

6.0 Gbps SATA PC Board and Cable Equalizer

 Check for Samples: [SN75LVCP600DRF](#)

FEATURES

- Multirate Operation, 1.5/3.0/6.0 Gbps
- Suitable to Receive 6.0 Gbps Data Over up to 40 Inches (1.0 meter) of FR4 PC Board
- Compensates up to 14dB Loss on the Receive Side and 1.2dB Loss on the Transmit Side at 3 GHz
- Integrated Output Squelch
- Auto Low Power Feature Lowers Power by >80% When Link is in Partial or Slumber Mode
 - <100 mW (Active Mode, typ)
 - <11 mW (Auto Low Power Mode, typ)

- Single 3.3 V Supply
- High Protection Against ESD Transient
 - HBM: 6,000 V
 - CDM: 1,500 V
- Ultra Small Footprint, 2x2 QFN Package

APPLICATIONS

- Notebooks, Desktops, Docking Stations, Servers, Workstations

DESCRIPTION

The SN75LVCP600 is a versatile single channel, SATA signal conditioner supporting data rates up to 6.0Gbps. The device complies with SATA physical link 2m and 3i specifications. The SN75LVCP600 operates from a single 3.3V supply and has 100-Ω line termination with self-biasing feature, making the device suitable for AC coupling. The inputs incorporate an out-of-band (OOB) detector, which automatically squelches the output when input differential voltage falls below threshold while maintaining a stable common-mode voltage compliant to SATA link. The device is also designed to handle spread spectrum clocking (SSC) transmission per SATA spec.

The SN75LVCP600 handles interconnect losses at its input with selectable equalization settings that can be programmed to match the loss in the channel. For data rates of 3Gbps and lower the LVCP600 equalizes signals for a span of up to 50 inches of FR4 board material. For data rates of 6Gbps the device compensates up to 40in of FR4 material. The equalization level is controlled by the setting of the signal control pin EQ.

Two de-emphasis levels can be selected on the transmit side to provide 0 or 1.2dB of additional high-frequency loss compensation at the output.

The device is hot-plug capable⁽¹⁾ preventing device damage under device *hot*-insertion such as async signal plug/removal, unpowered plug/removal, powered plug/removal, or surprise plug/removal.

(1) Requires use of AC coupling capacitors at differential inputs and outputs.

ORDERING INFORMATION⁽¹⁾

| PART NUMBER | PART MARKING | PACKAGE |
|-----------------|--------------|------------------------|
| SN75LVCP600DRFR | 600 | 8-Pin DRF reel (large) |
| SN75LVCP600DRFT | 600 | 8-Pin DRF reel (small) |

(1) For the most current package and ordering information, see the Package Option Addendum at the end of this document, or see the TI web site at www.ti.com.



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

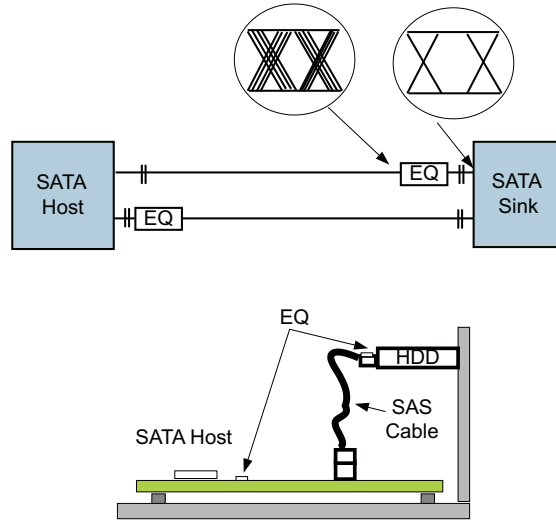


Figure 1. Typical Application

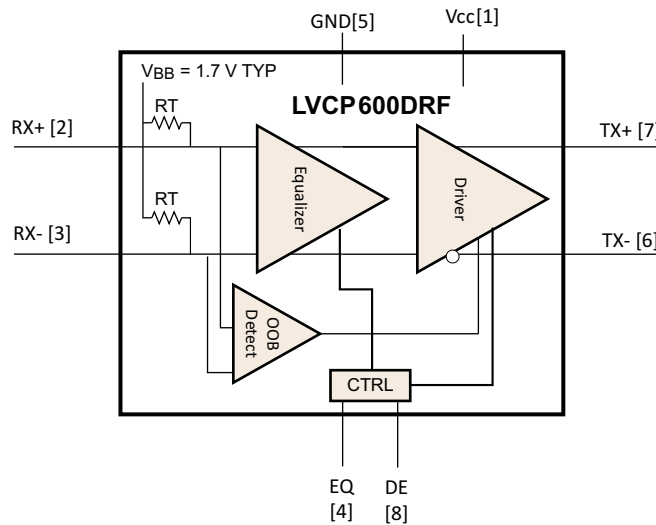
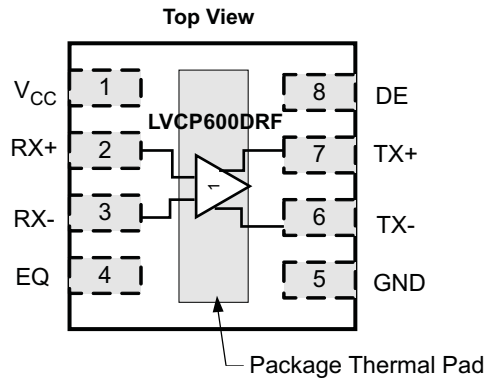


Figure 2. Data Flow Block Diagram

PIN ASSIGNMENTS

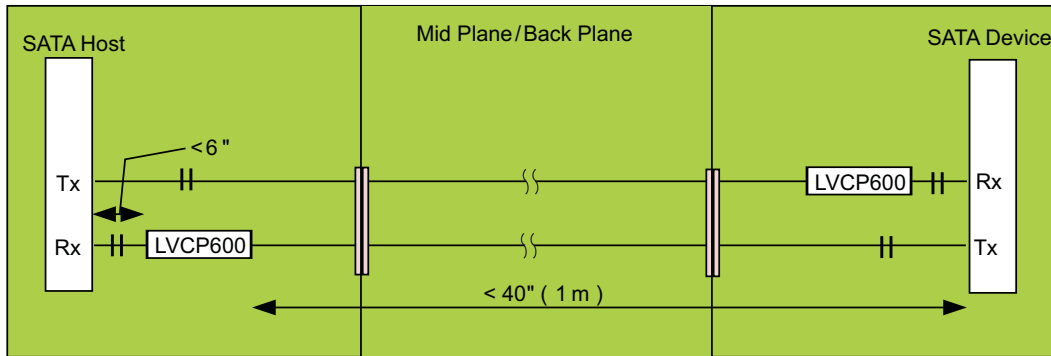


PIN FUNCTIONS

| PIN | | I/O TYPE | DESCRIPTION |
|------------------------------------|-----------------|-----------|---|
| NO. | NAME | | |
| HIGH SPEED DIFFERENTIAL I/O | | | |
| 2 | RX+ | I, VML | Non-inverting and inverting VML differential inputs. These pins are tied to an internal voltage bias by dual termination resistor circuit. |
| 3 | RX- | I, VML | |
| 7 | TX+ | I, VML | Non-inverting and inverting VML differential outputs. These pins are tied to an internal voltage bias by dual termination resistor circuit. |
| 6 | TX- | I, VML | |
| CONTROL PINS | | | |
| 4 | EQ | I, LVCMOS | Selects equalization settings per Table 1 . Internally tied to GND. |
| 8 | DE | I, LVCMOS | Selects de-emphasis settings per Table 1 . Internally tied to GND. |
| POWER | | | |
| 1 | V _{CC} | Power | Positive supply should be 3.3 V ±10% |
| 5 | GND | Power | Supply ground |

Table 1. EQ and DE Settings

| EQ | EQUALIZATION dB (at 6 Gbps) | DE | DE-EMPHASIS dB (at 6 Gbps) |
|----------------------|--------------------------------|----------------------|-------------------------------|
| 0 (<i>default</i>) | 7 | 0 (<i>default</i>) | 0 |
| 1 | 14 | 1 | -1.2 |

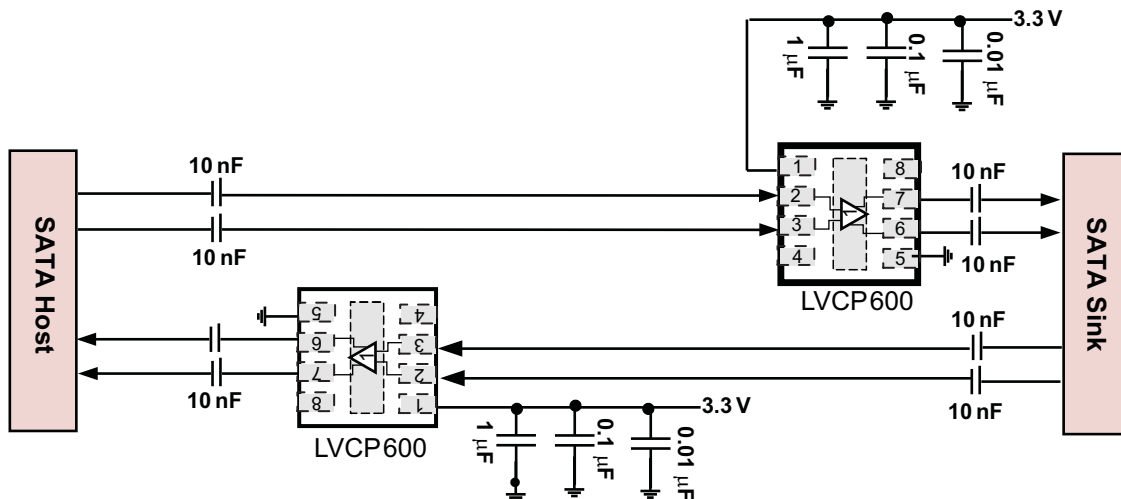


Note:

Trace lengths are suggested values based on TI HSPICE simulations (done over programmable limits of input EQ) to meet SATA loss and jitter spec.

Actual trace length supported by the LVCP600 may be more or less than suggested values and will depend on board layout, trace widths and number of connectors used in the high speed signal path.

Figure 3. Trace Length Example



Note:

- 1) Place supply caps close to device pin
- 2) EQ and DE selection at 7 dB and 0dB respectively
- 3) Actual EQ/DE settings will depend on device placement relative to host and SATA connector

Figure 4. Typical Device Implementation

The LVCP600 allows user to take the guess work of using a signal conditioning device in a SATA link. With the LVCP600 user has the option to use or remove the device based on signal conditioning needs. See [Figure 5](#).

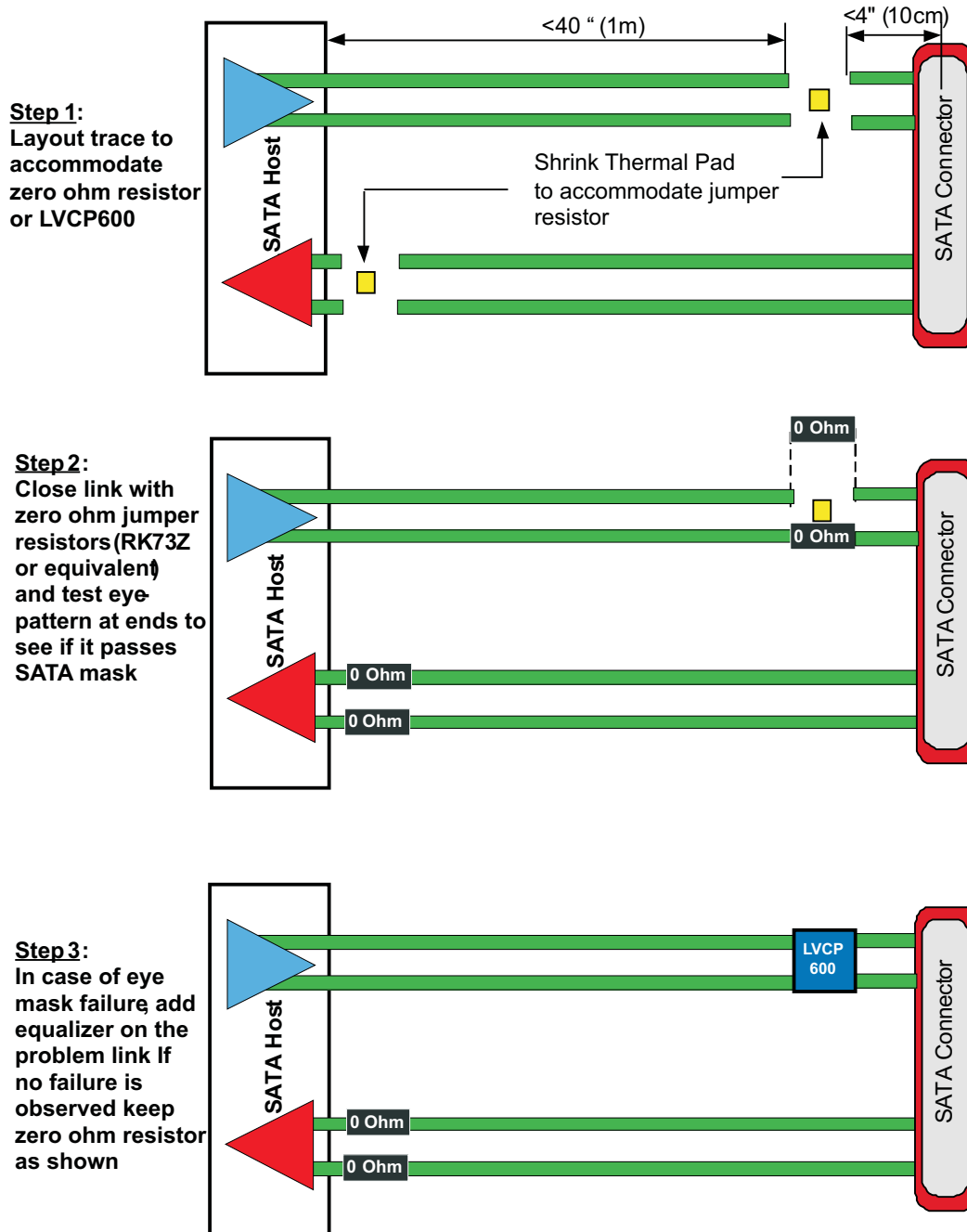


Figure 5. Implementation Guideline

OPERATION DESCRIPTION

INPUT EQUALIZATION

The SN75LVCP600 supports programmable equalization in its front stage; the equalization settings are shown in Table 1. The input equalizer is designed to recover a signal even when no eye is present at the receiver and will affectively support FR4 trace at the input anywhere from 4" to 40" at SATA 6G speed.

AUTO LOW POWER (ALP) MODE (see Figure 10)

Auto Low Power (ALP) Mode is triggered when a channel is in the electrical idle state for >10 μ s.

- The device enters and exits Low Power Mode by actively monitoring input signal (VID_{pp}) level. When the input signal is in the electrical idle state, i.e., $VID_{pp} < 50$ mV and stays in this state for >10 μ s, the device automatically enters the low power state. In this state the output is driven to VCM and the device selectively shuts off internal circuitry to lower power by >90% of its normal operating power. While in ALP mode the device continues actively to monitor input signal levels. When the input signal exceeds the SATA OOB upper threshold level, the device reverts to the active state. Exit time from Auto Low Power Mode is <50 ns (max).

OUT-OF-BAND (OOB) SUPPORT

The squelch detector circuit within the device enables full detection of OOB signaling as specified in the SATA spec. When differential signal amplitude at the receiver input is 50 mV_{pp} or less, the output is squelched. Differential signal amplitude of 90 mV_{pp} or more is detected as an activity and therefore passed to the output indicating activity. Squelch circuit ON/OFF time is 5 ns max. While in squelch mode outputs are held to VCM.

DEVICE POWER

The SN75LVCP600 is designed to operate from a single 3.3 V supply. Always practice proper power supply sequencing procedure. Apply V_{CC} first before any input signals are applied to the device. The power down sequence is in reverse order.

ABSOLUTE MAXIMUM RATINGS

over operating free-air temperature range (unless otherwise noted)⁽¹⁾

| | | VALUE | UNIT |
|-------------------------------------|-------------------------------------|------------------------------|------|
| Supply voltage range ⁽²⁾ | V_{CC} | -0.5 to 4 | V |
| Voltage range | Differential I/O | -0.5 to 4 | V |
| | Control I/O | -0.5 to $V_{CC} + 0.5$ | V |
| Electrostatic discharge | Human body model ⁽³⁾ | ± 6000 | V |
| | Charged-device model ⁽⁴⁾ | ± 1500 | V |
| | Machine model ⁽⁵⁾ | ± 200 | V |
| Continuous power dissipation | | See Dissipation Rating Table | |

- Stresses beyond those listed under absolute maximum ratings may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any conditions beyond those indicated under recommended operating conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- All voltage values, except differential voltages, are with respect to network ground terminal.
- Tested in accordance with JEDEC Standard 22, Test Method A114-B.
- Tested in accordance with JEDEC Standard 22, Test Method C101-A.
- Tested in accordance with JEDEC Standard 22, Test Method A115-A.

THERMAL INFORMATION

| THERMAL METRIC ⁽¹⁾ | | SN75LVCP600 | UNITS |
|-------------------------------|--|--------------|-------|
| | | DRF (8) PINS | |
| θ_{JA} | Junction-to-ambient thermal resistance | 97.8 | °C/W |
| θ_{JCTop} | Junction-to-case (top) thermal resistance | 81.9 | |
| θ_{JB} | Junction-to-board thermal resistance | 65.6 | |
| ψ_{JT} | Junction-to-top characterization parameter | 1.3 | |
| ψ_{JB} | Junction-to-board characterization parameter | 65.6 | |
| θ_{JCbott} | Junction-to-case (bottom) thermal resistance | 19.1 | |

- For more information about traditional and new thermal metrics, see the *IC Package Thermal Metrics* application report, [SPRA953](#).

RECOMMENDED OPERATING CONDITIONS

typical values for all parameters are at $V_{CC} = 3.3\text{ V}$ and $T_A = 25^\circ\text{C}$; all temperature limits are specified by design

| PARAMETER | | TEST CONDITIONS | MIN | TYP | MAX | UNITS |
|----------------|--------------------------------|-----------------|-----|-----|-----|------------------|
| V_{CC} | Supply voltage | | 3 | 3.3 | 3.6 | V |
| $C_{COUPLING}$ | Coupling capacitor | | | 12 | | nF |
| T_A | Operating free-air temperature | | 0 | | 85 | $^\circ\text{C}$ |

ELECTRICAL CHARACTERISTICS

over recommended operating conditions (unless otherwise noted)

| PARAMETER | | TEST CONDITIONS | MIN | TYP | MAX | UNITS |
|--------------------------|------------------------------------|---|-----|-----|------|------------------|
| DEVICE PARAMETERS | | | | | | |
| I_{CCMax} | Active mode supply current | EQ/DE = NC, K28.5 pattern at 6 Gbps, $V_{ID} = 700\text{ mV}_{pp}$ | | 29 | 40 | mA |
| I_{CCPS} | Auto power save mode I_{CC} | When auto low power conditions are met | | 3.3 | 5.9 | mA |
| | Maximum data rate | | | | 6.0 | Gbps |
| t_{PDelay} | Propagation delay | Measured using K28.5 pattern, See Figure 8 | | 275 | 350 | ps |
| AutoLP _{ENTRY} | Auto low power entry time | Electrical idle at input, See Figure 10 | | 11 | | μs |
| AutoLP _{EXIT} | Auto low power exit time | After first signal activity, See Figure 10 | | 33 | 50 | ns |
| OOB | | | | | | |
| V_{OOB} | Input OOB threshold | $F = 750\text{ MHz}$ | 50 | 70 | 90 | mV_{pp} |
| $D_{VdiffOOB}$ | OOB differential delta | | | | 25 | mV |
| D_{VCMOOB} | OOB common-mode delta | | | | 50 | mV |
| t_{OOB1} | OOB mode enter | See Figure 9 | | 1 | 5 | ns |
| t_{OOB2} | OOB mode exit | See Figure 9 | | 1 | 5 | ns |
| CONTROL LOGIC | | | | | | |
| V_{IH} | High-level input voltage | For all control pins | 1.4 | | | V |
| V_{IL} | Low-level input voltage | | | | 0.5 | V |
| V_{INHYS} | Input hysteresis | | | 115 | | mV |
| I_{IH} | High-level input current | $V_{IH} = V_{CC}$ (DE/EQ) | | | 20 | μA |
| I_{IL} | Low-level input current | $V_{IL} = 0\text{V}$ (DE/EQ) | | | 10 | μA |
| RECEIVER AC/DC | | | | | | |
| Z_{DIFFRX} | Differential input impedance | | 85 | 100 | 115 | Ω |
| Z_{SERX} | Single-ended input impedance | | 40 | | | Ω |
| V_{CMRX} | Common-mode voltage | | | 1.7 | | V |
| RL_{DIFFRX} | Differential mode return loss (RL) | $f = 150\text{ MHz} - 300\text{ MHz}$ | 18 | 26 | | dB |
| | | $f = 300\text{ MHz} - 600\text{ MHz}$ | 14 | 22 | | |
| | | $f = 600\text{ MHz} - 1.2\text{ GHz}$ | 10 | 17 | | |
| | | $f = 1.2\text{ GHz} - 2.4\text{ GHz}$ | 8 | 12 | | |
| | | $f = 2.4\text{ GHz} - 3.0\text{ GHz}$ | 3 | 11 | | |
| $RX_{DIFFRLSlope}$ | Differential mode RL slope | $f = 300\text{ MHz} - 6.0\text{ GHz}$ (see Figure 6) | | -13 | | dB/dec |
| RL_{CMRX} | Common-mode return loss | $f = 150\text{ MHz} - 300\text{ MHz}$ | 5 | 9.4 | | dB |
| | | $f = 300\text{ MHz} - 600\text{ MHz}$ | 5 | 17 | | |
| | | $f = 600\text{ MHz} - 1.2\text{ GHz}$ | 2 | 18 | | |
| | | $f = 1.2\text{ GHz} - 2.4\text{ GHz}$ | 1 | 9.9 | | |
| | | $f = 2.4\text{ GHz} - 3.0\text{ GHz}$ | 1 | 8.6 | | |
| V_{diffRX} | Differential input voltage PP | $f = 1.5\text{ GHz}$ and 3.0 GHz | 120 | | 1600 | mV/ppd |

ELECTRICAL CHARACTERISTICS (continued)

over recommended operating conditions (unless otherwise noted)

| PARAMETER | | TEST CONDITIONS | MIN | TYP | MAX | UNITS |
|---------------------------|-----------------------------------|---|------|------|-----|------------|
| IB _{RX} | Impedance balance | f = 150 MHz–300 MHz | 30 | 41 | | dB |
| | | f = 300 MHz–600 MHz | 30 | 41 | | |
| | | f = 600 MHz–1.2 GHz | 20 | 34 | | |
| | | f = 1.2 GHz–2.4 GHz | 10 | 24 | | |
| | | f = 2.4 GHz–3.0 GHz | 10 | 26 | | |
| | | f = 3.0 GHz–5.0 GHz | 4 | 18 | | |
| | | f = 5.0 GHz–6.5 GHz | 4 | 18 | | |
| T _{20-80RX} | Rise/fall time | Rise times and fall times measured between 20% and 80% of the signal. SATA 6 Gbps speed measured 1" from device pin | 62 | | 75 | ps |
| T _{skewRX} | Differential skew | Difference between the single-ended mid-point of the RX+ signal rising/falling edge, and the single-ended mid-point of the RX– signal falling/rising edge | | | 30 | ps |
| TRANSMITTER AC/DC | | | | | | |
| Z _{diffTX} | Pair differential impedance | | 85 | 100 | 122 | Ω |
| Z _{SETX} | Single-ended input impedance | | 40 | | | Ω |
| V _{Txtrans} | Sequencing transient voltage | Transient voltages on the serial data bus during power sequencing (lab load) | –1.2 | 0.3 | 1.2 | V |
| RL _{DiffTX} | Differential mode return loss | f = 150 MHz–300 MHz | 13 | 22 | | dB |
| | | f = 300 MHz–600 MHz | 8 | 21 | | |
| | | f = 600 MHz–1.2 GHz | 6 | 19 | | |
| | | f = 1.2 GHz–2.4 GHz | 6 | 14 | | |
| | | f = 2.4 GHz–3.0 GHz | 3 | 14 | | |
| TX _{DiffRLSlope} | Differential mode RL slope | f = 300 MHz – 3.0 GHz (see Figure 6) | | –13 | | dB/dec |
| RL _{CMTX} | Common-mode return loss | f = 150 MHz–300 MHz | 5 | 20 | | dB |
| | | f = 300 MHz–600 MHz | 5 | 16 | | |
| | | f = 600 MHz–1.2 GHz | 2 | 13 | | |
| | | f = 1.2 GHz–2.4 GHz | 1 | 8 | | |
| | | f = 2.4 GHz–3.0 GHz | 1 | 8 | | |
| IB _{TX} | Impedance balance | f = 150 MHz–300 MHz | 30 | 38 | | dB |
| | | f = 300 MHz–600 MHz | 30 | 38 | | |
| | | f = 600 MHz–1.2 GHz | 20 | 33 | | |
| | | f = 1.2 GHz–2.4 GHz | 10 | 25 | | |
| | | f = 2.4 GHz–3.0 GHz | 10 | 25 | | |
| | | f = 3.0 GHz–5.0 GHz | 4 | 21 | | |
| | | f = 5.0 GHz–6.5 GHz | 4 | 21 | | |
| Diff _{VppTX} | Differential output voltage swing | f = 3.0 GHz (under no interconnect loss) | 400 | 650 | 900 | mV/ppd |
| VCM _{AC_TX} | TX AC CM voltage | At 1.5 GHz | | 15 | 50 | mV/ppd |
| | | At 3.0 GHz | | 10 | 26 | dBmv (rms) |
| | | At 6.0 GHz | | 12 | 30 | |
| VCM _{TX} | Common-mode voltage | | | 1.70 | | V |
| T _{20-80TX} | Rise/Fall time | Rise times and fall times measured between 20% and 80% of the signal. At 6Gbps under no load conditions measured at the pin | 44 | 58 | 85 | ps |
| T _{skewTX} | Differential skew | Difference between the single-ended mid-point of the TX+ signal rising/falling edge, and the single-ended mid-point of the TX– signal falling/rising edge, D1, D0 = V _{CC} | | 2 | 20 | ps |
| TxR/F _{imb} | TX rise/fall imbalance | At 3 Gbps | | 6% | 20% | |
| TxA _{mplmb} | TX amplitude imbalance | | | 1% | 10% | |

ELECTRICAL CHARACTERISTICS (continued)

over recommended operating conditions (unless otherwise noted)

| PARAMETER | | TEST CONDITIONS | MIN | TYP | MAX | UNITS |
|---|-------------------------------|---|------|------|-----|------------------|
| TRANSMITTER JITTER⁽¹⁾ | | | | | | |
| DJ _{TX} | Residual deterministic jitter | VID = 500 mV _{pp} , UI = 333 ps, K28.5 control character, See Figure 7 | 0.12 | 0.19 | | UI _{pp} |
| RJ _{TX} | Random jitter | VID = 500 mV _{pp} , UI = 333 ps, K28.7 control character, See Figure 7 | 1.0 | 2.0 | | ps-rms |
| DJ _{TX} | Residual deterministic jitter | VID = 500 mV _{pp} , UI = 167 ps, K28.5 control character, See Figure 7 | 0.12 | 0.34 | | UI _{pp} |
| RJ _{TX} | Random jitter | VID = 500 mV _{pp} , UI = 167 ps, K28.7 control character, See Figure 7 | 0.95 | 2.0 | | ps-rms |

(1) $T_J = (14.1 \times RJ_{SD} + DJ)$ where RJ_{SD} is one standard deviation value of RJ Gaussian distribution. Jitter measurement is at the SATA connector and includes jitter generated at the package connection on the printed circuit board, and at the board interconnect as shown in Figure 7.

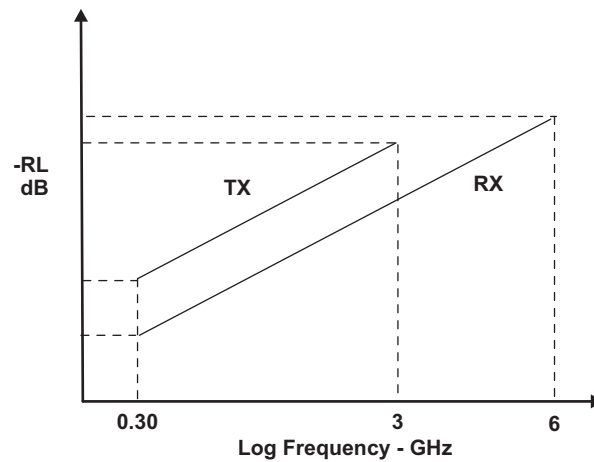


Figure 6. TX, RX Differential Return Loss Limits

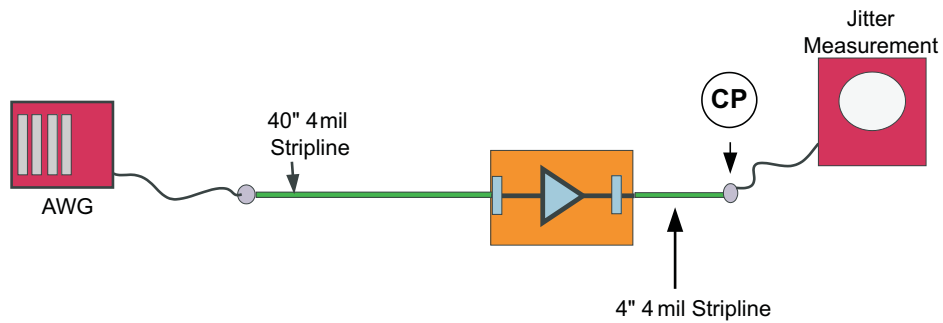


Figure 7. Jitter Measurement Test Condition

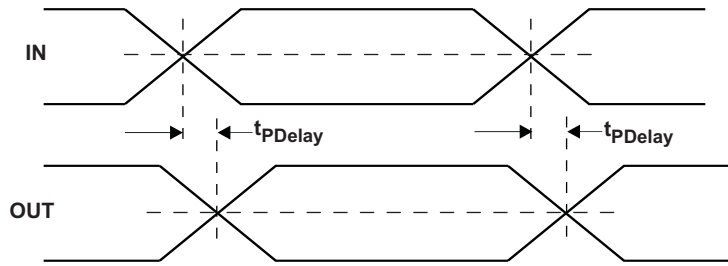


Figure 8. Propagation Delay Timing Diagram

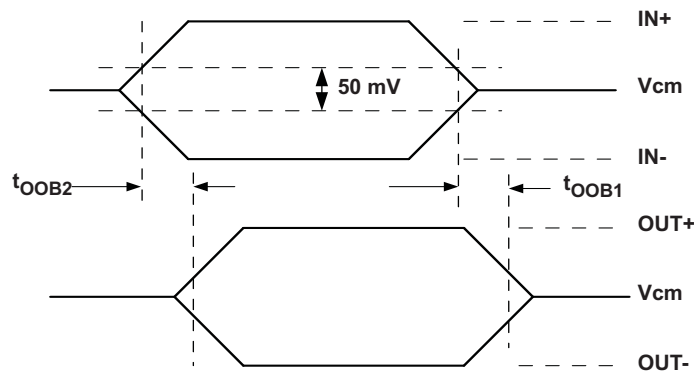


Figure 9. OOB Enter and Exit Timing

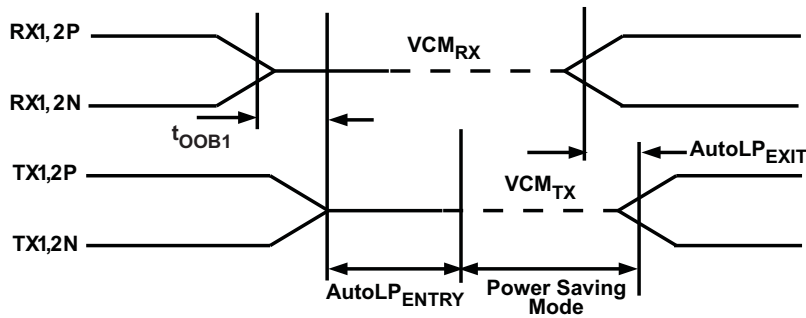


Figure 10. Auto Low Power Mode Entry and Exit Timing

APPLICATION INFORMATION

Figure 12 through Figure 24 show LVCP600 typical performance plots when connected to various trace lengths with $V_{CC} = 3.3\text{ V}$ and $T_A = 25^\circ\text{C}$. All eye diagrams measured using K28.5 pattern at 6 Gbps. Setup for the performance plots is shown in Figure 11.

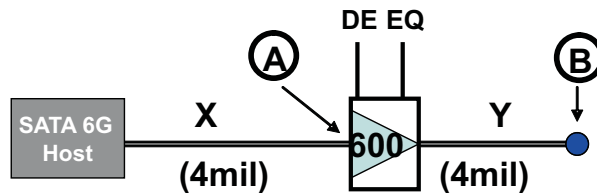


Figure 11. Test Points

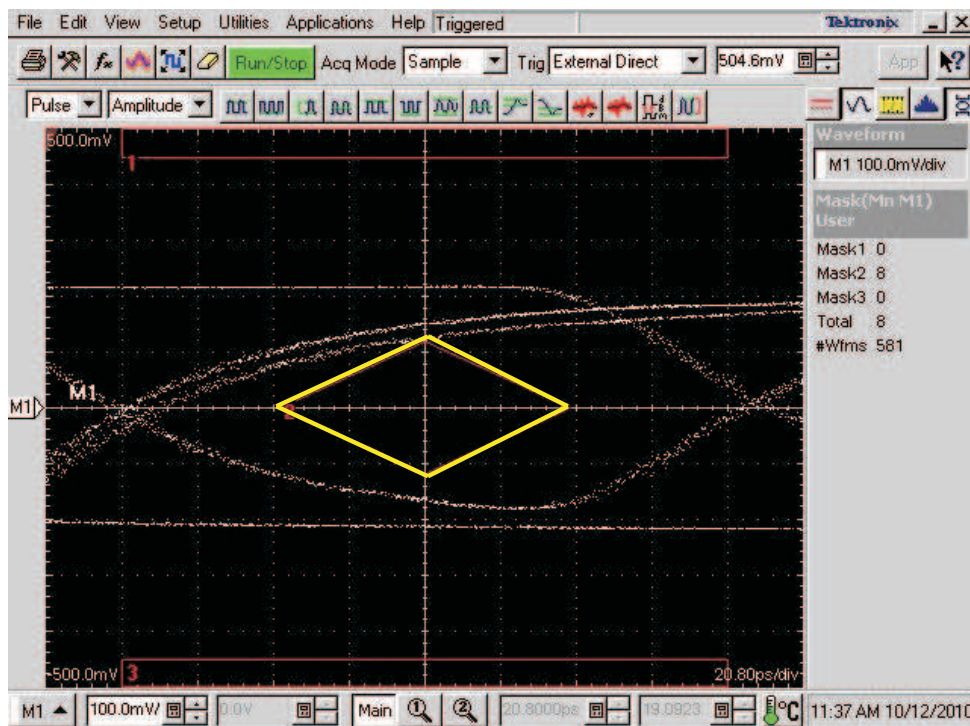


Figure 12. Eye Pattern at A → X= 8"; Y = 2" DE = 1 EQ = 0

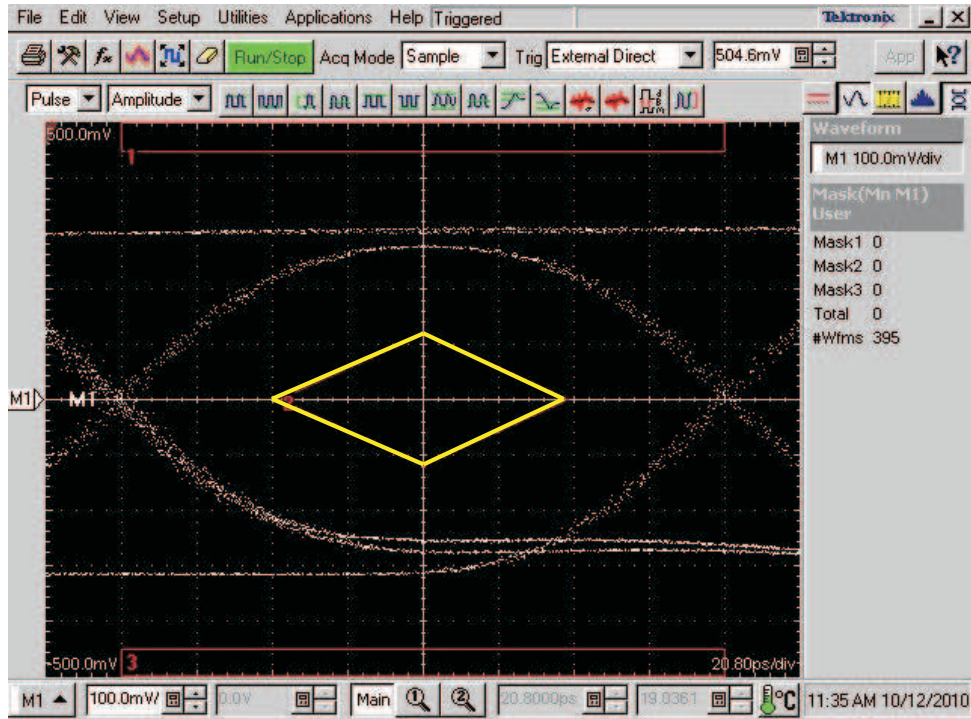


Figure 13. Eye Pattern at B → X= 8"; Y = 2" DE = 1 EQ = 0

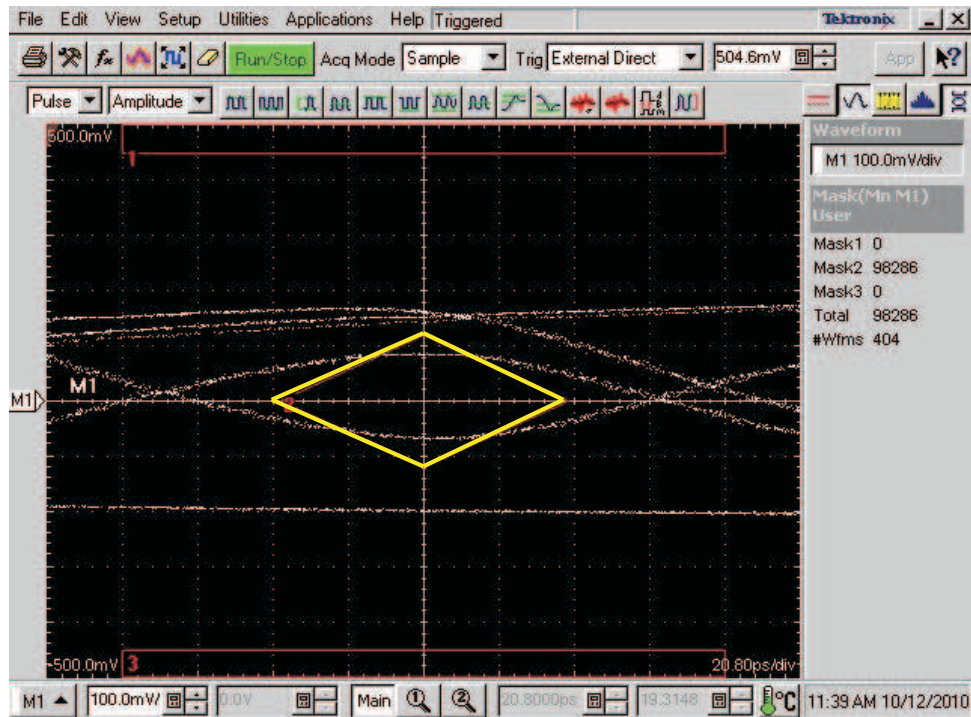


Figure 14. Eye Pattern at A → X= 16"; Y = 2" DE = 1 EQ = 0

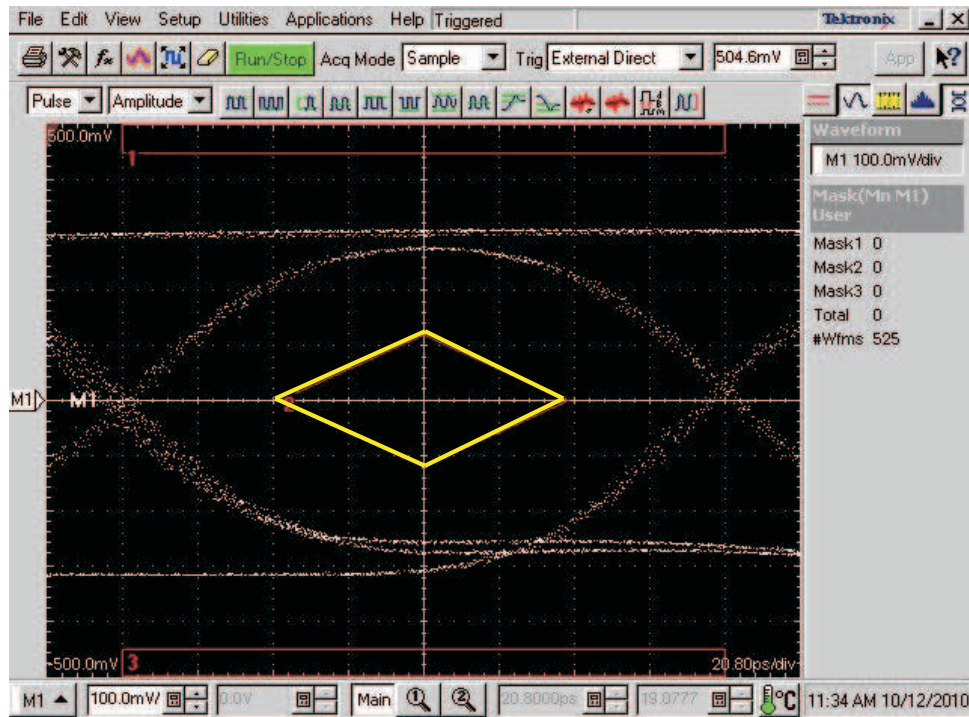


Figure 15. Eye Pattern at B → X= 16"; Y = 2" DE = 1 EQ = 0

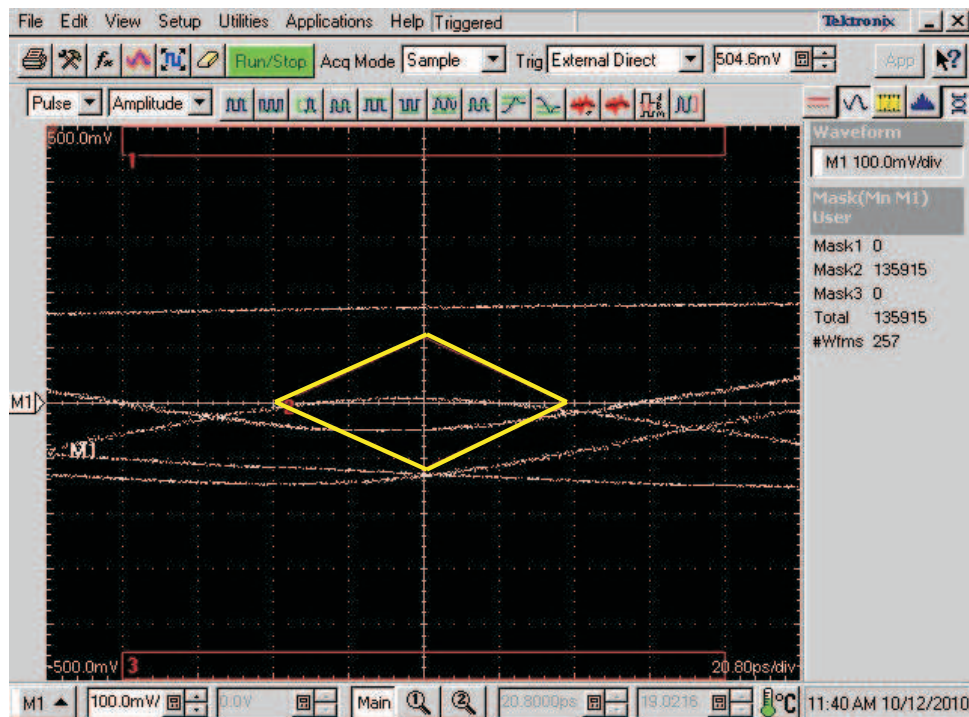


Figure 16. Eye Pattern at A → X= 24"; Y = 2" DE = 1 EQ = 0

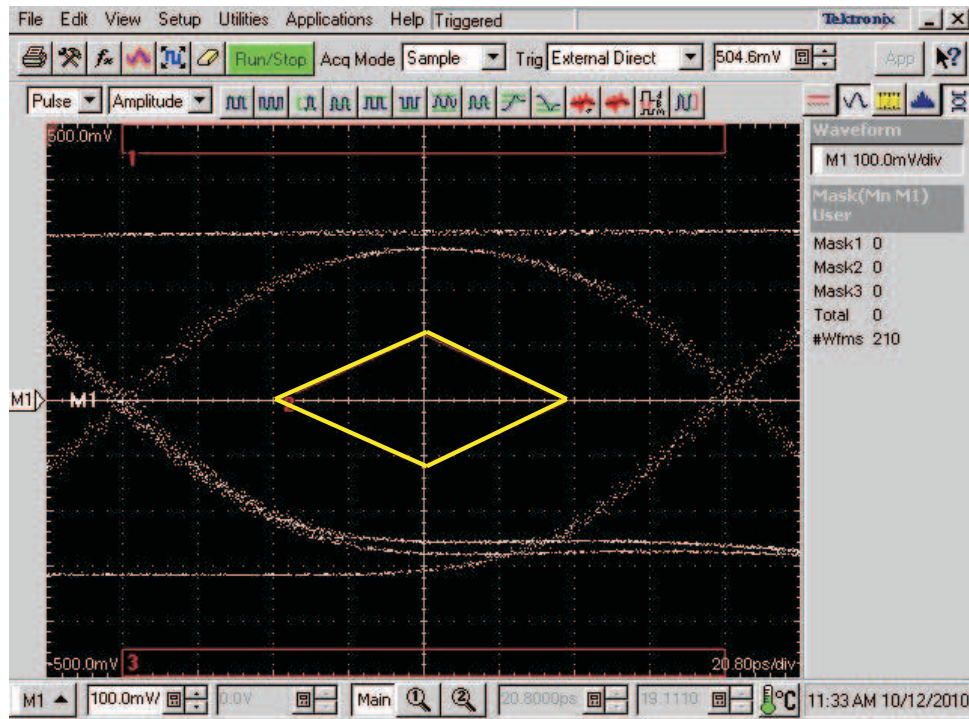


Figure 17. Eye Pattern at B → X= 24"; Y = 2" DE = 1 EQ = 0

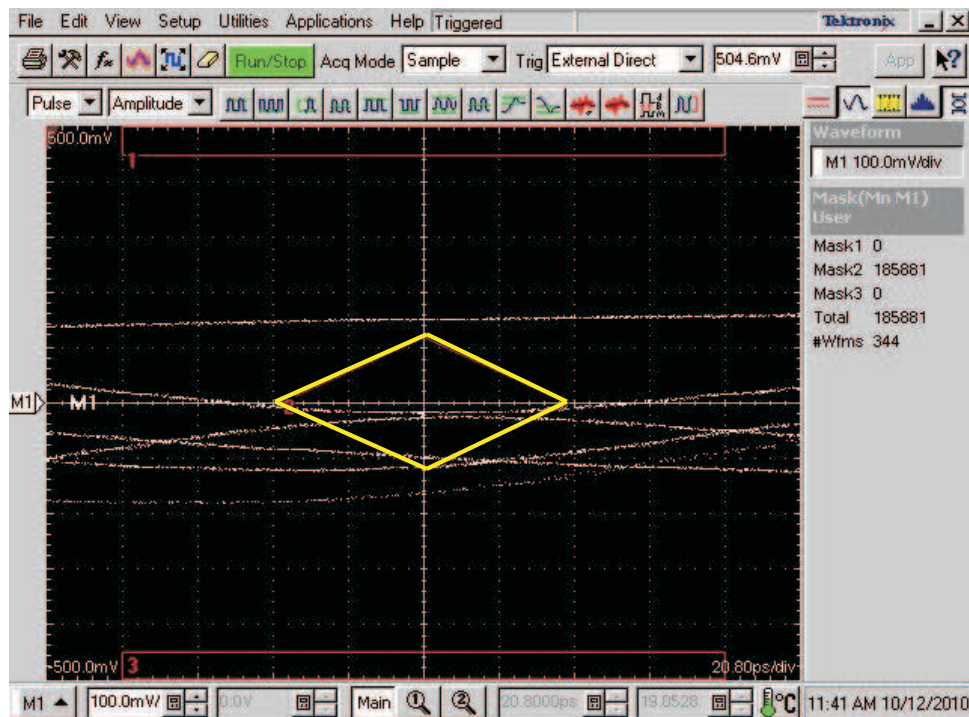


Figure 18. Eye Pattern at A → X= 32"; Y = 2" DE = 1 EQ = 1

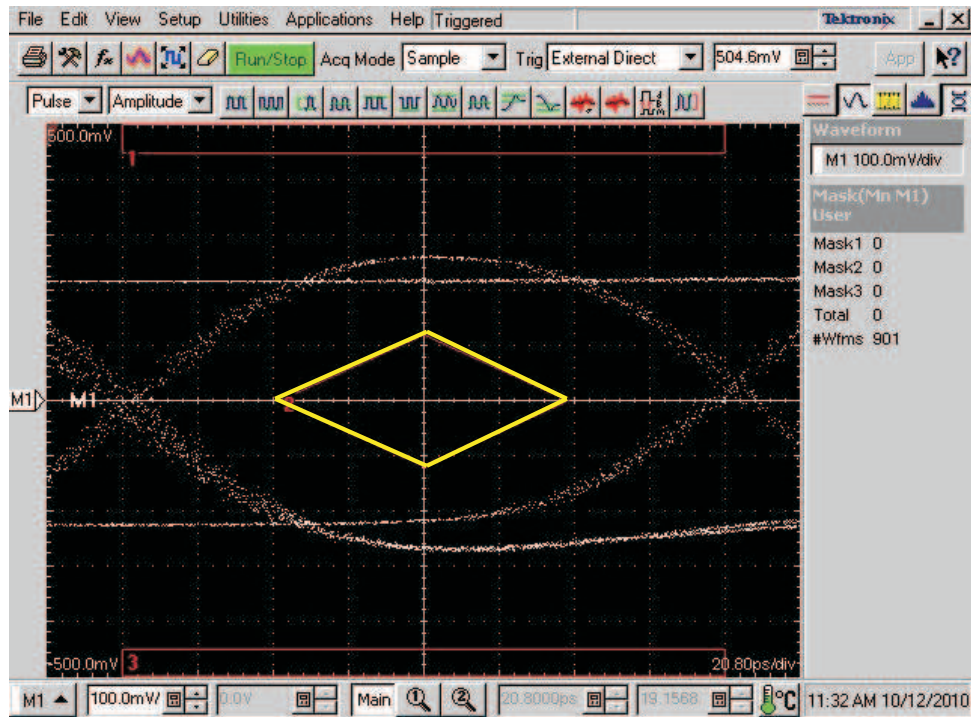


Figure 19. Eye Pattern at B → X= 32"; Y = 2" DE = 1 EQ = 1

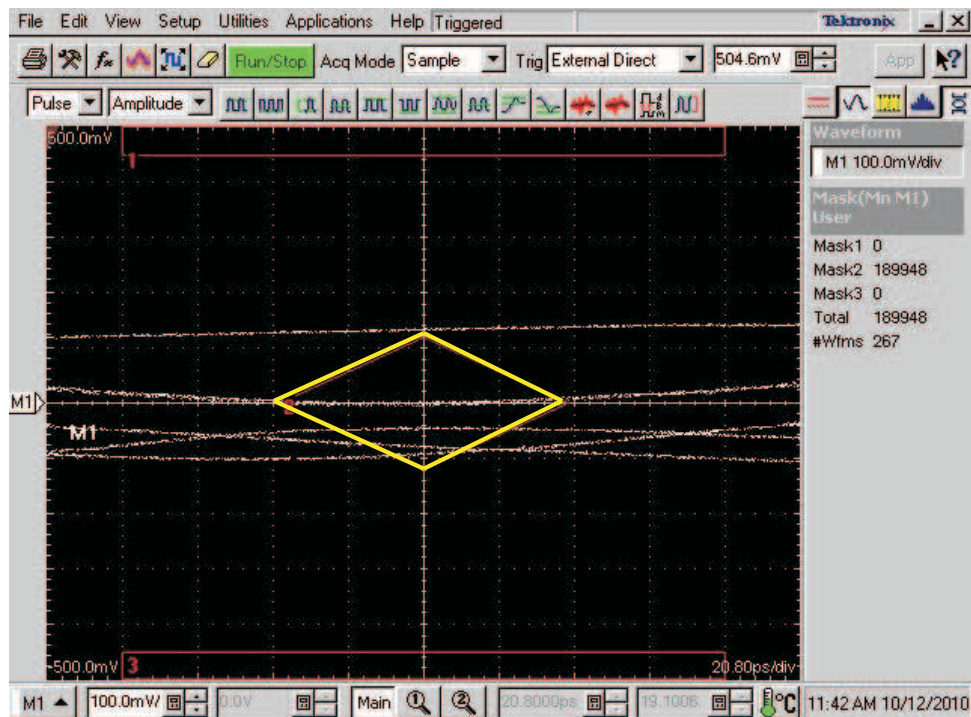


Figure 20. Eye Pattern at A → X= 40"; Y = 2" DE = 1 EQ = 1

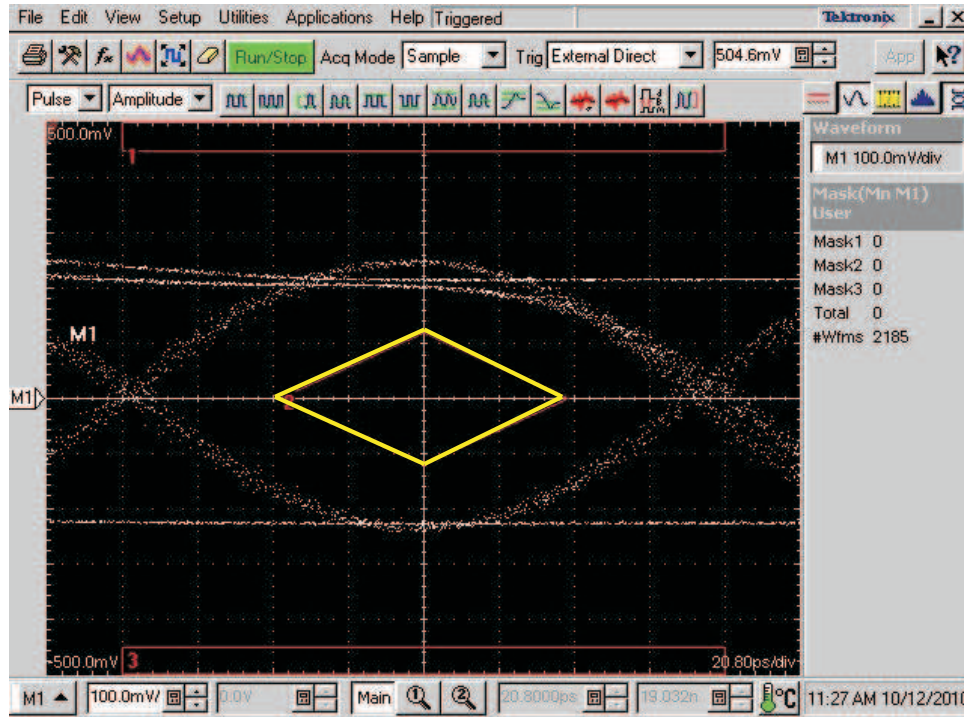


Figure 21. Eye Pattern at B → X= 40"; Y = 2" DE = 1 EQ = 1

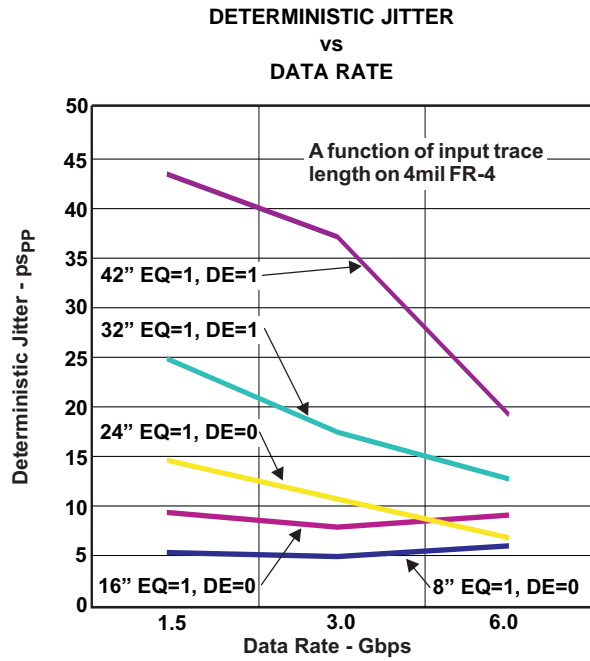


Figure 22.

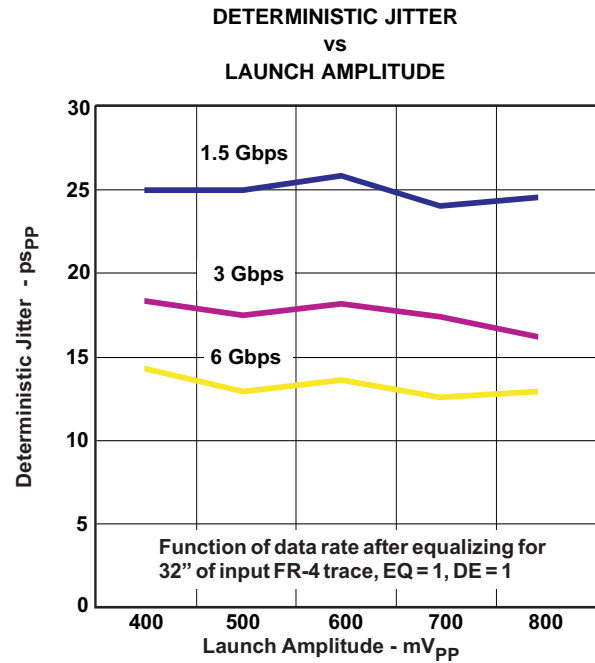


Figure 23.

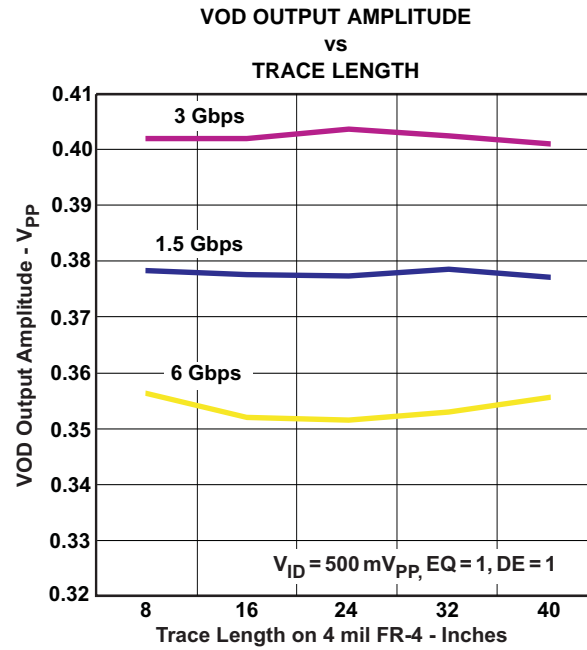


Figure 24.

PACKAGING INFORMATION

| Orderable Device | Status (1) | Package Type | Package Drawing | Pins | Package Qty | Eco Plan (2) | Lead/Ball Finish | MSL Peak Temp (3) | Op Temp (°C) | Top-Side Markings (4) | Samples |
|------------------|---------------|--------------|--------------------|------|-------------|----------------------------|------------------|----------------------|--------------|--------------------------|-------------------------|
| SN75LVCP600DRFR | ACTIVE | WSON | DRF | 8 | 3000 | Green (RoHS & no Sb/Br) | CU NIPDAU | Level-2-260C-1 YEAR | 0 to 85 | 600 | Samples |
| SN75LVCP600DRFT | ACTIVE | WSON | DRF | 8 | 250 | Green (RoHS & no Sb/Br) | CU NIPDAU | Level-2-260C-1 YEAR | 0 to 85 | 600 | Samples |

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) Only one of markings shown within the brackets will appear on the physical device.

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TAPE AND REEL INFORMATION



QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

| Device | Package Type | Package Drawing | Pins | SPQ | Reel Diameter (mm) | Reel Width W1 (mm) | A0 (mm) | B0 (mm) | K0 (mm) | P1 (mm) | W (mm) | Pin1 Quadrant |
|-----------------|--------------|-----------------|------|------|--------------------|--------------------|---------|---------|---------|---------|--------|---------------|
| SN75LVCP600DRFR | WSON | DRF | 8 | 3000 | 180.0 | 8.4 | 2.3 | 2.3 | 1.15 | 4.0 | 8.0 | Q2 |
| SN75LVCP600DRFT | WSON | DRF | 8 | 250 | 180.0 | 8.4 | 2.3 | 2.3 | 1.15 | 4.0 | 8.0 | Q2 |

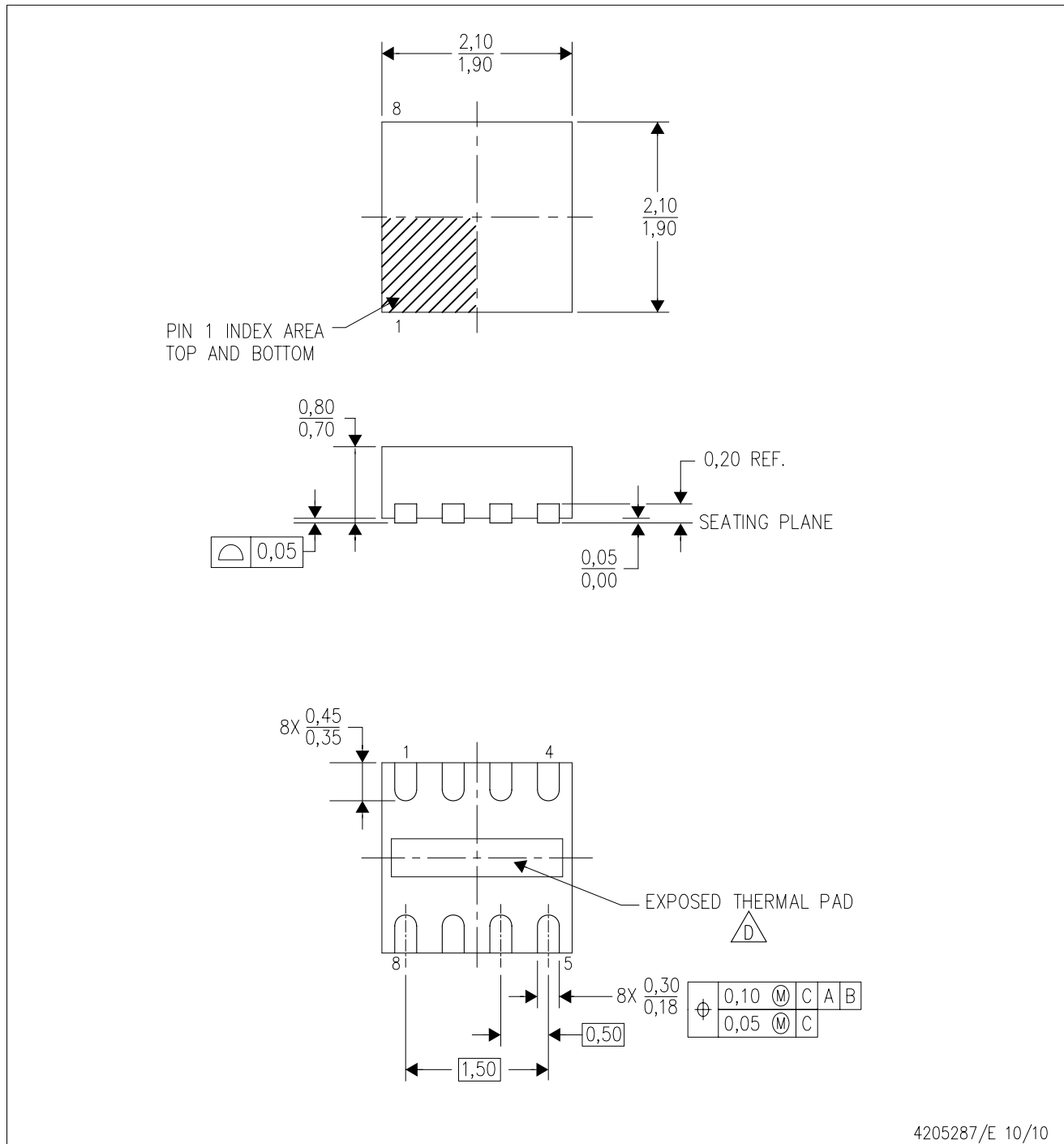
TAPE AND REEL BOX DIMENSIONS



*All dimensions are nominal

| Device | Package Type | Package Drawing | Pins | SPQ | Length (mm) | Width (mm) | Height (mm) |
|-----------------|--------------|-----------------|------|------|-------------|------------|-------------|
| SN75LVCP600DRFR | WSON | DRF | 8 | 3000 | 210.0 | 185.0 | 35.0 |
| SN75LVCP600DRFT | WSON | DRF | 8 | 250 | 210.0 | 185.0 | 35.0 |

DRF (S-PWSON-N8)

PLASTIC SMALL OUTLINE NO-LEAD



- NOTES:
- A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.
 - B. This drawing is subject to change without notice.
 - C. Quad Flatpack, No-Leads (QFN) package configuration.
 -  D. The Package thermal pad must be soldered to the board for thermal and mechanical performance. See product data sheet for details regarding the exposed thermal pad dimensions.
 - E. Falls within JEDEC MO-229.

THERMAL PAD MECHANICAL DATA

DRF (S-PWSON-N8)

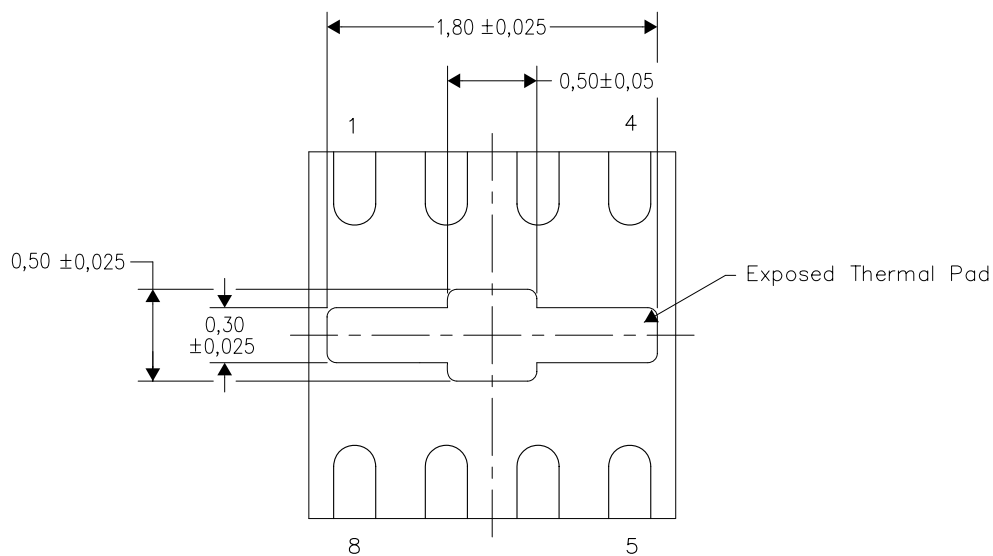
PLASTIC SMALL OUTLINE NO-LEAD

THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No-Lead (QFN) package and its advantages, refer to Application Report, QFN/SOIC PCB Attachment, Texas Instruments Literature No. SLUA271. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.



Bottom View

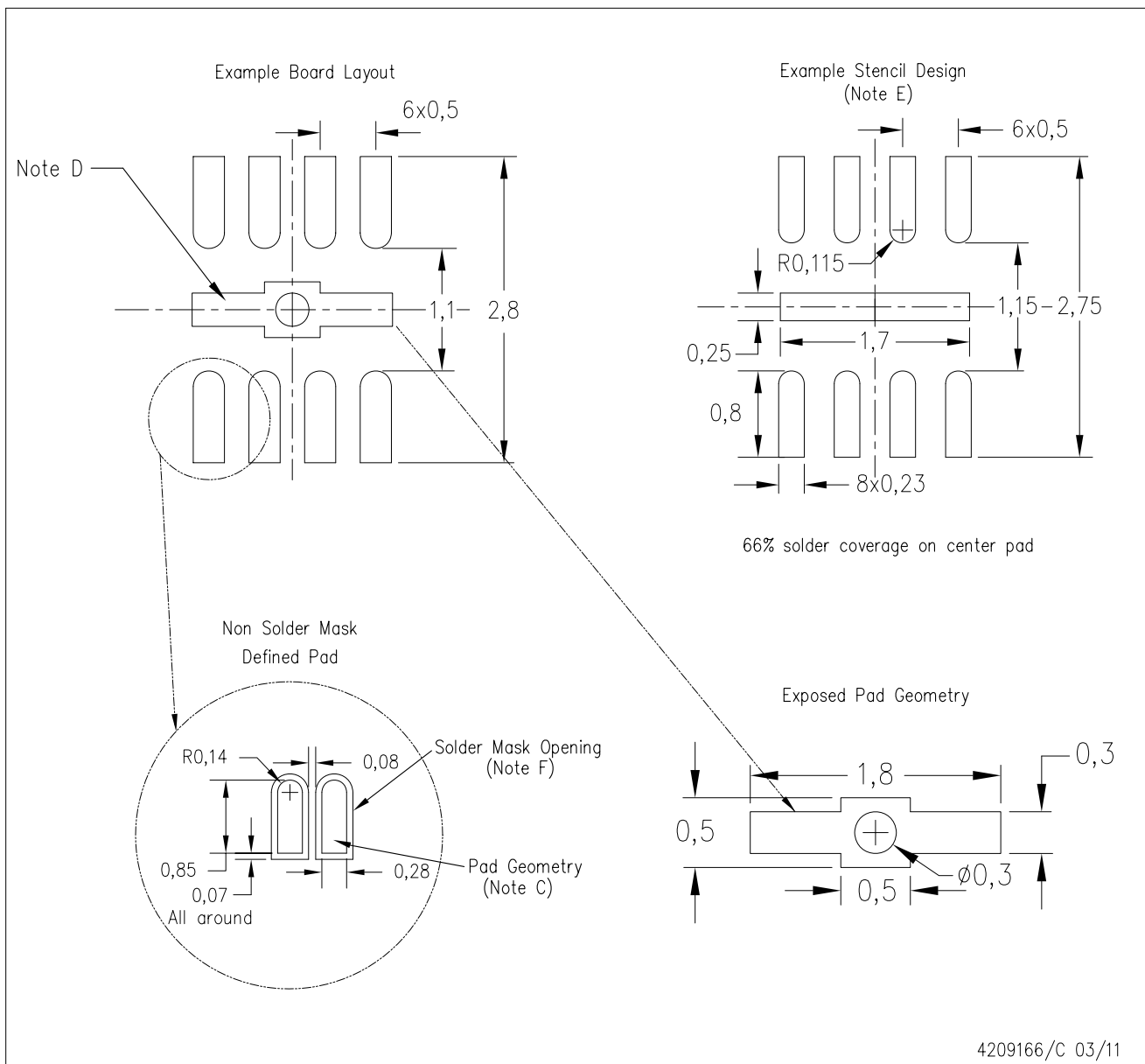
Exposed Thermal Pad Dimensions

4206840/G 04/11

NOTE: A. All linear dimensions are in millimeters

DRF (S-PWSON-N8)

PLASTIC SMALL OUTLINE NO-LEAD



- NOTES:
- All linear dimensions are in millimeters.
 - This drawing is subject to change without notice.
 - Publication IPC-7351 is recommended for alternate designs.
 - This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, Quad Flat-Pack Packages, Texas Instruments Literature No. SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com <<http://www.ti.com>>.
 - Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
 - Customers should contact their board fabrication site for minimum solder mask web tolerances between signal pads.

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