_h$				
Byte addresses			12 - 15 = 0C$_h$ - 0F$_h$				
This four-byte field is the password required for access to the OEM area of the MIC3003's memory and registers. This field is compared to the four-byte OEMPW field at serial address A2$_h$. If the two fields match, access is allowed to the OEM areas of the MIC3003 non-volatile memory at serial addresses A4$_h$ and A6$_h$. The OEM password may be set by writing the new value into OEMPWSET. The new password will not take effect until after a power-on reset occurs or a warm reset is performed using the RST bit in OEMCFG0. This allows the new password to be verified before it takes effect. These registers are non-volatile and will be maintained through power and reset cycles. A valid OEM password is required for access to this register.							

Byte	Weight
3	OEM Password, Most Significant Byte
2	OEM Password, 2nd Most Significant Byte
1	OEM Password, 2nd Least Significant Byte
0	OEM Password, Least Significant Byte

OEM Calibration 0 (OEMCAL0)							
D[7] reserved read-only	D[6] FLTDIS read/write	D[5] FSPIN read/write	D[4] WRINH read/write	D[3] APCCAL read/write	D[2] reserved read-only	D[1] reserved read-only	D[0] FRCOPS read/write
Default value			0000 0000$_b$ = 00$_h$				
Serial address			A6$_h$				
Byte address			16 = 10$_h$				
A valid OEM password is required for access to this register. This register is volatile and will not keep its value through power cycles.							

Bit		Function	Operation
D[7]		Reserved	Always write as zero; reads undefined.
D[6]	FLTDIS	Fault Comparator disable	0 = Fault Comparator faults enabled
			1 = Fault Comparator faults disabled
			When FLTDIS is high, the Fault Comparator does not run, and the following faults are not detected:
			• High IBIAS faults
			• High transmit power faults
			• Low receive power faults
			If a fault condition was detected prior to the assertion of FLTDIS, the fault flag will remain active until cleared, but cannot be set again until the Fault Comparator is allowed to run.
D[5]	FSPIN	Fault Comparator "spin-on-channel" mode select; do not enable ADC and FC spin-on-channel modes simultaneously.	0 = Normal Fault Comparator operation
			1 = Force the Fault Comparator to spin on one channel
			When the Fault Comparator spins on just one channel (selected via OEMCAL1[1:0]), the two channels not being examined will not, of course, respond to fault conditions.
D[4]	WRINH	Inhibit NVRAM write cycles.	0 = Normal NVRAM operation
			1 = Inhibit NVRAM writes
			When WRINH is high, writes to the MIC3003's internal memory do not occur. Registers that are non-volatile are written with the new value, but will not retain that value through a power cycle, since the NVRAM backing storage has not been modified.
D[3]	APCCAL	Selects APC DAC calibration mode - DACs may be controlled directly.	0 = Normal mode
			1 = DAC calibration mode.
			When DAC calibration mode is enabled, the temperature compensation lookup tables are disabled, so the DACs are presented with the values written into their corresponding registers.
D[2:1]		Reserved	Always write as zeros; reads undefined.
D[0]	FRCOPS	Forces outputs for board-level or system-level testing	0 = Normal operation
			1 = Force outputs for testing: The following outputs are driven to their active states:
			• TXFAULT (active polarity set in OEMCFG5)
			• /INT (only driven if the OEMIM bit is clear in FLTMSK)
			• RXLOS (active polarity and RXLOS selection are set in OEMCFG6)

OEM Calibration 1 (OEMCAL1)							
D[7] reserved read-only	D[6] ADSTP read/write	D[5] ADIDL read-only	D[4] ONESHOT read/write	D[3] ADSPIN read/write	D[2] SPIN[2] read/write	D[1] SPIN[1] read/write	D[0] SPIN[0] read/write
Default value			$0000\ 0000_b = 00_h$				
Serial address			$A6_h$				
Byte address			$17 = 11_h$				
A valid OEM password is required for access to this register. This register is volatile and will not keep its value through power cycles.							

Bit		Function	Operation
D[7]		Reserved	Always write as zero; reads undefined.
D[6]	ADSTP	Stop ADC: Halts the analog to digital converter	0 = ADC in normal operation 1 = ADC stopped When ADSTP is set, the five-channel ADC halts and no new ADC results are computed. Existing result registers are not changed. In addition, no new temperature compensation is applied to the APC loop and fault parameters, and no comparisons of the ADC results against the alarm and warning limits is performed.
D[5]	ADIDL	ADC idle flag	0 = ADC is busy 1 = ADC is idle ADIDL may be used in conjunction with ONESHOT to determine when the single ADC conversion is complete. After ONESHOT is set, the ADC runs until completion and then halts. Software may poll ADIDL to detect this completion before interrofating the result.
D[4]	ONESHOT	Triggers one-shot A/D conversion cycle	0 = Normal ADC operation 1 = ADC one-shot mode Setting ONESHOT high starts the ADC and causes it to stop after the next conversion is complete. After the conversion, the ADC remains stopped until ONESHOT is set low. Multiple single ADC conversions may be executed by repeatedly writing one to ONESHOT.
D[3]	ADSPIN	Selects ADC spin-on-channel mode; do not enable ADC and FC spin-on-channel modes simultaneously	0 = Normal ADC operation 1 = ADC spin-on-channel
D[2:0]	SPIN[2:0]	ADC and Fault Comparator (FC) channel select for spin-on-channel mode; do not enable ADC and FC spin-on-channel modes simultaneously	**ADC:** 000 = temperature 001 = voltage 010 = VILD 011 = VMPD 100 = VRX

LUT Index (LUTINDX)							
D[7] read-only	D[6] read-only	D[5] read-only	D[4] read-only	D[3] read-only	D[2] read-only	D[1] read-only	D[0] read-only
Default value (before the first set of 16 temperature measurements has been accumulated)			0000 0000b = 00h				
Serial address			A6h				
Byte address			18 = 12h				
The look-up table index is derived from the current temperature measurement as follows: INDEX = T_{AVG} / 2 where T_{AVG} is the current average temperature, averaged over a set of 16 samples. This register allows the current table index to be read by the host. The appropriate table base address must be added to LUTINDX to form a complete table index in physical memory. A valid OEM password is required for access to this register.							

OEM Configuration 3 (OEMCFG3)							
D[7] QGPOM read/write	D[6] TXFPOL read/write	D[5] GPOD read/write	D[4] GPOM read/write	D[3] GPOC read/write	D[2] TXFINM read/write	D[1] LOSDIS read/write	D[0] INTCAL read/write
Default value			0000 1000b = 08h				
Serial address			A6h				
Byte address			19 = 13h				
This register is non-volatile and will be maintained through power and reset cycles. A valid OEM password is required for access to this register.							

	Bit	Function	Operation
D[7]	QGPOM	QGPO pin function select	0: GPO (general purpose output)
			1: Reset output
D[6]	TXFPOL	TXFIN active polarity select; a fault is indicated when TXFIN = TXFPOL	0 = The TXFIN input is active-low
			1 = The TXFIN input is active-high
			This bit is ignored if TXFINM = 0
D[5]	GPOD	GPO output drive	0 = GPO output is open-drain
			1 = GPO output is push-pull
			This bit is ignored if GPOM = 0
D[4]	GPOM	GPO/RRSOUT pin mode select	0 = RRSOUT output
			1 = GPO output
D[3]	GPOC	GPO output control	0 = Set the GPO pin low
			1 = Set the GPO pin high
			This bit is ignored if GPOM = 0
D[2]	TXFINM	SHDN/TXFIN pin mode select	0 = SHDN output
			1 = TXFIN input
D[1]	LOSDIS	RXLOS comparator and output disable	0 = RXLOS fault enabled, and the RXLOS output is enabled for normal operation
			1 = RXLOS fault disabled; also, the RXLOS output is disabled, and will remain low
D[0]	INTCAL	Calibration mode select	0 = External calibration mode
			1 = Internal calibration(the MIC3003's ALU applies slope and offset coefficients to the ALU results as necessary)

BIAS DAC Setting (APCDAC) Current VBIAS Setting							
D[7] read-only	D[6] read-only	D[5] read-only	D[4] read-only	D[3] read-only	D[2] read-only	D[1] read-only	D[0] read-only
Default value			$0000\ 0000_b = 00_h$				
Serial address			$A6_h$				
Byte address			$20 = 14_h$				
This register reflects (reads back) the value being sent to the BIAS DAC (APCSET0, APCSET1, or APCSET2 whichever is selected, with temperature compensation applied). A valid OEM password is required for access to this register.							

Modulation DAC Setting (MODDAC) Current VMOD Setting							
D[7] read-only	D[6] read-only	D[5] read-only	D[4] read-only	D[3] read-only	D[2] read-only	D[1] read-only	D[0] read-only
Default value			$0000\ 0000_b = 00_h$				
Serial address			$A6_h$				
Byte address			$21 = 15_h$				
This register reflects (reads back) the value set in the being sent to the modulation DAC (MODSET0, MODSET1, or MODSET2 whichever is selected, with temperature compensation applied).							
A valid OEM password is required for access to this register.							

OEM Readback Register (OEMRD)							
D[7] Reserved read-only	D[6] KILL_LASER read-only	D[5] RXP_FLT Read-only	D[4] INT read-only	D[3] APCSAT read-only	D[2] IBFLT read-only	D[1] TXFLT read-only	D[0] RRSOUT read-only
Default value			$0000\ 0000_b = 00_h$				
Serial address			$A6_h$				
Byte address			$22 = 16_h$				
This register reflects (reads back) the status of the bits corresponding to the parameters defined below.							
A valid OEM password is required for access to this register.							

Bit		Function	Operation
D[7]		Reserved	Always reads as a zero
D[6]	KILL_LASER	State of the internal laser disable signal	0: The MIC3003 is disabling the laser 1: The laser is enabled to operate
D[5]	RXP_FLT	Registered Fault Comparator detection of a receive power fault	0: No Fault Comparator receive power fault 1: The Fault Comparator has detected a receive power fault
D[4]	INT	Mirrors state of /INT but active-high	1 = The interrupt is asserted 0 = No pending interrupt.
D[3]	APCSAT	Registered APC saturation fault	1 = APC saturation fault detected 0 = Normal operation.
D[2]	IBFLT	Registered Fault Comparator detection of an IBIAS fault	1 = IBIAS fault detected 0 = Normal operation
D[1]	TXFLT	Registered Fault Comparator detection of a transmit power fault	1 = Registered transmit power fault 0 = normal operation
D[0]	TRSOUT	State of the rate select output pin, TRSOUT	1 = high; 0 = low

Signal Detect Threshold (LOSFLTn)

D[7] read/write	D[6] read/write	D[5] read/write	D[4] read/write	D[3] read/write	D[2] read/write	D[1] read/write	D[0] read/write
Default value			0000 0000$_b$ = 00$_h$				
Serial address			A6$_h$				
Byte address			23 = 17$_h$				

This eight-bit register works in conjunction with the LOSFLT register to provide hysteresis and control the operation of the loss of signal comparator. The comparator's output, RXLOS, is asserted when the input on VRX falls below the level in LOSFLT. The output will then be de-asserted when the VRX signal rises above the level in LOSFLTn.

The input signal is subject to scaling by the RXPOT. If the LOS comparator is disabled, i.e., LOSDIS = 1, this register is ignored.

A valid OEM password is required for access to this register. This register is non-volatile and will be maintained through power and reset cycles.

RX EEPOT Tap Selection (RXPOT)

D[7] reserved read-only	D[6] reserved read-only	D[5] reserved read-only	D[4] RXPOT[4] read/write	D[3] RXPOT[3] read/write	D[2] RXPOT[2] read/write	D[1] RXPOT[1] read/write	D[0] RXPOT[0] read/write
Default value			0000 0000$_b$ = 00$_h$				
Serial address			A6$_h$				
Byte address			24 = 18$_h$				

This register is non-volatile and will be maintained through power and reset cycles. A valid OEM password is required for access to these registers.

Bit(s)	Function	Operation
D[7:5]	Reserved	Reserved. Always write as zero; reads return zeros.
D[4:0]	RXPOT tap selection:	Adjust gain of the receive power measurement: 00000 = No divider action 00001 = 31/32 00010 = 30/32 • • • 11110 = 2/32 11111 = 1/32

OEM Configuration 4 (OEMCFG4)

D[7] TXF_WRN read/write	D[6] TXF_ALM read/write	D[5] LAT_WRN read/write	D[4] LAT_ALM read/write	D[3] ISTART[3] read/write	D[2] ISTART[2] read/write	D[1] ISTART[1] read/write	D[0] ISTART[0] read/write
Default value			0000 0000$_b$ = 00$_h$				
Serial address			A6$_h$				
Byte address			25 = 19$_h$				

This register is non-volatile and will be maintained through power and reset cycles. A valid OEM password is required for access to these registers.

Bit(s)	Function		Operation
D[7]	TXF_WRN	Allows warnings to assert TXFAULT	0: Warnings do not assert TXFAULT
			1: Warnings assert TXFAULT
			The RXPWR low warning flag does not assert TXFAULT
D[6]	TXF_ALM	Allows alarms to assert TXFAULT	0: Alarms do not assert TXFAULT
			1: Alarms assert TXFAULT
			The RXPWR low alarm flag does not assert TXFAULT
D[5]	LAT_WRN	Warning latch	0: Warnings flags are latched. They are cleared by reading the register or toggling TXDISABLE.
			1: Warnings flags are not latched., i.e. they are set and reset with the warning condition. The flags are also cleared by reading the register or toggling TXDISABLE.
D[4]	LAT_ALM	Alarm latch	0: Alarms flags are latched. They are cleared by reading the register or toggling TXDISABLE.
			1: Alarms flags are not latched., i.e. they are set and reset with alarm condition. The flags are also cleared by reading the register or toggling TXDISABLE.
D[3:0]	ISTART[3:0]	I_{START} current level selection.	I_{START} current level selection: 0000 = No I_{START} current 0001 - 1111 = 0.375 mA x I_{START}[3:0] I_{START} is used to speed up the laser start-up after a fault occurs. The charging current of the compensation capacitor starts from I_{START} instead of ramping up from 0.

OEM Configuration 5 (OEMCFG5)							
D[7] read/write	D[6] read/write	D[5] read/write	D[4] read/write	D[3] read/write	D[2] read/write	D[1] read/write	D[0] read/write
Default value			0000 0000$_b$ = 00$_h$				
Serial address			A6$_h$				
Byte address			26 = 1A$_h$				
This register is non-volatile and will be maintained through power and reset cycles. A valid OEM password is required for access to these registers.							

Bit(s)	Function	Operation
D[7]	SHDN output enable / disable	0: SHDN is enabled. TXFAULT will trigger the SHDN output
		1: SHDN is disabled. TXFAULT has no effect on the SHDN output
		This applies when pin 12 is set as the SHDN output
D[6]	Lookup table temperature offset control	0: The temperature result used for the LUT access averaging algorithm does not have the offset coefficient applied
		1: The temperature result used for the LUT access averaging algorithm is offset by the signed 6-bit (.5 C resolution) offset coefficient.
D[5]	Temperature result register offset control	0: The temperature result register does not have the offset coefficient applied
		1: The temperature result register is offset by the signed 6-bit (.5 C resolution) offset coefficient.
D[4]	Polarity of TXFAULT	0: TXFAULT is active-high
		1: TXFAULT is active-low
D[3]	SMBus multipart support	0: Multipart mode off
		1: Multipart mode on
D[2]	OEM password location	0: A6h: 120-123 (78h-7Bh)
		1: A6h: 123-126 (7Bh-7Eh)
D[1]	SMBUS timeout enable / disable	0: SMBUS timeout enabled
		1: SMBUS timeout disabled

OEM Configuration 6 (OEMCFG6)							
D[7] read/write	D[6] read/write	D[5] read/write	D[4] read/write	D[3] read/write	D[2] read/write	D[1] read/write	D[0] read/write
Default value			$0000\ 0000_b = 00_h$				
Serial address			$A6_h$				
Byte address			$27 = 1B_h$				
This register is non-volatile and will be maintained through power and reset cycles. A valid OEM password is required for access to these registers.							

Bit(s)	Function	Operation
D[7]	A0h space access control	0: A0h space is used according to SFF-8472
		1: A0h space does not respond to any access request. SMBus transactions to A0h receive a NACK response.
D[6:5]	RXPWR Linearization Intervals	00: Temperature-based coefficient selection
		01: Programmable delimiters
		10: Hard coded delimiters
		11: Reserved
		Options 01 and 10 partition the receive power result range into eight regions, each of which has its own slope and offset coefficients. The delimiters identify the region boundaries within the 12-bit receive power range of results, and may be either hard-coded (10) or programmable (01).
		Use of hard-coded delimiters frees up extra A6 memory space for scratchpad use.
D[4]	TXDISABLE debounce enable / disable	0: TXDISABLE is not debounced
		1: TXDISABLE is debounced. Glitches less than 5 ms are rejected.
		Set this bit to 1 if a mechanical switch is used for TXDISABLE. Set to 0 for normal operation to assure compliance to the SFP MSA.
D[3]	RXLOS Polarity	0: The RXLOS output is low for normal operation and high with a loss of signal condition.
		1: The RXLOS output is high for normal operation (signal detected) and low with a loss of signal (no signal detected) condition.
D[2]	RXLOS/TRSOUT Select	0: RXLOS is selected for output
		1: TRRSOUT is selected for output
D[1]	Temperature resolution	0: Temperature is measured to a resolution of 1 ºC (eight-bit resolution)
		1: Temperature is measured to a resolution of 0.5 ºC (nine-bit resolution)
D[0]	TXFAULT clear mode	0: TXFAULT remains set until TXDISABLE is toggled
		1: TXFAULT is in continuous mode and follows the state of the faults
		In continuous mode, the fault conditions asserting TXFAULT are not registered and turn on and off According to the MIC3003's operation.

Power-On Hour Meter Data (POHDATA)							
D[7] read/write	D[6] read/write	D[5] read/write	D[4] read/write	D[3] read/write	D[2] read/write	D[1] read/write	D[0] read/write
Default values for all bytes when the MIC3003 is shipped			0000 0000$_b$ = 00$_h$				
Serial address			A6$_h$				
Byte addresses			32 – 34 and 36 - 38 = 20$_h$ - 22$_h$ and 24$_h$ - 26$_h$				
These registers are used for backing up the POH result during power cycles. At power-up, the POH meter selects the larger of the two values as the initial count. Incremental results are stored in alternate register pairs. The power-on hour meter may be reset or preset by writing to these registers.							
These registers are not typically intended to be used by the OEM or end user. The current value of the power-on hours meter may be read from the POHh and POHl registers.							
If it is necessary to preset the power-on hours meter to a specific value, please consult the factory for the exact format to be written to the POHDATA registers.							
These registers are non-volatile and will be maintained through power and reset cycles. A valid OEM password is required for access to these registers.							

Byte	Weight
3	POHA, high-byte
2	POHA, low-byte
1	POHB, high-byte
0	POHB, low-byte

OEM Scratchpad Registers (SCRATCHn)	
Default value	$0000\ 0000_b = 00_h$
Serial address	$A6_h$
Byte addresses	136 - 207 (88 - CF_h)
	208 - 221 (D0 - DD_h): This area is part of the OEM scratch pad only if the hard-coded delimiters option for receive power linearization is used.
	222 - 250 (DE - FA_h)
The scratchpad registers are general-purpose non-volatile memory locations. They can be freely read from and written to any time the MIC3003 is in OEM mode.	

RX Power Coefficient Look-up Table (RXLUTn)	
Default values	Offset coefficients are set to a default value of zero
	Slope coefficients are set to a default value of 1.0
Serial address	$A6_h$
Byte addresses	40 - 71 = 28_h - 47_h
These registers hold the receive power slope and offset coefficients used to calibrate the MIC3003's ADC receive power result in internal calibration mode.	
These registers are non-volatile and will be maintained through power and reset cycles. A valid OEM password is required for access to these registers.	

A6 Address	Bytes	Definition
40 (28h)	RXSLP0h	RX slope 0, high byte
41 (29h)	RXSLP0l	RX slope 0, low byte
42 (2Ah)	RXOFF0h	RX offset 0, high byte
43 (2Bh)	RXOFF0l	RX offset 0, low byte
44 (2Ch)	RXSLP1h	RX slope 1, high byte
45 (2Dh)	RXSLP1l	RX slope 1, low byte
46 (2Eh)	RXOFF1h	RX offset 1, high byte
47 (2Fh)	RXOFF1l	RX offset 1, low byte
•	•	
•	•	
•	•	
68 (44h)	RXSLP7h	RX slope 7, high byte
69 (45h)	RXSLP7l	RX slope 7, low byte
70 (46h)	RXOFF7h	RX offset 7, high byte
71 (47h)	RXOFF7l	RX offset 7, low byte

| Calibration Constants (CALCOEFn)	
Default values	Offset coefficients are set to a default value of zero
	Slope coefficients are set to a default value of 1.0
Serial address	A6$_h$
Byte addresses	74 - 87 = 4Ah - 57h
These registers hold the slope and offset coefficients used to calibrate the MIC3003's ADC results in internal calibration mode. Note that the temperature offset is also used in external calibration mode; but this can be disabled in OEMCFG5 (or by setting the offset coefficient to zero). These registers are non-volatile and will be maintained through power and reset cycles. A valid OEM password is required for access to these registers.	

A6 Address	Bytes	Definition
74 (4Ah)	TOFFh	Temperature offset (six-bit signed offset, .5 C resolution)
75 (4Bh)	TOFF0l	Not used
76 (4Ch)	VSLP0h	Voltage slope, high byte
77 (4Dh)	VSLP0l	Voltage slope, low byte
78 (4Eh)	VOFFh	Voltage offset, high byte
79 (4Fh)	VOFF0l	Voltage offset, low byte
80 (50h)	ISLP0h	Bias current slope, high byte
81 (51h)	ISLP0l	Bias current slope, low byte
82 (52h)	IOFFh	Bias current offset, high byte
83 (53h)	IOFF0l	Bias current offset, low byte
84 (54h)	TXSLPh	TX Power slope, high byte
85 (55h)	TXSLPl	TX Power slope, low byte
86 (56h)	TXOFFh	TX Power offset, high byte
87 (57h)	TXOFFl	TX Power offset, low byte

Manufacturer ID Register (MFG_ID) Identifies Micrel as the manufacturer of the device. Always returns 2Ah							
D[7] read-only	D[6] read-only	D[5] read-only	D[4] read-only	D[3] read-only	D[2] read-only	D[1] read-only	D[0] read-only
Default value		0010 1010$_b$ = 2A$_h$					
Serial address		A6$_h$					
Byte address		254 = FE$_h$					
The value in this register, in combination with the DEV_ID register, serves to identify the MIC3003 and its revision number to software. This register is read-only.							

Bit(s)	Function	Operation
D[7:0]	Identifies Micrel as the manufacturer of the device. Always returns 2A$_h$.	Read-only. Always returns A$_h$

Device ID Register (DEV_ID)							
D[7] read-only	D[6] read-only	D[5] read-only	D[4] read-only	D[3] read-only	D[2] read-only	D[1] read-only	D[0] read-only
MIC3003 Device ID always reads 0 at D[7-6] and 11 at D[5-4]				Die Revision			
Default value			0011 XXXX$_b$ = 3X$_h$				
Serial address			A6$_h$				
Byte address			255 = FF$_h$				
The value in this register, in combination with the MFG_ID register, serve to identify the MIC3003 and its revision number to software. This register is read-only.							

Package Information
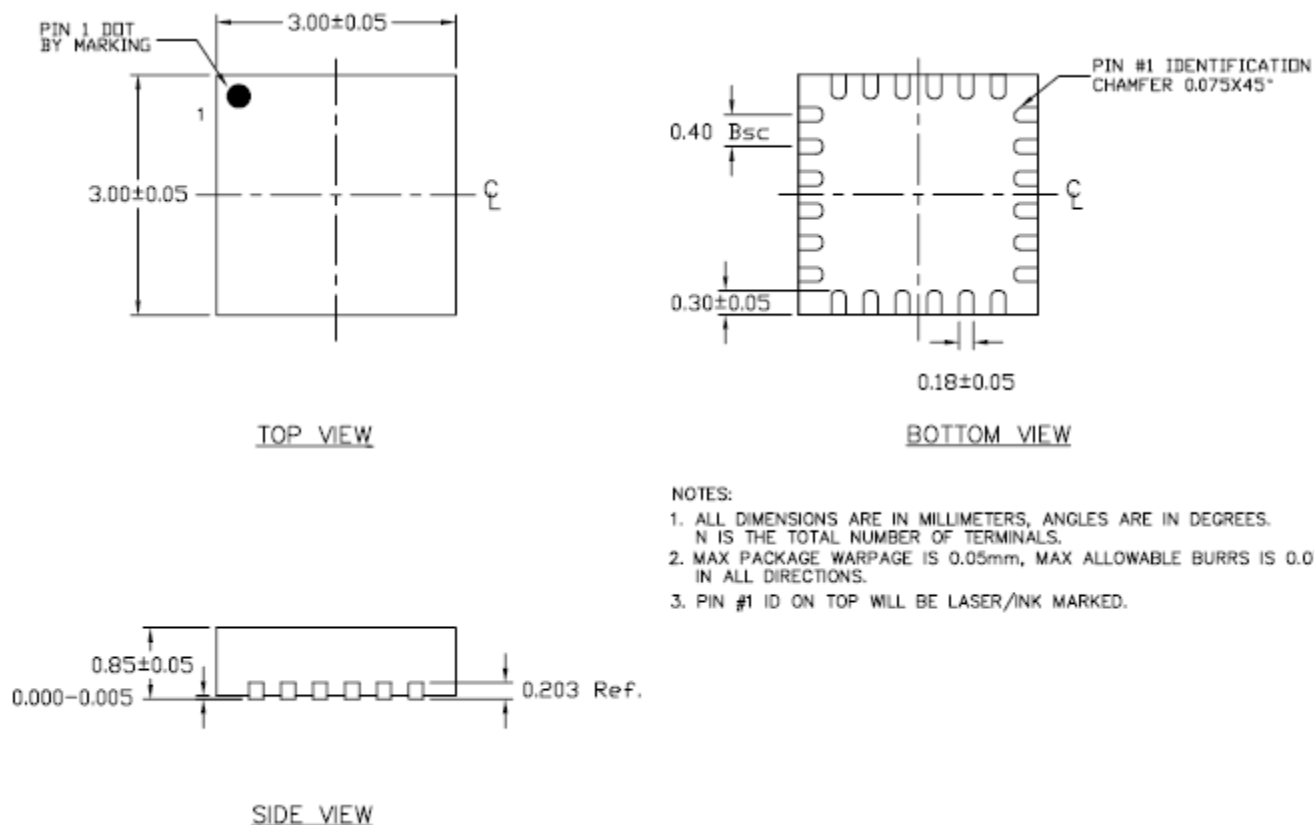
24-Pin MLF® (MLF-24)

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Благодаря сотрудничеству с мировыми поставщиками мы осуществляем комплексные и плановые поставки широчайшего спектра электронных компонентов.

Собственная эффективная логистика и склад в обеспечивает надежную поставку продукции в точно указанные сроки по всей России.

Мы осуществляем техническую поддержку нашим клиентам и предпродажную проверку качества продукции. На все поставляемые продукты мы предоставляем гарантию .

Осуществляем поставки продукции под контролем ВП МО РФ на предприятия военно-промышленного комплекса России , а также работаем в рамках 275 ФЗ с открытием отдельных счетов в уполномоченном банке. Система менеджмента качества компании соответствует требованиям ГОСТ ISO 9001.

Минимальные сроки поставки, гибкие цены, неограниченный ассортимент и индивидуальный подход к клиентам являются основой для выстраивания долгосрочного и эффективного сотрудничества с предприятиями радиоэлектронной промышленности, предприятиями ВПК и научно-исследовательскими институтами России.

С нами вы становитесь еще успешнее!

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