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# FSB50760SFS

## Motion SPM® 5 SuperFET® Series

### Features

- UL Certified No. E209204 (UL1557)
- 600 V  $R_{DS(on)} = 530 \text{ m}\Omega$ (Max) SuperFET MOSFET 3-Phase Inverter with Gate Drivers and Protection
- Built-in Bootstrap Diodes Simplify PCB Layout
- Separate Open-Source Pins from Low-Side MOSFETs for Three-Phase Current-Sensing
- Active-HIGH Interface, Works with 3.3 / 5 V Logic, Schmitt-trigger Input
- Optimized for Low Electromagnetic Interference
- HVIC Temperature-Sensing Built-in for Temperature Monitoring
- HVIC for Gate Driving and Under-Voltage Protection
- Isolation Rating: 1500 Vrms / 1 min.
- Moisture Sensitive Level (MSL) 3
- RoHS Compliant

### Applications

- 3-Phase Inverter Driver for Small Power AC Motor Drives

### Related Source

- [RD-402 - Reference Design for Motion SPM 5 SuperFET Series](#)
- [AN-9082 - Motion SPM5 Series Thermal Performance by Contact Pressure](#)
- [AN-9080 - User's Guide for Motion SPM 5 Series V2](#)

### General Description

The FSB50760SFS is an advanced Motion SPM® 5 module providing a fully-featured, high-performance inverter output stage for AC Induction, BLDC and PMSM motors such as refrigerators, fans and pumps. These modules integrate optimized gate drive of the built-in MOSFETs(SuperFET® technology) to minimize EMI and losses, while also providing multiple on-module protection features including under-voltage lockouts and thermal monitoring. The built-in high-speed HVIC requires only a single supply voltage and translates the incoming logic-level gate inputs to the high-voltage, high-current drive signals required to properly drive the module's internal MOSFETs. Separate open-source MOSFET terminals are available for each phase to support the widest variety of control algorithms.



3D Package Drawing (Click to Activate 3D Content)

### Package Marking & Ordering Information

Device Marking	Device	Package	Reel Size	Packing Type	Quantity
50760SFS	FSB50760SFS	SPM5Q-023	330mm	Tape-Reel	450

## Absolute Maximum Ratings

**Inverter Part** (each MOSFET unless otherwise specified.)

Symbol	Parameter	Conditions	Rating	Unit
$V_{DSS}$	Drain-Source Voltage of Each MOSFET		600	V
$*I_{D\ 25}$	Each MOSFET Drain Current, Continuous	$T_C = 25^\circ\text{C}$	3.6	A
$*I_{D\ 80}$	Each MOSFET Drain Current, Continuous	$T_C = 80^\circ\text{C}$	2.7	A
$*I_{DP}$	Each MOSFET Drain Current, Peak	$T_C = 25^\circ\text{C}$ , $PW < 100\ \mu\text{s}$	9.4	A
$*I_{DRMS}$	Each MOSFET Drain Current, Rms	$T_C = 80^\circ\text{C}$ , $F_{PWM} < 20\ \text{kHz}$	1.9	$A_{rms}$
$*P_D$	Maximum Power Dissipation	$T_C = 25^\circ\text{C}$ , For Each MOSFET	14.5	W

**Control Part** (each HVIC unless otherwise specified.)

Symbol	Parameter	Conditions	Rating	Unit
$V_{CC}$	Control Supply Voltage	Applied Between $V_{CC}$ and COM	20	V
$V_{BS}$	High-side Bias Voltage	Applied Between $V_B$ and $V_S$	20	V
$V_{IN}$	Input Signal Voltage	Applied Between IN and COM	$-0.3 \sim V_{CC} + 0.3$	V

**Bootstrap Diode Part** (each bootstrap diode unless otherwise specified.)

Symbol	Parameter	Conditions	Rating	Unit
$V_{RRMB}$	Maximum Repetitive Reverse Voltage		600	V
$*I_{FB}$	Forward Current	$T_C = 25^\circ\text{C}$	0.5	A
$*I_{FPB}$	Forward Current (Peak)	$T_C = 25^\circ\text{C}$ , Under 1ms Pulse Width	1.5	A

## Thermal Resistance

Symbol	Parameter	Conditions	Rating	Unit
$R_{\theta JC}$	Junction to Case Thermal Resistance	Each MOSFET under Inverter Operating Condition (1st Note 1)	8.6	$^\circ\text{C}/\text{W}$

## Total System

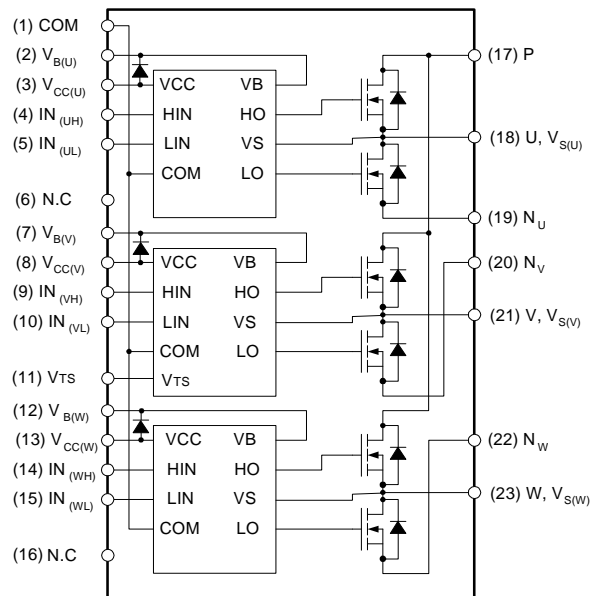
Symbol	Parameter	Conditions	Rating	Unit
$T_J$	Operating Junction Temperature		$-40 \sim 150$	$^\circ\text{C}$
$T_{STG}$	Storage Temperature		$-40 \sim 125$	$^\circ\text{C}$
$V_{ISO}$	Isolation Voltage	60 Hz, Sinusoidal, 1 Minute, Connect Pins to Heat Sink Plate	1500	$V_{rms}$

### 1st Notes:

- For the measurement point of case temperature  $T_C$ , please refer to Figure 4.
- Marking " \* " is calculation value or design factor.

## Pin descriptions

Pin Number	Pin Name	Pin Description
1	COM	IC Common Supply Ground
2	$V_{B(U)}$	Bias Voltage for U-Phase High-Side MOSFET Driving
3	$V_{CC(U)}$	Bias Voltage for U-Phase IC and Low-Side MOSFET Driving
4	$IN_{(UH)}$	Signal Input for U-Phase High-Side
5	$IN_{(UL)}$	Signal Input for U-Phase Low-Side
6	N.C	No Connection
7	$V_{B(V)}$	Bias Voltage for V-Phase High Side MOSFET Driving
8	$V_{CC(V)}$	Bias Voltage for V-Phase IC and Low Side MOSFET Driving
9	$IN_{(VH)}$	Signal Input for V-Phase High-Side
10	$IN_{(VL)}$	Signal Input for V-Phase Low-Side
11	$V_{TS}$	Output for HVIC Temperature Sensing
12	$V_{B(W)}$	Bias Voltage for W-Phase High-Side MOSFET Driving
13	$V_{CC(W)}$	Bias Voltage for W-Phase IC and Low-Side MOSFET Driving
14	$IN_{(WH)}$	Signal Input for W-Phase High-Side
15	$IN_{(WL)}$	Signal Input for W-Phase Low-Side
16	N.C	No Connection
17	P	Positive DC-Link Input
18	U, $V_{S(U)}$	Output for U-Phase & Bias Voltage Ground for High-Side MOSFET Driving
19	$N_U$	Negative DC-Link Input for U-Phase
20	$N_V$	Negative DC-Link Input for V-Phase
21	V, $V_{S(V)}$	Output for V-Phase & Bias Voltage Ground for High-Side MOSFET Driving
22	$N_W$	Negative DC-Link Input for W-Phase
23	W, $V_{S(W)}$	Output for W Phase & Bias Voltage Ground for High-Side MOSFET Driving



**Figure 1. Pin Configuration and Internal Block Diagram (Bottom View)**

### 1st Notes:

- Source terminal of each low-side MOSFET is not connected to supply ground or bias voltage ground inside Motion SPM® 5 product. External connections should be made as indicated in Figure 3.

**Electrical Characteristics** ( $T_J = 25^\circ\text{C}$ ,  $V_{CC} = V_{BS} = 15\text{ V}$  unless otherwise specified.)**Inverter Part** (each MOSFET unless otherwise specified.)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$BV_{DSS}$	Drain - Source Breakdown Voltage	$V_{IN} = 0\text{ V}$ , $I_D = 1\text{ mA}$ (2nd Note 1)	600	-	-	V
$I_{DSS}$	Zero Gate Voltage Drain Current	$V_{IN} = 0\text{ V}$ , $V_{DS} = 600\text{ V}$	-	-	1	mA
$R_{DS(on)}$	Static Drain - Source Turn-On Resistance	$V_{CC} = V_{BS} = 15\text{ V}$ , $V_{IN} = 5\text{ V}$ , $I_D = 2\text{ A}$	-	460	530	m $\Omega$
$V_{SD}$	Drain - Source Diode Forward Voltage	$V_{CC} = V_{BS} = 15\text{ V}$ , $V_{IN} = 0\text{ V}$ , $I_D = -2\text{ A}$	-	-	1.1	V
$t_{ON}$	Switching Times	$V_{PN} = 300\text{ V}$ , $V_{CC} = V_{BS} = 15\text{ V}$ , $I_D = 2\text{ A}$ $V_{IN} = 0\text{ V} \leftrightarrow 5\text{ V}$ , Inductive Load $L = 3\text{ mH}$ High- and Low-Side MOSFET Switching (2nd Note 2)	-	1200	-	ns
$t_{OFF}$			-	970	-	ns
$t_{rr}$			-	160	-	ns
$E_{ON}$			-	120	-	$\mu\text{J}$
$E_{OFF}$			-	10	-	$\mu\text{J}$
RBSOA	Reverse Bias Safe Operating Area	$V_{PN} = 400\text{ V}$ , $V_{CC} = V_{BS} = 15\text{ V}$ , $I_D = I_{DP}$ , $V_{DS} = BV_{DSS}$ , $T_J = 150^\circ\text{C}$ High- and Low-Side MOSFET Switching (2nd Note 3)	Full Square			

**Control Part** (each HVIC unless otherwise specified.)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$I_{QCC}$	Quiescent $V_{CC}$ Current	$V_{CC} = 15\text{ V}$ , $V_{IN} = 0\text{ V}$ Applied Between $V_{CC}$ and COM	-	-	200	$\mu\text{A}$
$I_{QBS}$	Quiescent $V_{BS}$ Current	$V_{BS} = 15\text{ V}$ , $V_{IN} = 0\text{ V}$ Applied Between $V_{B(U)} - U$ , $V_{B(V)} - V$ , $V_{B(W)} - W$	-	-	100	$\mu\text{A}$
$UV_{CCD}$	Low-Side Under-Voltage Protection (Figure 8)	$V_{CC}$ Under-Voltage Protection Detection Level	7.4	8.0	9.4	V
$UV_{CCR}$		$V_{CC}$ Under-Voltage Protection Reset Level	8.0	8.9	9.8	V
$UV_{BSD}$	High-Side Under-Voltage Protection (Figure 9)	$V_{BS}$ Under-Voltage Protection Detection Level	7.4	8.0	9.4	V
$UV_{BSR}$		$V_{BS}$ Under-Voltage Protection Reset Level	8.0	8.9	9.8	V
$V_{TS}$	HVIC Temperature Sensing Voltage Output	$V_{CC} = 15\text{ V}$ , $T_{HVIC} = 25^\circ\text{C}$ (2nd Note 4)	600	790	980	mV
$V_{IH}$	ON Threshold Voltage	Logic HIGH Level	-	-	2.9	V
$V_{IL}$	OFF Threshold Voltage	Logic LOW Level	0.8	-	-	V

**Bootstrap Diode Part** (each bootstrap diode unless otherwise specified.)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{FB}$	Forward Voltage	$I_F = 0.1\text{ A}$ , $T_C = 25^\circ\text{C}$ (2nd Note 5)	-	2.5	-	V
$t_{rrB}$	Reverse Recovery Time	$I_F = 0.1\text{ A}$ , $T_C = 25^\circ\text{C}$	-	80	-	ns

**2nd Notes:**

- $BV_{DSS}$  is the absolute maximum voltage rating between drain and source terminal of each MOSFET inside Motion SPM® 5 product.  $V_{PN}$  should be sufficiently less than this value considering the effect of the stray inductance so that  $V_{PN}$  should not exceed  $BV_{DSS}$  in any case.
- $t_{ON}$  and  $t_{OFF}$  include the propagation delay of the internal drive IC. Listed values are measured at the laboratory test condition, and they can be different according to the field applications due to the effect of different printed circuit boards and wirings. Please see Figure 6 for the switching time definition with the switching test circuit of Figure 7.
- The peak current and voltage of each MOSFET during the switching operation should be included in the Safe Operating Area (SOA). Please see Figure 7 for the RBSOA test circuit that is same as the switching test circuit.
- $V_{IS}$  is only for sensing-temperature of module and cannot shutdown MOSFETs automatically.
- Built-in bootstrap diode includes around 15  $\Omega$  resistance characteristic. Please refer to Figure 2.

## Recommended Operating Condition

Symbol	Parameter	Conditions	Min.	Typ.	Max.	Unit
$V_{PN}$	Supply Voltage	Applied Between P and N	-	300	450	V
$V_{CC}$	Control Supply Voltage	Applied Between $V_{CC}$ and COM	13.5	15.0	16.5	V
$V_{BS}$	High-Side Bias Voltage	Applied Between $V_B$ and $V_S$	13.5	15.0	16.5	V
$V_{IN(ON)}$	Input ON Threshold Voltage	Applied Between IN and COM	3.0	-	$V_{CC}$	V
$V_{IN(OFF)}$	Input OFF Threshold Voltage		0	-	0.6	V
$t_{dead}$	Blanking Time for Preventing Arm-Short	$V_{CC} = V_{BS} = 13.5 \sim 16.5 \text{ V}$ , $T_J \leq 150^\circ\text{C}$	1.0	-	-	$\mu\text{s}$
$f_{PWM}$	PWM Switching Frequency	$T_J \leq 150^\circ\text{C}$	-	20	-	kHz



**Figure 2. Built-in Bootstrap Diode Characteristics (Typical)**



Figure 3. Recommended MCU Interface and Bootstrap Circuit with Parameters

3rd Notes:

- Parameters for bootstrap circuit elements are dependent on PWM algorithm. For 15 kHz of switching frequency, typical example of parameters is shown above.
- RC-coupling (R<sub>5</sub> and C<sub>5</sub>) and C<sub>4</sub> at each input of Motion SPM 5 product and MCU (Indicated as Dotted Lines) may be used to prevent improper signal due to surge-noise.
- Bold lines should be short and thick in PCB pattern to have small stray inductance of circuit, which results in the reduction of surge-voltage. Bypass capacitors such as C<sub>1</sub>, C<sub>2</sub> and C<sub>3</sub> should have good high-frequency characteristics to absorb high-frequency ripple-current.

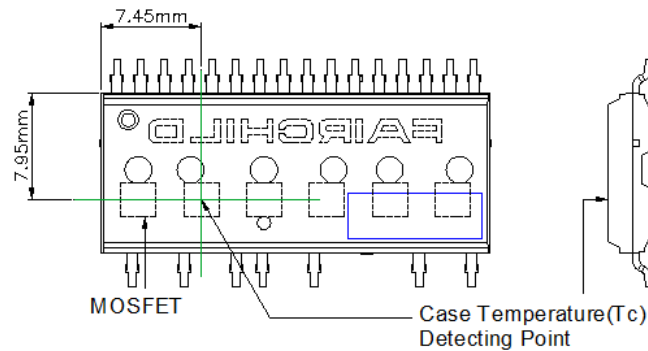


Figure 4. Case Temperature Measurement

3rd Notes:

- Attach the thermocouple on top of the heat-sink of SPM 5 package (between SPM 5 package and heatsink if applied) to get the correct temperature measurement.

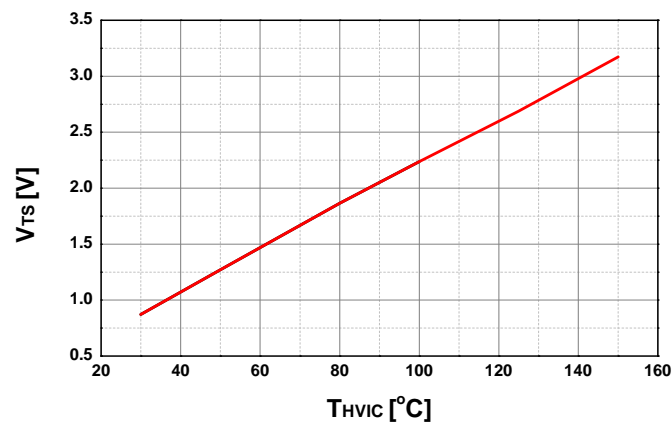


Figure 5. Temperature Profile of V<sub>TS</sub> (Typical)

The diagram illustrates the internal structure of the Motion SPM® 5 product, which is a single-leg inverter. It features a central semiconductor chip with several pins and internal components. The pins are labeled: VCC, HIN, LIN, COM, VTS, VB, HO, VS, and LO. The internal components include a MOSFET driver stage (HIN, LIN, COM, VTS) and a power MOSFET stage (VB, HO, VS, LO). The power MOSFET stage is connected to a load inductor (L) and a DC source (V<sub>DC</sub>). The output current is labeled I<sub>D</sub>. The diagram also shows a gate driver circuit (V<sub>CC</sub>, HIN, LIN, COM, VTS) and a bootstrap capacitor (C<sub>BS</sub>) connected to the gate of the power MOSFET. The output voltage across the inductor is labeled V<sub>DS</sub>.

The diagram illustrates the UV Protection Status logic. The **Input Signal** is a periodic square wave. The **UV Protection Status** transitions from **RESET** to **DETECTION** when the **Low-side Supply,  $V_{CC}$**  drops below the  **$UV_{CCD}$**  threshold. It returns to **RESET** when the supply rises above the  **$UV_{CCR}$**  threshold. The **MOSFET Current** is shown as pulses that occur only when the status is in **RESET** and the input signal is high. The current stops during the **DETECTION** period.

The diagram illustrates the UV Protection Status logic. The **Input Signal** is a periodic square wave. The **UV Protection Status** is initially in the **RESET** state. When the **High-side Supply,  $V_{BS}$**  falls below the  **$UV_{BSD}$**  threshold, the status transitions to **DETECTION**. When  $V_{BS}$  rises above the  **$UV_{BSR}$**  threshold, the status returns to **RESET**. The **MOSFET Current** shows pulses during the **RESET** state and no current during the **DETECTION** state.

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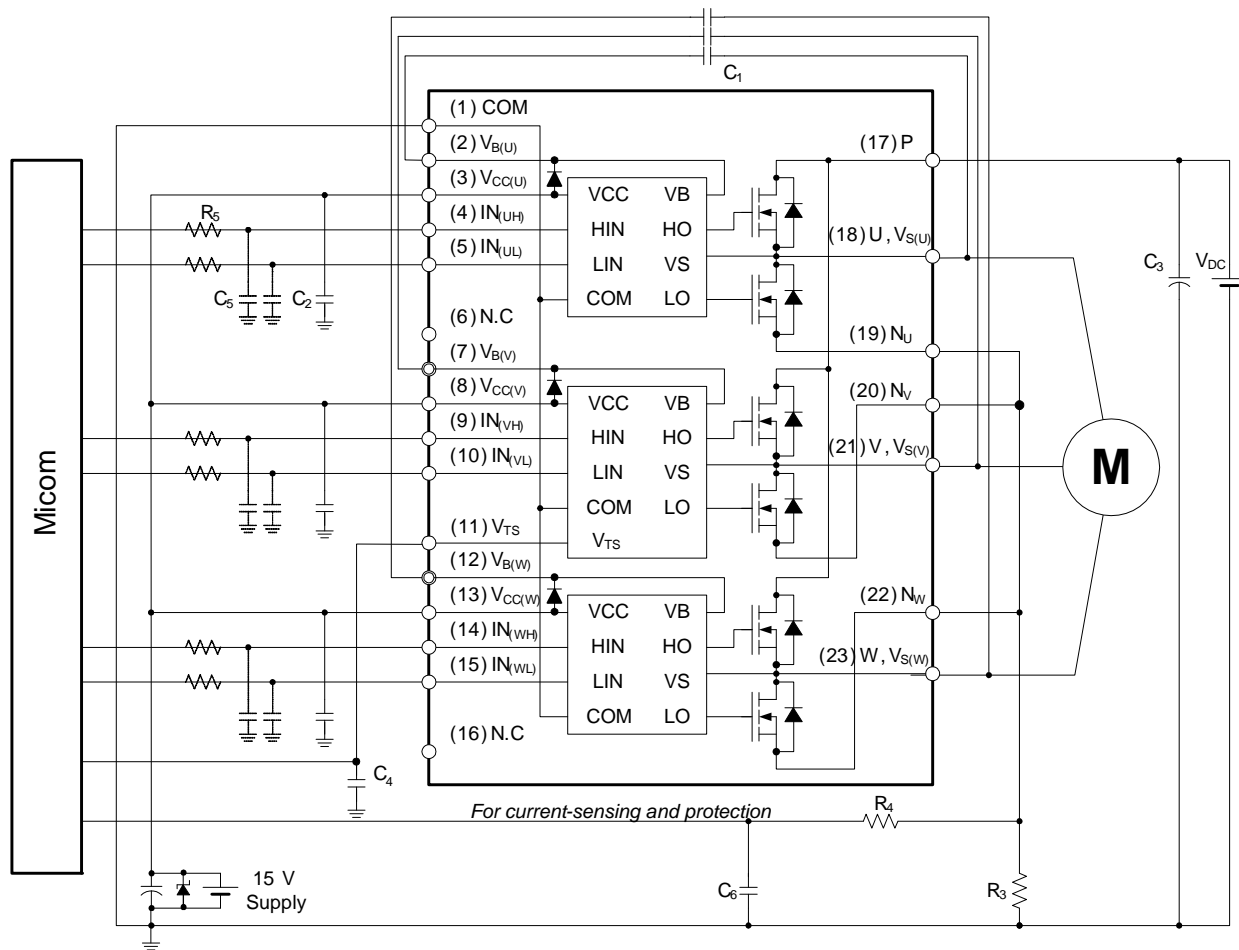
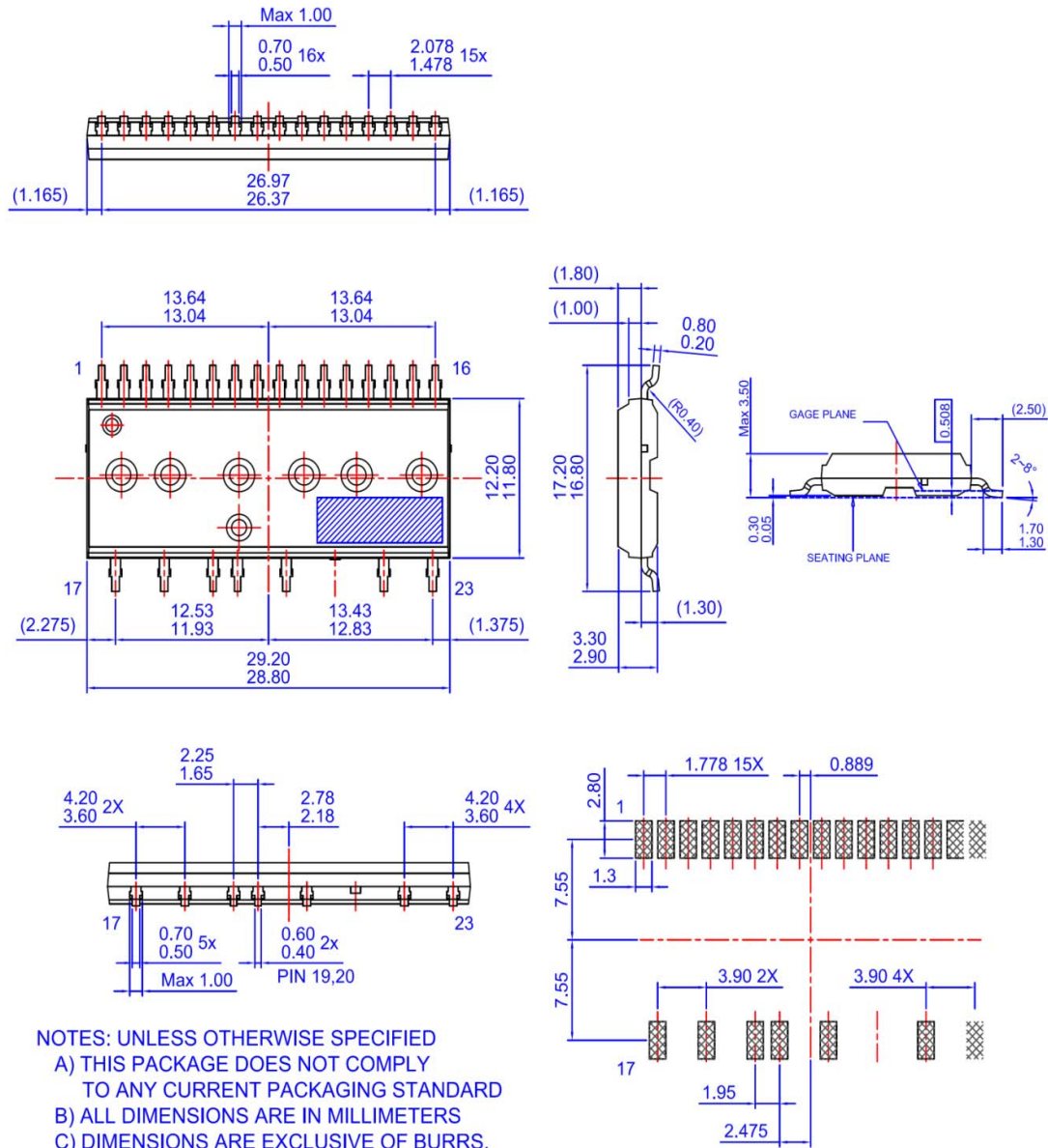


Figure 10. Example of Application Circuit

4th Notes:

1. About pin position, refer to Figure 1.
2. RC-coupling ( $R_5$  and  $C_5$ ,  $R_4$  and  $C_6$ ) and  $C_4$  at each input of Motion SPM® 5 product and MCU are useful to prevent improper input signal caused by surge-noise.
3. The voltage-drop across  $R_3$  affects the low-side switching performance and the bootstrap characteristics since it is placed between COM and the source terminal of the low-side MOSFET. For this reason, the voltage-drop across  $R_3$  should be less than 1 V in the steady-state.
4. Ground-wires and output terminals, should be thick and short in order to avoid surge-voltage and malfunction of HVIC.
5. All the filter capacitors should be connected close to Motion SPM 5 product, and they should have good characteristics for rejecting high-frequency ripple current.

# Detailed Package Outline Drawings (FSB50760SFS)



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TinyBoost®  
TinyBuck®  
TinyCalc™  
TinyLogic®  
TINYOPTO™  
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No Identification Needed	Full Production	Datasheet contains final specifications. Fairchild Semiconductor reserves the right to make changes at any time without notice to improve the design.
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