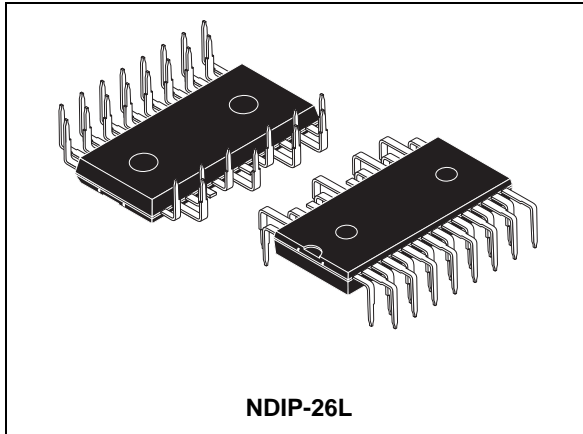


## SLLIMM™-nano (small low-loss intelligent molded module) IPM, 3 A - 600 V 3-phase IGBT inverter bridge

Datasheet - production data



### Features

- IPM 3 A, 600 V, 3-phase IGBT inverter bridge including control ICs for gate driving and freewheeling diodes
- Optimized for low electromagnetic interference
- $V_{CE(sat)}$  negative temperature coefficient
- 3.3 V, 5 V, 15 V CMOS/TTL input comparators with hysteresis and pull-down resistor
- Undervoltage lockout
- Internal bootstrap diode
- Interlocking function
- Optimized pinout for easy board layout
- 85 k $\Omega$  NTC for temperature control (UL1434 CA 2 and 4)

### Applications

- 3-phase inverters for motor drives
- Dish washers, refrigerator compressors, heating systems, air-conditioning fans, draining and recirculation pumps

### Description

This intelligent power module implements a compact, high-performance AC motor drive in a simple, rugged design. It is composed of six IGBTs with freewheeling diodes and three half-bridge HVICs for gate driving, providing low electromagnetic interference (EMI) characteristics with optimized switching speed. The package is optimized for thermal performance and compactness in built-in motor applications, or other low power applications where assembly space is limited. SLLIMM™ is a trademark of STMicroelectronics.

**Table 1. Device summary**

Order code	Marking	Package	Packaging
STGIPN3H60AT	GIPN3H60AT	NDIP-26L	Tube

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# 1 Internal schematic diagram and pin configuration

Figure 1. Internal schematic diagram

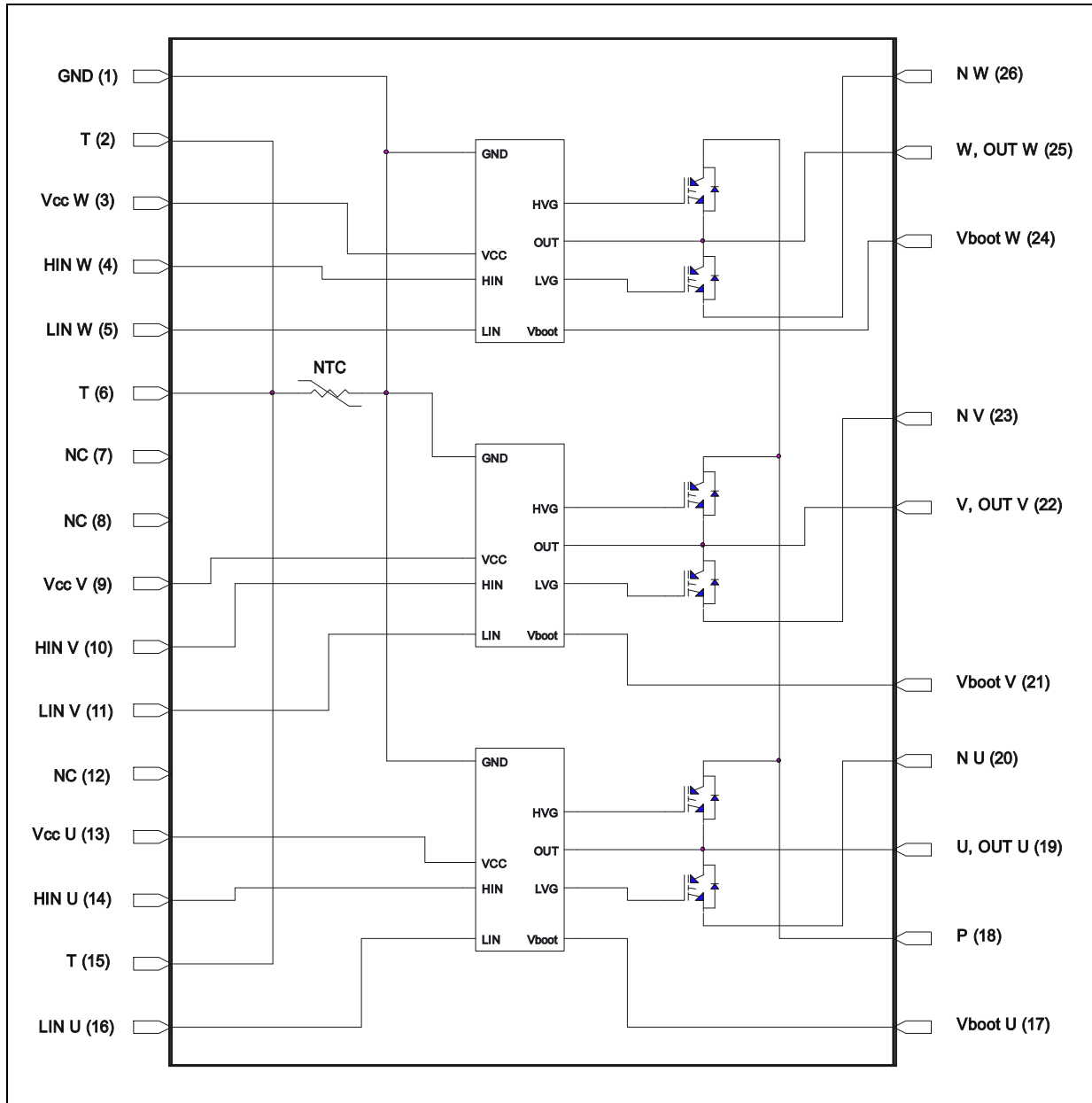
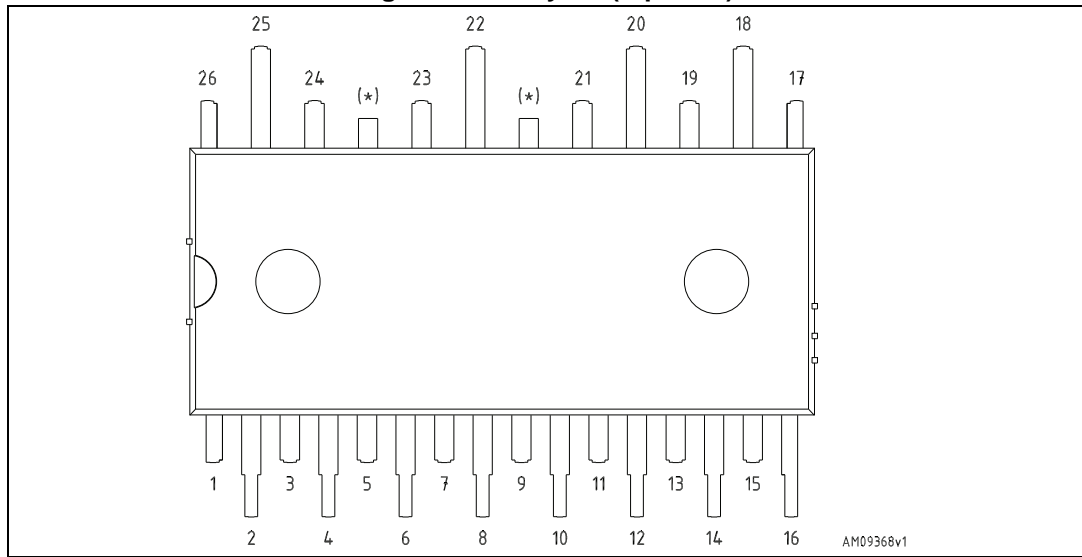


Table 2. Pin description

Pin	Symbol	Description
1	GND	Ground
2	T	NTC thermistor terminal
3	V <sub>CC</sub> W	Low voltage power supply W phase
4	HIN W	High side logic input for W phase
5	LIN W	Low side logic input for W phase
6	T	NTC thermistor terminal
7	NC	Not connected
8	NC	Not connected
9	V <sub>CC</sub> V	Low voltage power supply V phase
10	HIN V	High side logic input for V phase
11	LIN V	Low side logic input for V phase
12	NC	Not connected
13	V <sub>CC</sub> U	Low voltage power supply for U phase
14	HIN U	High side logic input for U phase
15	T	NTC thermistor terminal
16	LIN U	Low side logic input for U phase
17	V <sub>boot</sub> U	Bootstrap voltage for U phase
18	P	Positive DC input
19	U	U phase output
20	N <sub>U</sub>	Negative DC input for U phase
21	V <sub>boot</sub> V	Bootstrap voltage for V phase
22	V	V phase output
23	N <sub>V</sub>	Negative DC input for V phase
24	V <sub>boot</sub> W	Bootstrap voltage for W phase
25	W	W phase output
26	N <sub>W</sub>	Negative DC input for W phase

Figure 2. Pin layout (top view)



(\*) Dummy pin internally connected to P (positive DC input).

## 2 Electrical ratings

### 2.1 Absolute maximum ratings

**Table 3. Inverter part**

Symbol	Parameter	Value	Unit
$V_{CES}$	Each IGBT collector emitter voltage ( $V_{IN}^{(1)} = 0$ )	600	V
$\pm I_C^{(2)}$	Each IGBT continuous collector current at $T_C = 25\text{ °C}$	3	A
$\pm I_{CP}^{(3)}$	Each IGBT pulsed collector current	18	A
$P_{TOT}$	Each IGBT total dissipation at $T_C = 25\text{ °C}$	8	W

1. Applied between  $HIN_i$ ,  $LIN_i$  and  $G_{ND}$  for  $i = U, V, W$ .
2. Calculated according to the iterative formula:

$$I_C(T_C) = \frac{T_{j(max)} - T_C}{R_{thj-c} \times V_{CE(sat)(max)}(T_{j(max)}, I_C(T_C))}$$

3. Pulse width limited by max junction temperature.

**Table 4. Control part**

Symbol	Parameter	Min.	Max.	Unit
$V_{OUT}$	Output voltage applied between $OUT_U$ , $OUT_V$ , $OUT_W$ - GND	$V_{boot} - 18$	$V_{boot} + 0.3$	V
$V_{CC}$	Low voltage power supply	- 0.3	18	V
$V_{boot}$	Bootstrap voltage	- 0.3	618	V
$V_{IN}$	Logic input voltage applied between $HIN_i$ , $LIN_i$ and $G_{ND}$ for $i = U, V, W$	- 0.3	$V_{CC} + 0.3$	V
$\Delta V_{OUT/dT}$	Allowed output slew rate		50	V/ns

**Table 5. Total system**

Symbol	Parameter	Value	Unit
$V_{ISO}$	Isolation withstand voltage applied between each pin and heatsink plate (AC voltage, $t = 60\text{ sec.}$ )	1000	V
$T_J$	Power chips operating junction temperature	-40 to 150	°C
$T_C$	Module case operating temperature	-40 to 125	°C

## 2.2 Thermal data

Table 6. Thermal data

Symbol	Parameter	Value	Unit
$R_{thJA}$	Thermal resistance junction-ambient	50	°C/W

### 3 Electrical characteristics

$T_j = 25\text{ }^\circ\text{C}$  unless otherwise specified.

**Table 7. Inverter part**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$V_{CE(sat)}$	Collector-emitter saturation voltage	$V_{CC} = V_{boot} = 15\text{ V}$ , $V_{IN}^{(1)} = 0 - 5\text{ V}$ , $I_C = 1\text{ A}$	-	2.15	2.6	V
		$V_{CC} = V_{boot} = 15\text{ V}$ , $V_{IN}^{(1)} = 0 - 5\text{ V}$ , $I_C = 1\text{ A}$ , $T_J = 125\text{ }^\circ\text{C}$	-	1.65		
$I_{CES}$	Collector-cut off current ( $V_{IN}^{(1)} = 0$ "logic state")	$V_{CE} = 550\text{ V}$ , $V_{CC} = V_{Boot} = 15\text{ V}$	-		250	$\mu\text{A}$
$V_F$	Diode forward voltage	$V_{IN}^{(1)} = 0$ "logic state", $I_C = 1\text{ A}$	-		1.7	V
<b>Inductive load switching time and energy</b>						
$t_{on}$	Turn-on time	$V_{DD} = 300\text{ V}$ , $V_{CC} = V_{boot} = 15\text{ V}$ , $V_{IN}^{(1)} = 0 - 5\text{ V}$ , $I_C = 1\text{ A}$ (see <a href="#">Figure 4</a> )	-	275		ns
$t_{c(on)}$	Crossover time (on)		-	90		
$t_{off}$	Turn-off time		-	890		
$t_{c(off)}$	Crossover time (off)		-	125		
$t_{rr}$	Reverse recovery time		-	50		$\mu\text{J}$
$E_{on}$	Turn-on switching losses		-	18		
$E_{off}$	Turn-off switching losses	-	13			

1. Applied between  $HIN_i$ ,  $\overline{LIN}_i$  and  $G_{ND}$  for  $i = U, V, W$  ( $\overline{LIN}_i$  inputs are active-low).

*Note:*  $t_{ON}$  and  $t_{OFF}$  include the propagation delay time of the internal drive.  $t_{C(ON)}$  and  $t_{C(OFF)}$  are the switching time of IGBT itself under the internally given gate driving condition.

**Figure 3. Switching time test circuit**

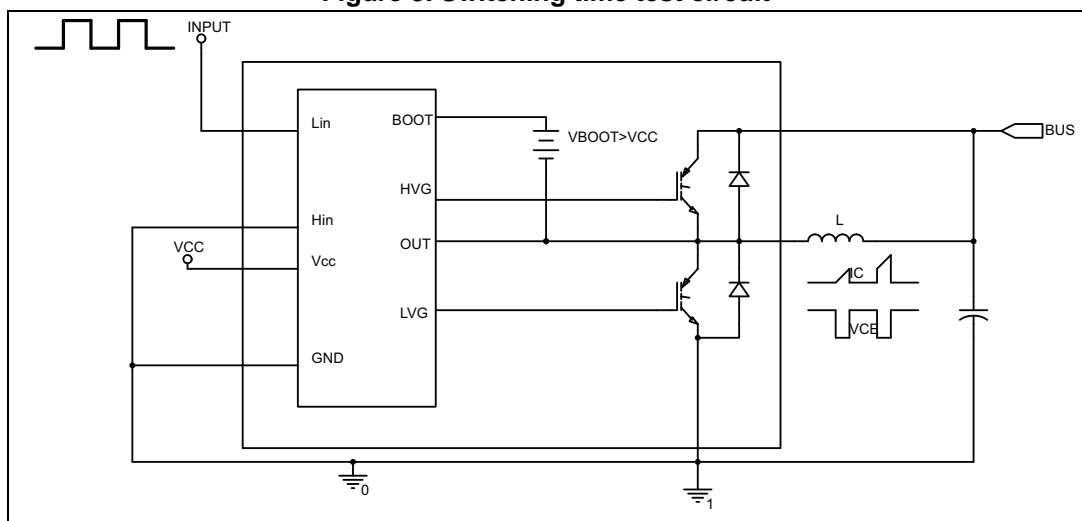
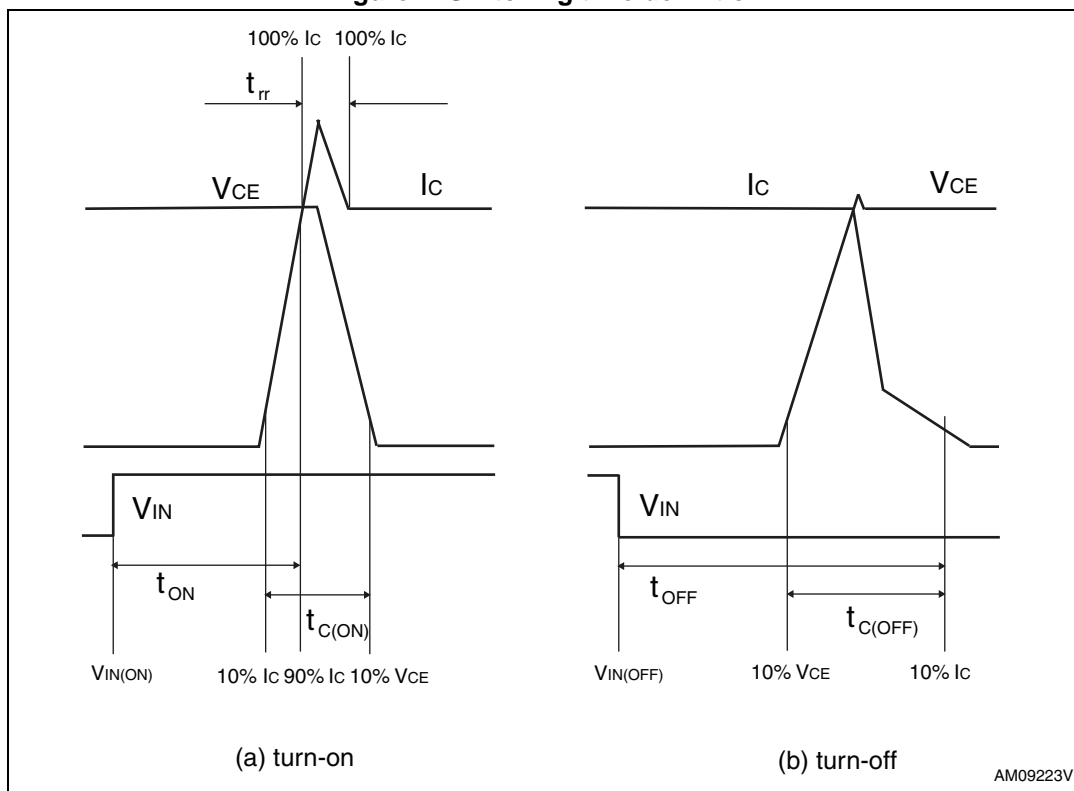




Figure 4. Switching time definition



### 3.1 Control part

Table 8. Low voltage power supply ( $V_{CC} = 15\text{ V}$  unless otherwise specified)

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$V_{CCthON}$	Undervoltage turn-on threshold		9.1	9.6	10.1	V
$V_{CCthOFF}$	Undervoltage turn-off threshold		7.9	8.3	8.8	V
$V_{CChys}$	Undervoltage hystereses		0.9			V
$I_{qccu}$	Undervoltage quiescent supply current	$V_{CC} < 7.9\text{ V}$		250	330	$\mu\text{A}$
$I_{qcc}$	Quiescent current	$V_{CC} = 15\text{ V}$		350	450	$\mu\text{A}$

**Table 9. Bootstrapped voltage ( $V_{CC} = 15\text{ V}$  unless otherwise specified)**

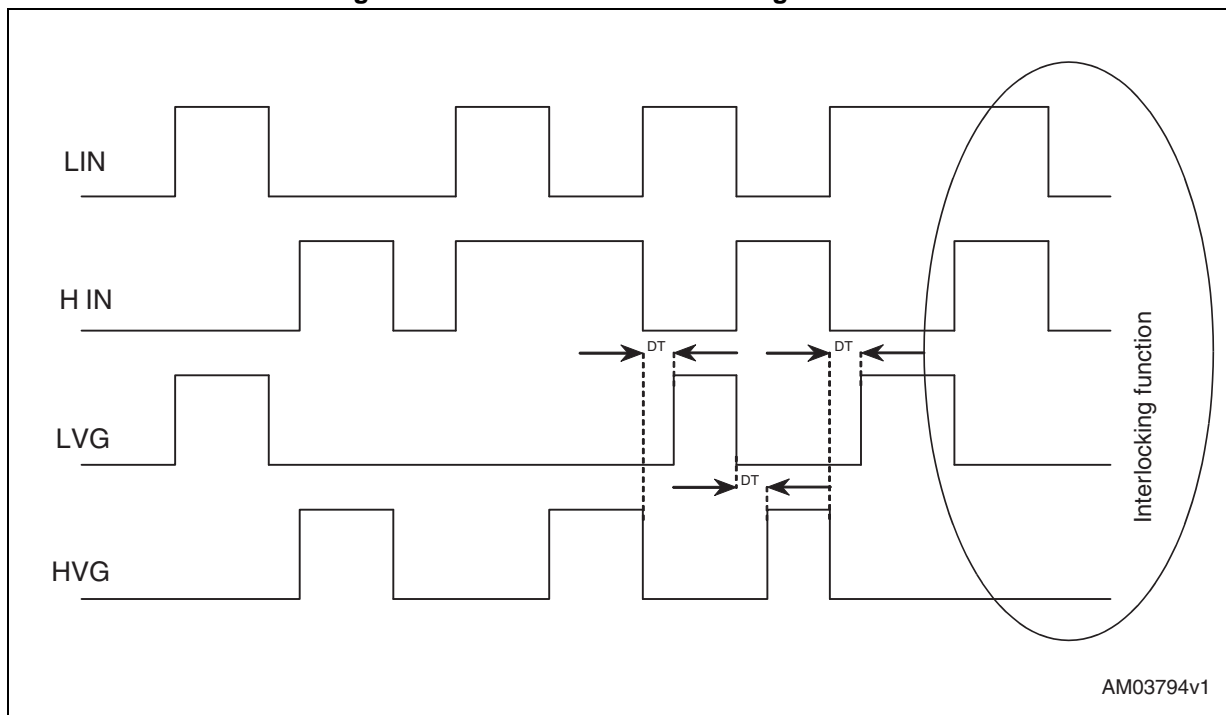
Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$V_{boot\_thON}$	Undervoltage turn-on threshold		8.5	9.5	10.5	V
$V_{boot\_thOFF}$	Undervoltage turn-off threshold		7.2	8.3	9.2	V
$V_{boot\_hys}$	Undervoltage hysteresis		0.9			V
$I_{qboot}$	Quiescent current				250	$\mu\text{A}$
$R_{DS(on)}$	Bootstrap driver on-resistance	$V_{CC} > 12.5\text{ V}$		125		$\Omega$

**Table 10. Logic inputs ( $V_{CC} = 15\text{ V}$  unless otherwise specified) <sup>(1)</sup>**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$V_{il}$	Low level logic input voltage				1.1	V
$V_{ih}$	High level logic input voltage		1.8			V
$I_{il}$	Low level logic input current	$V_{IN}^{(2)} = 0$	-1			$\mu\text{A}$
$I_{ih}$	High level logic input current	$V_{IN}^{(1)} = 15\text{ V}$		20	70	$\mu\text{A}$
Dt	Dead time			320		ns

1. See [Figure 5: Dead time and interlocking definition](#).
2. Applied between  $HIN_i$ ,  $LIN_i$  and  $G_{ND}$  for  $i = U, V, W$

**Figure 5. Dead time and interlocking definition**



3.1.1 NTC thermistor

Table 11. NTC thermistor

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit.
R <sub>25</sub>	Resistance	T = 25 °C		85		kΩ
R <sub>100</sub>	Resistance	T = 100 °C		5388		Ω
B	B-constant	T = 25 °C to 100 °C		4092		K
T	Operating temperature		-25		125	°C

Equation 1: resistance variation vs. temperature

$$R(T) = R_{25} \cdot e^{B \left( \frac{1}{T} - \frac{1}{298} \right)}$$

Where T are temperatures in Kelvins

Figure 6. NTC resistance vs. temperature

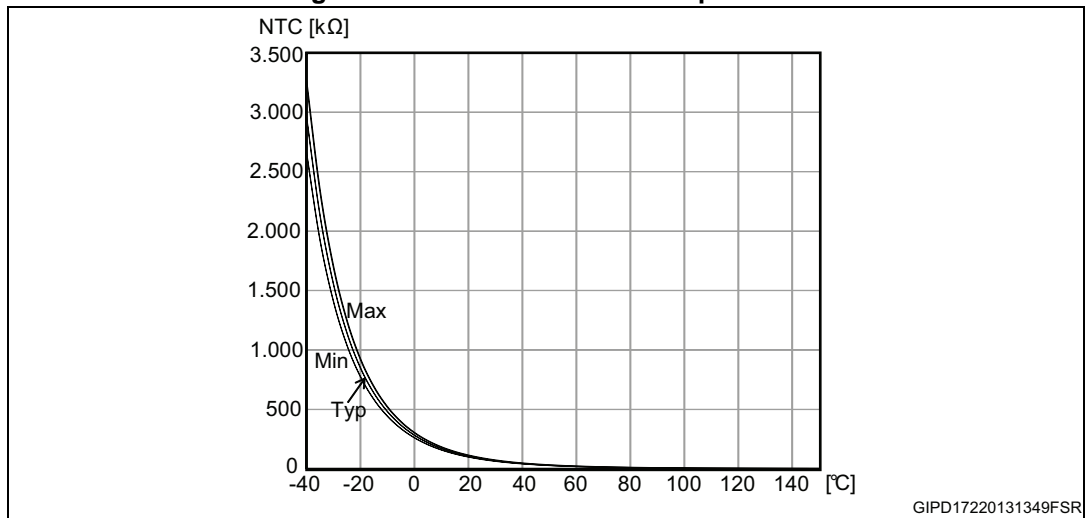
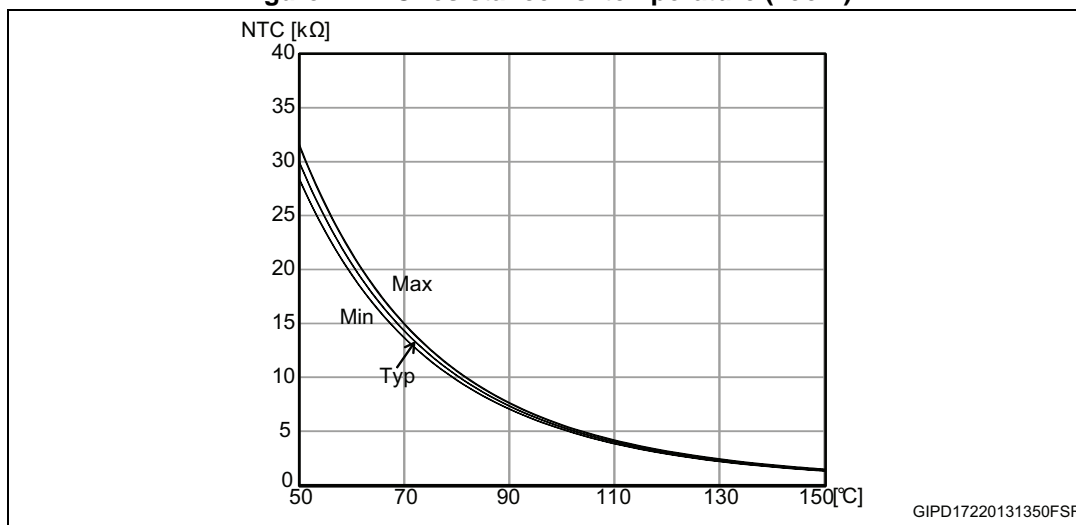
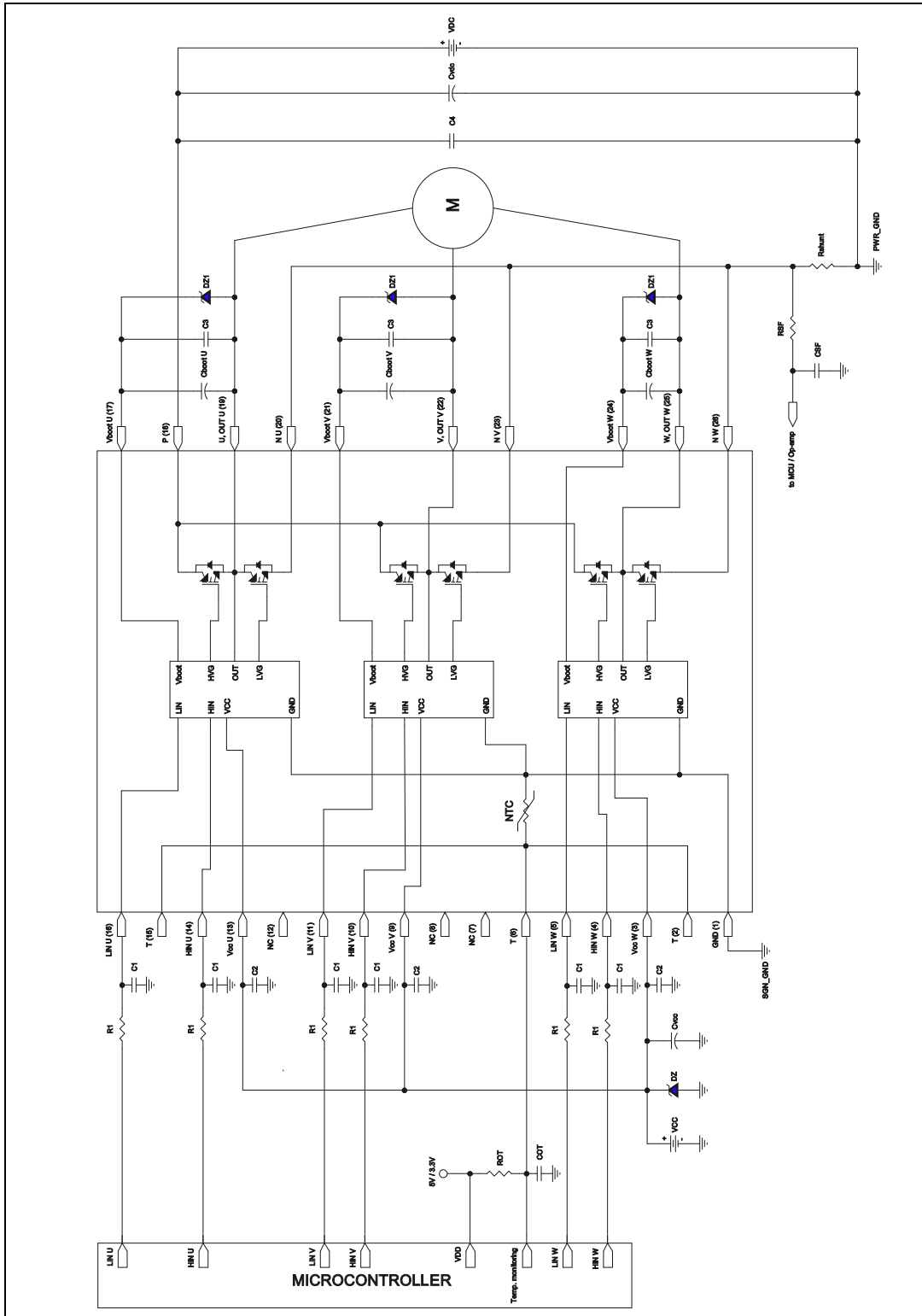


Figure 7. NTC resistance vs. temperature (zoom)



# 4 Application information

Figure 8. Typical application circuit



## 4.1 Recommendations

- Input signals HIN, LIN are active-high logic. A 500 k $\Omega$  (typ.) pull-down resistor is built-in for each input. To prevent input signal oscillation, the wiring of each input should be as short as possible and the use of RC filters (R1, C1) on each input signal is suggested. The filters should be done with a time constant of about 100ns and must be placed as close as possible to the IPM input pins.
- The bypass capacitor C<sub>vcc</sub> (aluminum or tantalum) is recommended to reduce the transient circuit demand on the power supply. In addition, a decoupling capacitor C2 (from 100 to 220 nF, ceramic with low ESR) is suggested, to reduce high frequency switching noise distributed on the power supply lines. It must be placed as close as possible to each V<sub>cc</sub> pin and in parallel to the bypass capacitor.
- The use of RC filter (RSF, CSF) for current monitoring is recommended to improve noise immunity. The filter must be placed as close as possible to the microcontroller or to the Op-amp.
- The decoupling capacitor C3 (from 100 to 220 nF, ceramic with low ESR), in parallel to each C<sub>boot</sub>, is recommended in order to filter high frequency disturbances.
- The Zener diodes DZ1 between the V<sub>cc</sub> pins and GND and in parallel to each C<sub>boot</sub> is suggested in order to prevent overvoltage.
- The decoupling capacitor C4 (from 100 to 220 nF, ceramic with low ESR) in parallel to the electrolytic capacitor C<sub>vdc</sub> is recommended, in order to prevent surge destruction. Both capacitors C4 and C<sub>vdc</sub> should be placed as close as possible to the IPM (C4 has priority over C<sub>vdc</sub>).
- By integrating an application-specific type HVIC inside the module, direct coupling to the MCU terminals without an opto-coupler is possible.
- Low inductance shunt resistors should be used for phase leg current sensing
- In order to avoid malfunctions, the wiring between N pins, the shunt resistor and PWR\_GND should be as short as possible.
- It is recommended to connect SGN\_GND to PWR\_GND at only one point (near the terminal of shunt resistor), in order to avoid any malfunction due to power ground fluctuation.

**Table 12. Recommended operating conditions**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
V <sub>PN</sub>	Supply voltage	Applied between P-Nu, Nv, Nw		300	500	V
V <sub>CC</sub>	Control supply voltage	Applied between V <sub>CC</sub> -GND	12	15	17	V
V <sub>BS</sub>	High side bias voltage	Applied between V <sub>BOOTi</sub> -OUT <sub>i</sub> for i = U, V, W	11.5		17	V
t <sub>dead</sub>	Blanking time to prevent Arm-short	For each input signal	1.5			$\mu$ s
f <sub>PWM</sub>	PWM input signal	-40°C < T <sub>c</sub> < 100°C -40°C < T <sub>j</sub> < 125°C			25	kHz
T <sub>C</sub>	Case operation temperature				100	°C

Note: For further details refer to AN4043.

## 5 Package mechanical data

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK® packages, depending on their level of environmental compliance. ECOPACK® specifications, grade definitions and product status are available at: [www.st.com](http://www.st.com). ECOPACK is an ST trademark.

Figure 9. NDIP-26L drawing

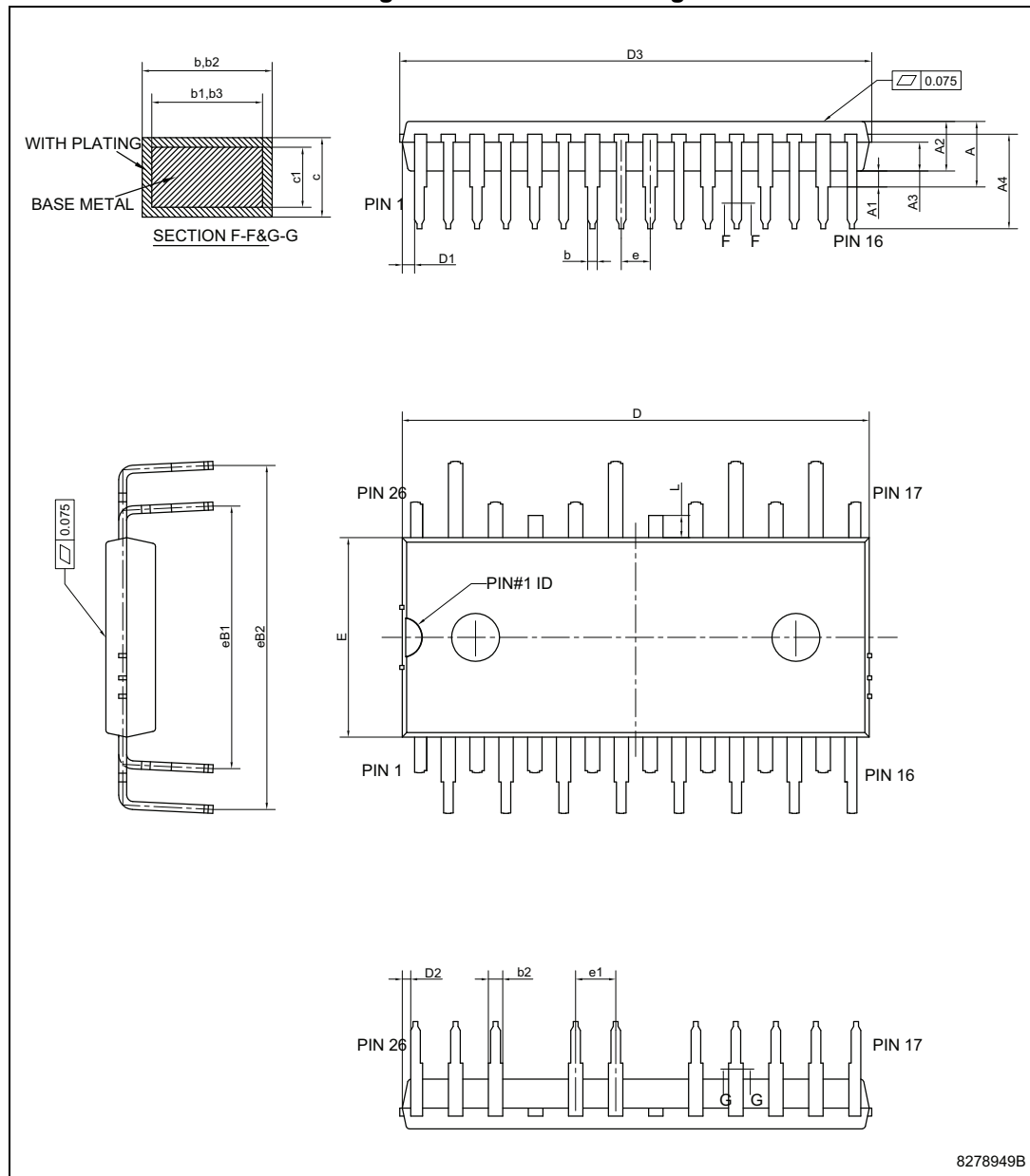


Table 13.NDIP-26L mechanical data

Dim.	mm.		
	Min.	Typ.	Max.
A			4.40
A1	0.80	1.00	1.20
A2	3.00	3.10	3.20
A3	1.70	1.80	1.90
A4	5.70	5.90	6.10
b	0.53		0.72
b1	0.52	0.60	0.68
b2	0.83		1.02
b3	0.82	0.90	0.98
c	0.46		0.59
c1	0.45	0.50	0.55
D	29.05	29.15	29.25
D1	0.50	0.77	1.00
D2	0.35	0.53	0.70
D3			29.55
E	12.35	12.45	12.55
e	1.70	1.80	1.90
e1	2.40	2.50	2.60
eB1	16.10	16.40	16.70
eB2	21.18	21.48	21.78
L	1.24	1.39	1.54





## 6 Revision history

Table 14. Document revision history

Date	Revision	Changes
30-Sep-2014	1	Initial release.

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