



# MOTOROLA

## MPQ6001,N MPQ6002,N MPQ6501,N MPQ6502,N

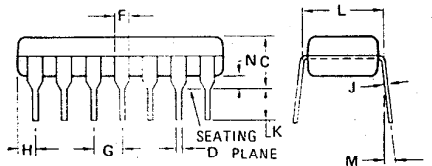
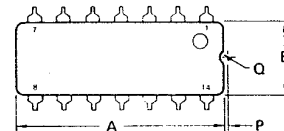
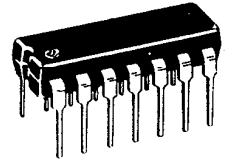
### QUAD DUAL-IN-LINE SILICON ANNULAR<sup>♦</sup> COMPLEMENTARY PAIR TRANSISTORS

... designed for high-speed switching circuits, DC to VHF amplifier applications and complementary circuitry.

- Maximum Power Dissipation @  $T_A = 25^\circ\text{C}$   
 $P_D = 1.25$  Watts – MPQ6001,MPQ6002,MPQ6501,MPQ6502  
 $= 0.9$  Watts – MPQ6001N,MPQ6002N,MPQ6501N,MPQ6502N
- DC Current Gain Specified – 1.0 to 300 mAdc
- High Current-Gain-Bandwidth Product –  
 $f_T = 400$  MHz (Typ) @  $I_C = 50$  mAdc
- NPN Transistor Similar to 2N2218 or 2N2219
- PNP Transistor Similar to 2N2904 or 2N2905
- Compact Size Compatible with IC Automatic Insertion Equipment

### QUAD DUAL-IN-LINE SILICON COMPLEMENTARY PAIR TRANSISTORS

CASE 646  
PLASTIC PACKAGE



#### MAXIMUM RATINGS

Rating	Symbol	Value	Unit	
Collector-Emitter Voltage	$V_{CEO}$	30	Vdc	
Collector-Base Voltage	$V_{CB}$	60	Vdc	
Emitter-Base Voltage	$V_{EB}$	5.0	Vdc	
Collector Current – Continuous	$I_C$	500	mAdc	
		Each Transistor	Four Transistors Equal Power	
Total Power Dissipation @ $T_A = 25^\circ\text{C}$ (1) MPQ6001, MPQ6002, MPQ6501, MPQ6502 MPQ6001N, MPQ6002N, MPQ6501N, MPQ6502N Derate above $25^\circ\text{C}$	$P_D$	0.65 0.5	1.25 0.9	Watts mW/ $^\circ\text{C}$
Total Power Dissipation @ $T_C = 25^\circ\text{C}$ MPQ6001, MPQ6002, MPQ6501, MPQ6502 MPQ6001N, MPQ6002N, MPQ6501N, MPQ6502N Derate above $25^\circ\text{C}$	$P_D$	5.18 4.0	10 7.2	Watts mW/ $^\circ\text{C}$
Operating and Storage Junction Temperature Range	$T_J, T_{stg}$	-55 to +150	$^\circ\text{C}$	

DIM	MILLIMETERS		INCHES	
	MIN	MAX	MIN	MAX
A	18.16	18.80	0.715	0.740
B	6.10	6.60	0.240	0.260
C	4.06	4.57	0.160	0.180
D	0.38	0.51	0.015	0.020
F	1.02	1.52	0.040	0.060
G	2.54 BSC		0.100 BSC	
H	1.32	1.83	0.052	0.072
J	0.20	0.30	0.008	0.012
K	2.92	3.43	0.115	0.135
L	7.37	7.87	0.290	0.310
M	-	$10^\circ$	-	$10^\circ$
N	0.51	1.02	0.020	0.040
P	0.13	0.38	0.005	0.015
Q	0.51	0.76	0.020	0.030

NOTES:

- LEADS WITHIN 0.13 mm (0.005) RADIUS OF TRUE POSITION AT SEATING PLANE AT MAXIMUM MATERIAL CONDITION.
- DIMENSION "L" TO CENTER OF LEADS WHEN FORMED PARALLEL

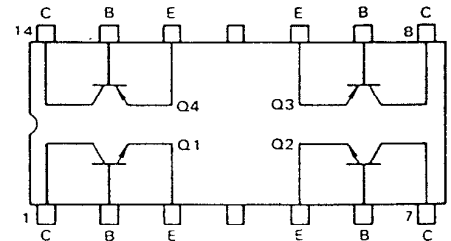
(1) Second Breakdown occurs at power levels greater than 3 times the power dissipation rating.

#### THERMAL CHARACTERISTICS

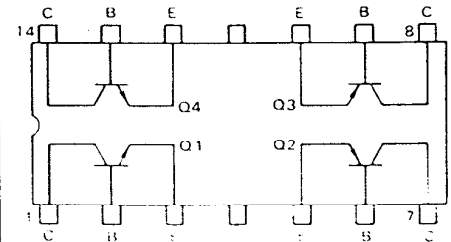
Characteristic	Junction to Case	Junction to Ambient	Unit
Thermal Resistance			$^\circ\text{C}/\text{W}$
Each Die			
MPQ6001, MPQ6002, MPQ6501, MPQ6502	125	193	
MPQ6001N, MPQ6002N, MPQ6501N, MPQ6502N	151	250	
Effective, 4 Die			
MPQ6001, MPQ6002, MPQ6501, MPQ6502	41.6	100	
MPQ6001N, MPQ6002N, MPQ6501N, MPQ6502N	52	134	
Coupling Factors			%
Q1-Q4 or Q2-Q3	MPQ6001, MPQ6002 MPQ6501, MPQ6502	30	60
	MPQ6001N, MPQ6002N MPQ6501N, MPQ6502N	34	70
	MPQ6001, MPQ6002 MPQ6501, MPQ6502	2.0	24
Q1-Q2 or Q3-Q4	MPQ6001N, MPQ6002N MPQ6501N, MPQ6502N	2.0	26

#### CONNECTION DIAGRAM

MPQ6001,N, MPQ6002,N,



MPQ6501,N, MPQ6502,N



<sup>♦</sup> Annular Semiconductors Patented by Motorola Inc.

## THERMAL COUPLING AND EFFECTIVE THERMAL RESISTANCE

In multiple chip devices, coupling of heat between die occurs. The junction temperature can be calculated as follows:

$$(1) \Delta T_{J1} = R_{\theta 1} P_{D1} + R_{\theta 2} K_{\theta 2} P_{D2} + R_{\theta 3} K_{\theta 3} P_{D3} + R_{\theta 4} K_{\theta 4} P_{D4}$$

Where  $\Delta T_{J1}$  is the change in junction temperature of die 1  
 $R_{\theta 1}$  thru 4 is the thermal resistance of die 1 through 4  
 $P_{D1}$  thru 4 is the power dissipated in die 1 through 4  
 $K_{\theta 2}$  thru 4 is the thermal coupling between die 1 and die 2 through 4.

An effective package thermal resistance can be defined as follows:

$$(2) R_{\theta(EFF)} = \Delta T_{J1} / P_{DT}$$

where:  $P_{DT}$  is the total package power dissipation.

Assuming equal thermal resistance for each die, equation (1) simplifies to

$$(3) \Delta T_{J1} = R_{\theta 1} (P_{D1} + K_{\theta 2} P_{D2} + K_{\theta 3} P_{D3} + K_{\theta 4} P_{D4})$$

For the conditions where  $P_{D1} = P_{D2} = P_{D3} = P_{D4}$ ,  $P_{DT} = 4 P_D$ , equation (3) can be further simplified and by substituting into equation (2) results in

$$(4) R_{\theta(EFF)} = R_{\theta 1} (1 + K_{\theta 2} + K_{\theta 3} + K_{\theta 4}) / 4$$

Values for the coupling factors when either the case or the ambient is used as a reference are given in the table on page 1. If significant power is to be dissipated in two die, die at the opposite ends of the package should be used so that lowest possible junction temperatures will result.

### ELECTRICAL CHARACTERISTICS ( $T_A = 25^\circ\text{C}$ unless otherwise noted)

Characteristic	Symbol	Min	Typ	Max	Unit
<b>OFF CHARACTERISTICS</b>					
Collector-Emitter Breakdown Voltage(1) ( $I_C = 10 \text{ mAdc}$ , $I_B = 0$ )	$BV_{CEO}$	30	—	—	Vdc
Collector-Base Breakdown Voltage ( $I_C = 10 \mu\text{Adc}$ , $I_E = 0$ )	$BV_{CBO}$	60	—	—	Vdc
Emitter-Base Breakdown Voltage ( $I_E = 10 \mu\text{Adc}$ , $I_C = 0$ )	$BV_{EBO}$	5.0	—	—	Vdc
Collector Cutoff Current ( $V_{CB} = 50 \text{ Vdc}$ , $I_E = 0$ )	$I_{CBO}$	—	—	30	nAdc
Emitter Cutoff Current ( $V_{EB} = 3.0 \text{ Vdc}$ , $I_C = 0$ )	$I_{EBO}$	—	—	30	nAdc
<b>ON CHARACTERISTICS</b>					
DC Current Gain(1) ( $I_C = 1.0 \text{ mAdc}$ , $V_{CE} = 10 \text{ Vdc}$ )	$h_{FE}$	25 50	—	—	—
( $I_C = 10 \text{ mAdc}$ , $V_{CE} = 10 \text{ Vdc}$ )		35 75	—	—	—
( $I_C = 150 \text{ mAdc}$ , $V_{CE} = 10 \text{ Vdc}$ )		40 100	—	—	—
( $I_C = 300 \text{ mAdc}$ , $V_{CE} = 10 \text{ Vdc}$ )		20 30	—	—	—
Collector-Emitter Saturation Voltage (1) ( $I_C = 150 \text{ mAdc}$ , $I_B = 15 \text{ mAdc}$ ) ( $I_C = 300 \text{ mAdc}$ , $I_B = 30 \text{ mAdc}$ )	$V_{CE(sat)}$	—	—	0.4 1.4	Vdc
Base-Emitter Saturation Voltage (1) ( $I_C = 150 \text{ mAdc}$ , $I_B = 15 \text{ mAdc}$ ) ( $I_C = 300 \text{ mAdc}$ , $I_B = 30 \text{ mAdc}$ )	$V_{BE(sat)}$	—	—	1.3 2.0	Vdc
<b>DYNAMIC CHARACTERISTICS</b>					
Current-Gain-Bandwidth Product (1) ( $I_C = 50 \text{ mAdc}$ , $V_{CE} = 20 \text{ Vdc}$ , $f = 100 \text{ MHz}$ )	$f_T$	200	350	—	MHz
Output Capacitance ( $V_{CB} = 10 \text{ Vdc}$ , $I_E = 0$ , $f = 100 \text{ kHz}$ )	$C_{ob}$	—	6.0 4.5	8.0 8.0	pF
Input Capacitance ( $V_{EB} = 2.0 \text{ Vdc}$ , $I_C = 0$ , $f = 100 \text{ kHz}$ )	$C_{ib}$	—	20 17	30 30	pF
<b>SWITCHING CHARACTERISTICS</b>					
Turn-On Time ( $V_{CC} = 30 \text{ Vdc}$ , $V_{BE(on)} = 0.5 \text{ Vdc}$ , $I_C = 150 \text{ mAdc}$ , $I_B = 15 \text{ mAdc}$ , Figure 1)	$t_{on}$	—	30	—	ns
Turn-Off Time ( $V_{CC} = 30 \text{ Vdc}$ , $I_C = 150 \text{ mAdc}$ , $I_B = I_{B2} = 15 \text{ mAdc}$ , Figure 2)	$t_{off}$	—	225	—	ns

(1) Pulse Test: Pulse Width  $\leq 300 \mu\text{s}$ , Duty Cycle = 2%.

### NPN SATURATED SWITCHING TIME TEST CIRCUITS

For NPN Switching Tests, reverse the diodes, voltage polarities, and input pulses.

FIGURE 1 - NPN TURN-ON TIME

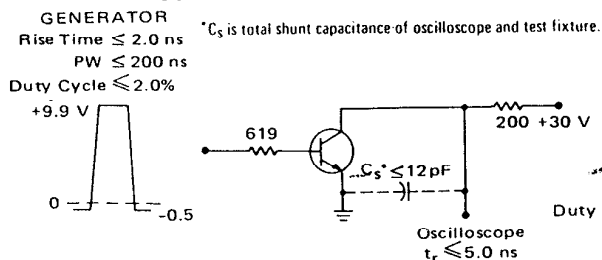
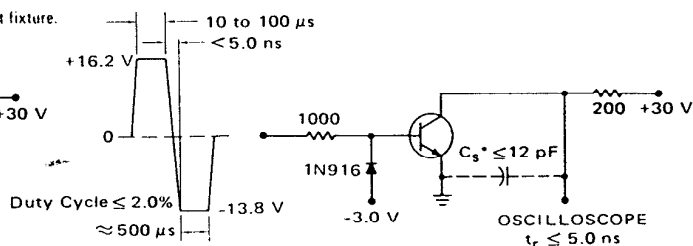


FIGURE 2 - NPN TURN-OFF TIME



**MOTOROLA Semiconductor Products Inc.**

NPN DATA

FIGURE 3 – NORMALIZED DC CURRENT GAIN

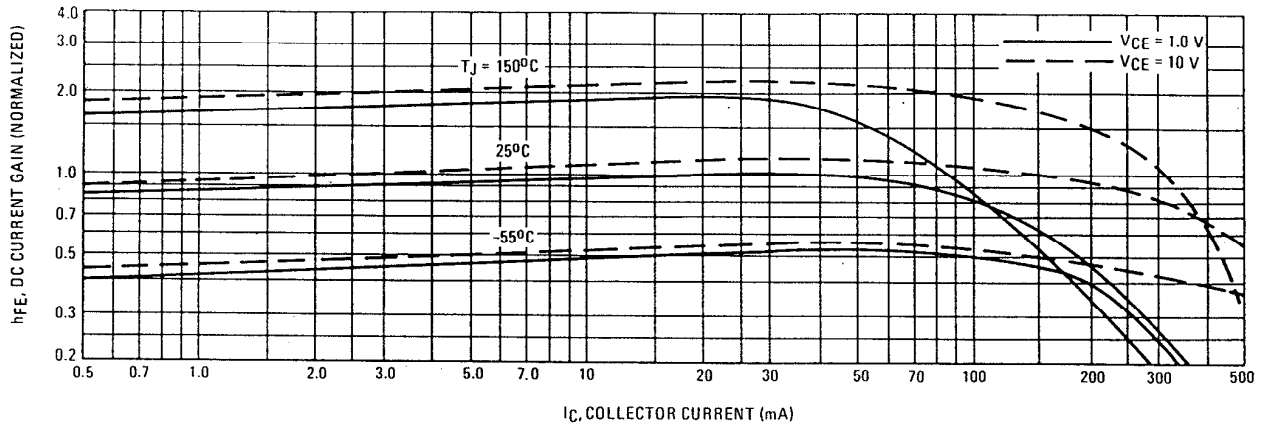


FIGURE 4 – "ON" VOLTAGES

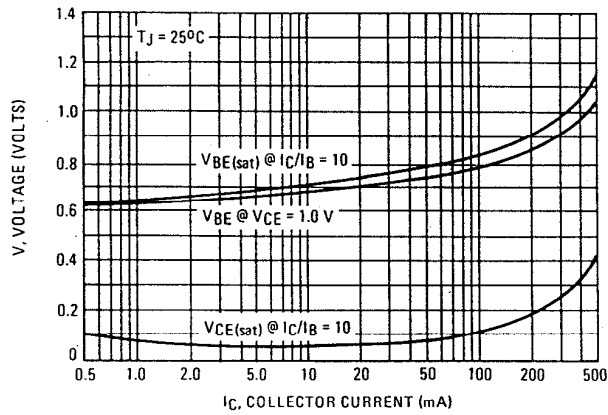
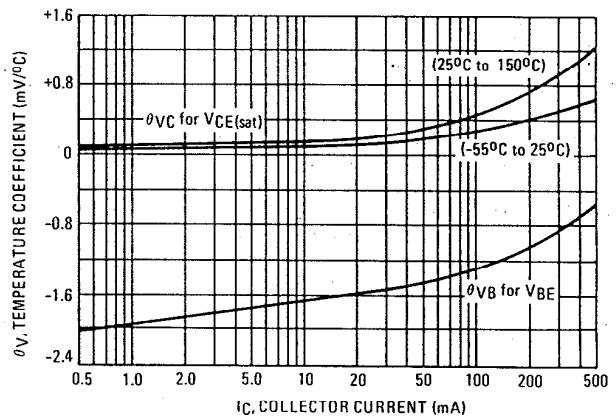


FIGURE 5 – TEMPERATURE COEFFICIENTS



NOISE FIGURE  
( $V_{CE} = 10\text{ Vdc}$ ,  $T_A = 25^\circ\text{C}$ )

FIGURE 6 – FREQUENCY EFFECTS

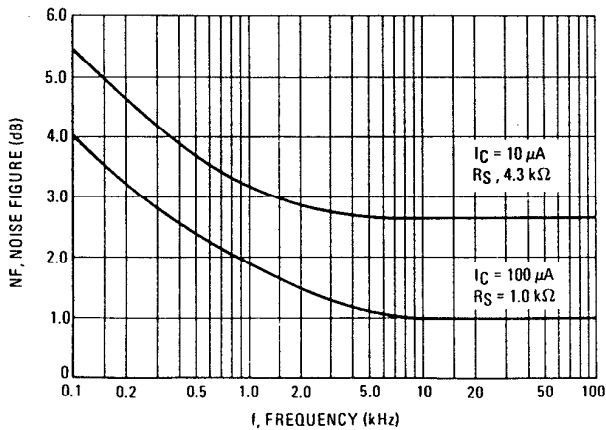
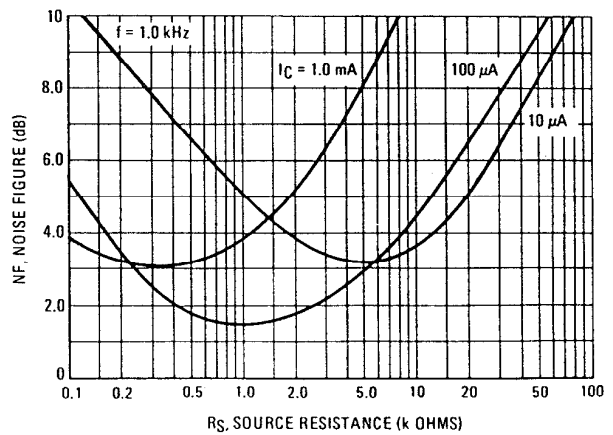
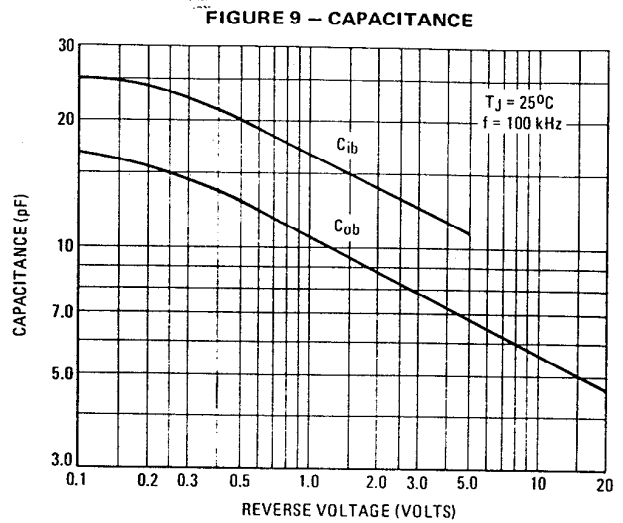
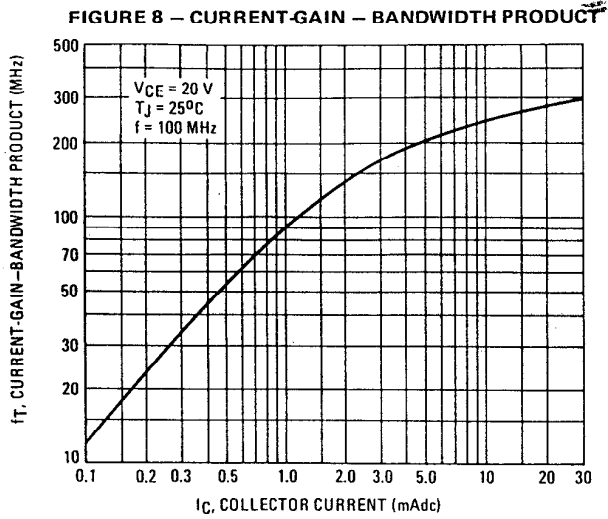


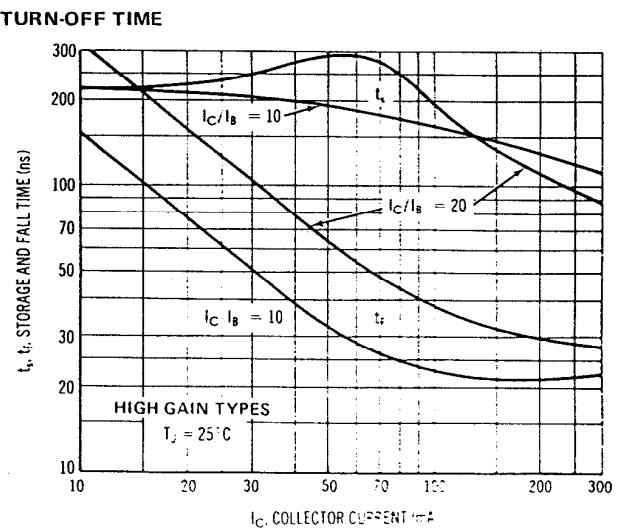
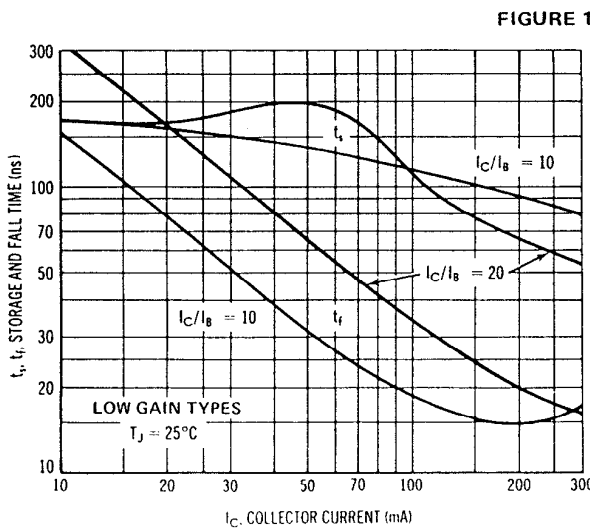
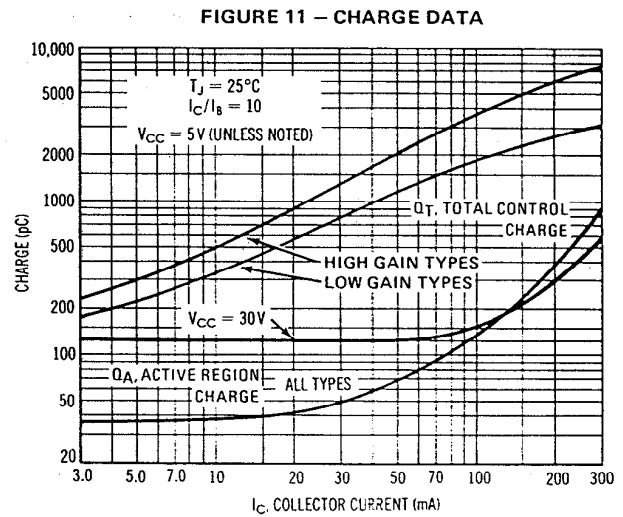
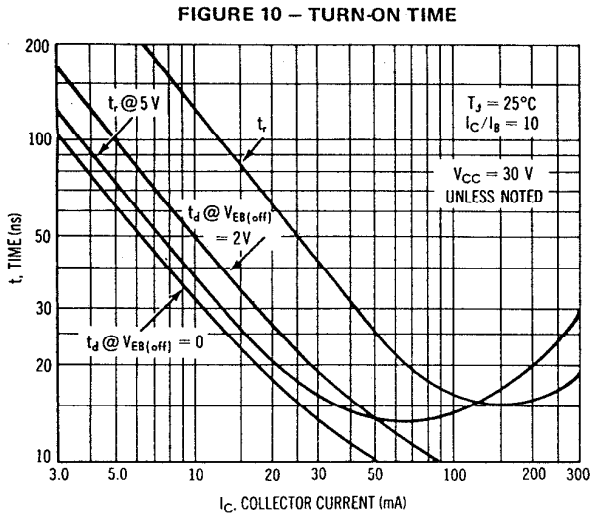
FIGURE 7 – SOURCE RESISTANCE EFFECTS



NPN DATA



SWITCHING TIME CHARACTERISTICS



PNP DATA

FIGURE 13 – DC CURRENT GAIN

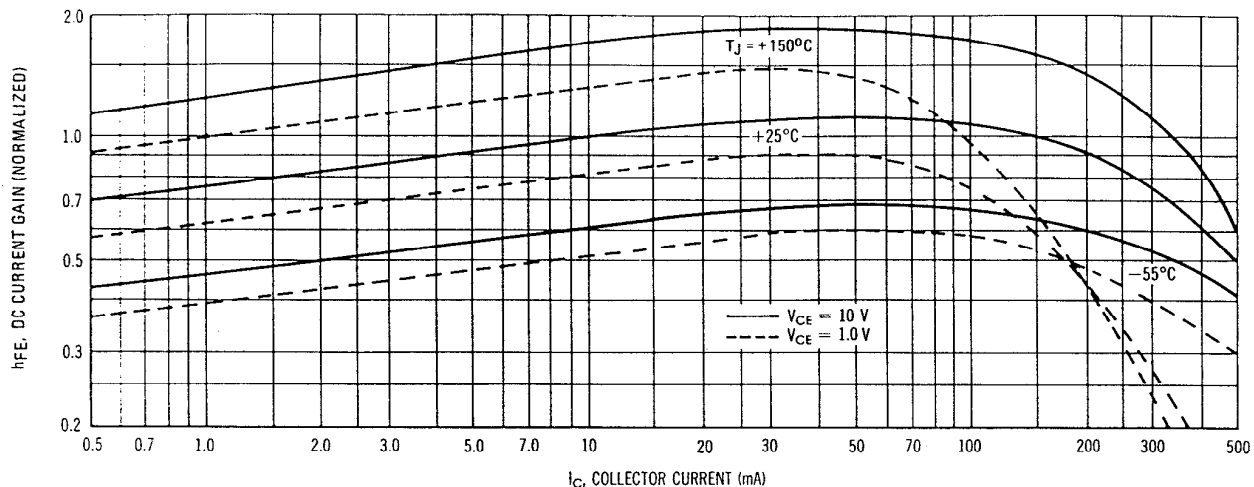


FIGURE 14 – "ON" VOLTAGES

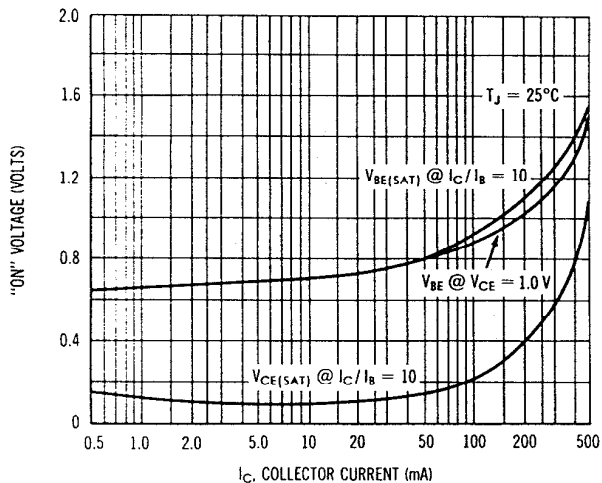
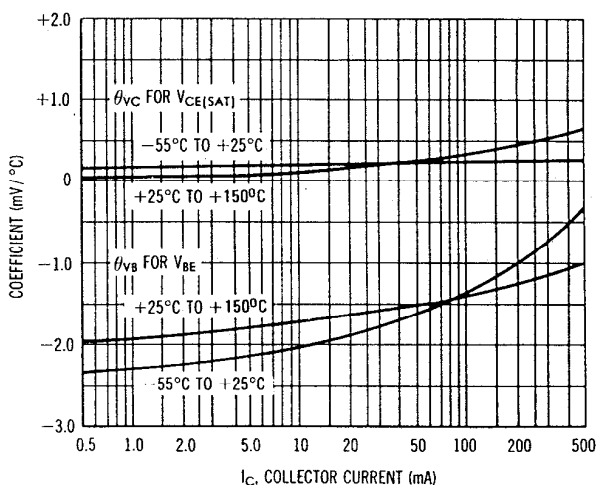


FIGURE 15 – TEMPERATURE COEFFICIENTS



NOISE FIGURE  
(V<sub>CE</sub> = 10 V, T<sub>A</sub> = 25°C)

FIGURE 16 – FREQUENCY EFFECTS

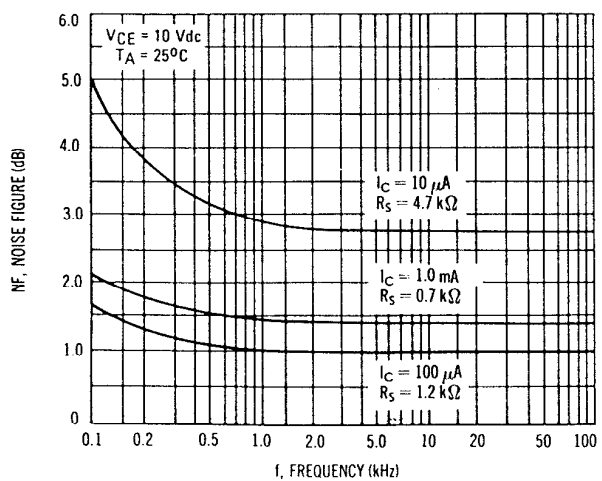
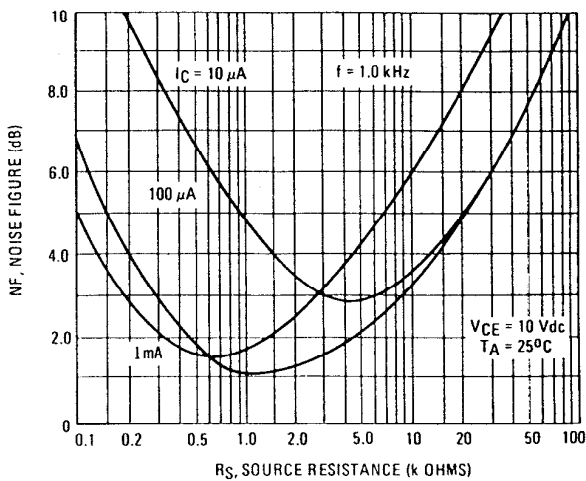


FIGURE 17 – SOURCE RESISTANCE EFFECTS



PNP DATA

FIGURE 18 – CURRENT-GAIN BANDWIDTH PRODUCT

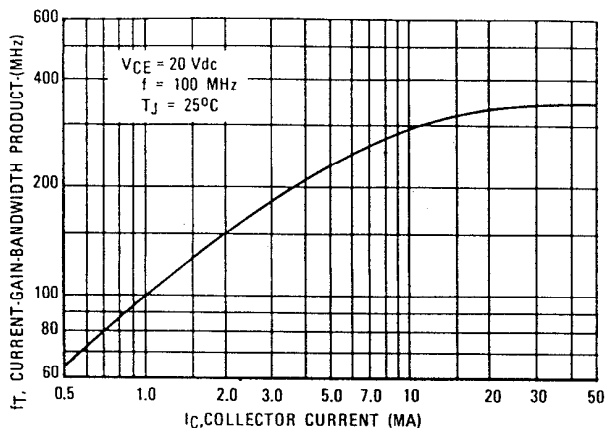


FIGURE 19 – CAPACITANCE

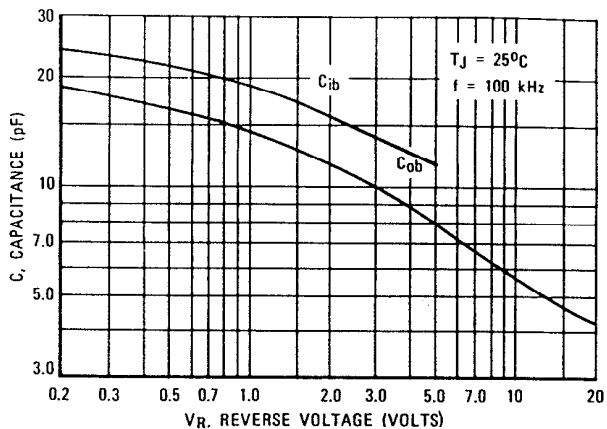


FIGURE 20 – TURN ON TIME

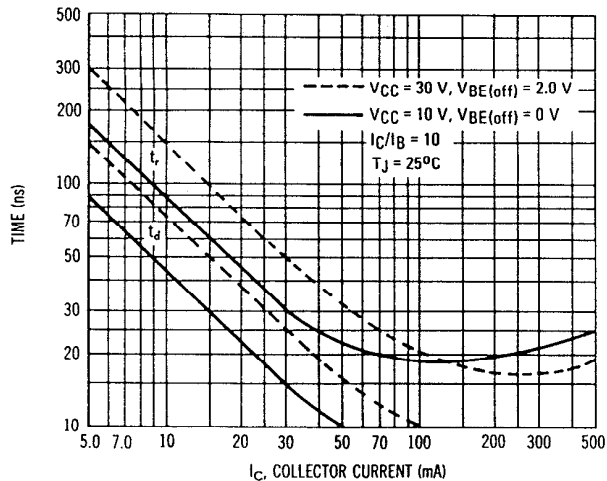


FIGURE 21 – CHARGE DATA

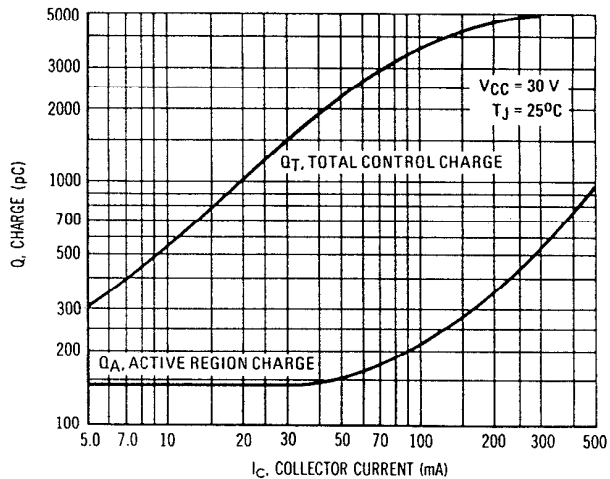


FIGURE 22 – STORAGE TIME

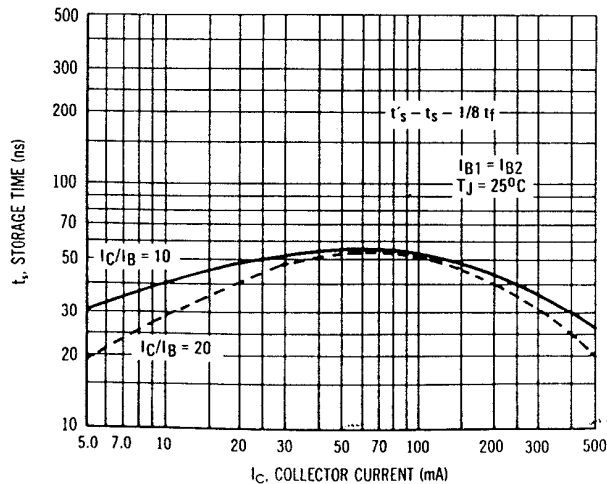
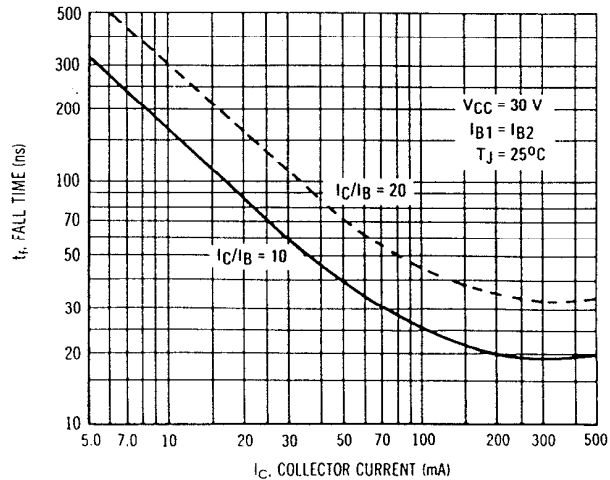


FIGURE 23 – FALL TIME





## Стандарт Электрон Связь

Мы молодая и активно развивающаяся компания в области поставок электронных компонентов. Мы поставляем электронные компоненты отечественного и импортного производства напрямую от производителей и с крупнейших складов мира.

Благодаря сотрудничеству с мировыми поставщиками мы осуществляем комплексные и плановые поставки широчайшего спектра электронных компонентов.

Собственная эффективная логистика и склад в обеспечивает надежную поставку продукции в точно указанные сроки по всей России.

Мы осуществляем техническую поддержку нашим клиентам и предпродажную проверку качества продукции. На все поставляемые продукты мы предоставляем гарантию .

Осуществляем поставки продукции под контролем ВП МО РФ на предприятия военно-промышленного комплекса России , а также работаем в рамках 275 ФЗ с открытием отдельных счетов в уполномоченном банке. Система менеджмента качества компании соответствует требованиям ГОСТ ISO 9001.

Минимальные сроки поставки, гибкие цены, неограниченный ассортимент и индивидуальный подход к клиентам являются основой для выстраивания долгосрочного и эффективного сотрудничества с предприятиями радиоэлектронной промышленности, предприятиями ВПК и научно-исследовательскими институтами России.

С нами вы становитесь еще успешнее!

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