AL1696

HIGHLY INTEGRATION OFFLINE DIMMABLE LED DRIVER

Description

The AL1696 is a highly integrated high power factor constant current converter for mains dimmable LED driver applications. The AL1696 controls the system operating in boundary conduction mode (BCM) to ease in EMI/EMC qualification and testing to meet the latest regulatory standards.

The AL1696 with build-in MOSFET solution reduces the bill of material (BOM) cost by eliminating the need of auxiliary winding and external high voltage MOSFET.

The AL1696 can be compatible with a wide range of dimmers including leading edge and trailing edge dimmers. Some dimmers can achieve deep dimming down to 1%. The AL1696 dimming curve is compliant with the standard of NEMA SSL6.

The AL1696 is available in SO-7 package, which provides an extra pin spacing between the high voltage MOSFET's drain and low voltage pins to increase electricity isolation.
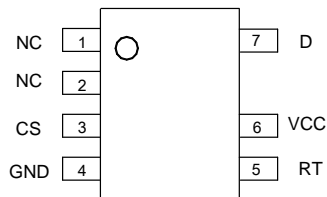
Features

- Internal MOSFET up to 300V/3A, 500V/2A, 600V/2A
- Tight Current Sense Tolerance: ±3%
- Low Startup Current: 150µA
- Low Operation Current: 120µA (Switching Frequency at 5kHz)
- Single Winding Inductor
- Wide Range of Dimmer Compatibility
- Dimming Curve Compliant with NEMA SSL6
- Internal Protections
  - Under Voltage Lockout (UVLO)
  - Leading-Edge Blanking (LEB)
  - Cycle-by-cycle Over Current Protection (OCP)
  - Output Short Protection (OSP)
  - Over-Temperature Protection (OTP)
- SO-7 Package
- **Totally Lead-Free & Fully RoHS Compliant (Notes 1 & 2)**
- **Halogen and Antimony Free. "Green" Device (Note 3)**

Pin Assignments

(Top View)

NC 1	7 D
NC 2	
CS 3	6 VCC
GND 4	5 RT

SO-7

Applications

- Mains Dimmable LED Lamps
- Offline LED Power Supply Driver

Notes: 1. No purposely added lead. Fully EU Directive 2002/95/EC (RoHS) & 2011/65/EU (RoHS 2) compliant.
2. See http://www.diodes.com/quality/lead_free.html for more information about Diodes Incorporated's definitions of Halogen- and Antimony-free, "Green" and Lead-free.
3. Halogen- and Antimony-free "Green" products are defined as those which contain <900ppm bromine, <900ppm chlorine (<1500ppm total Br + Cl) and <1000ppm antimony compounds.
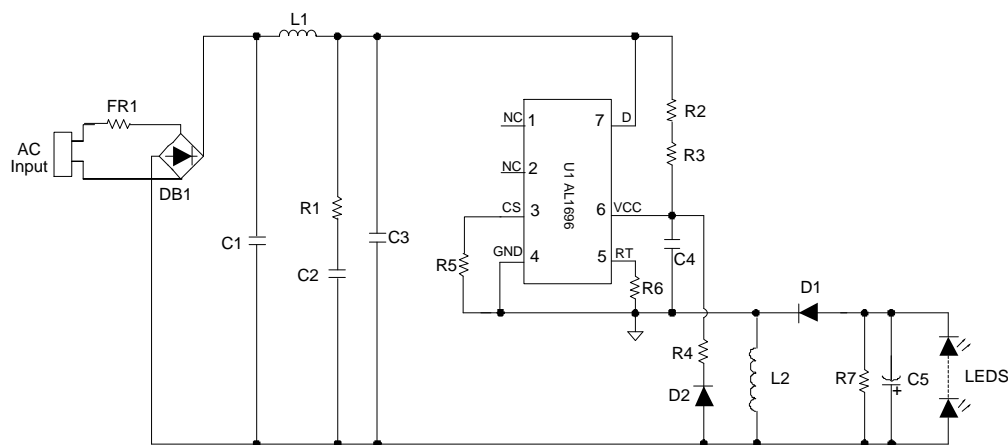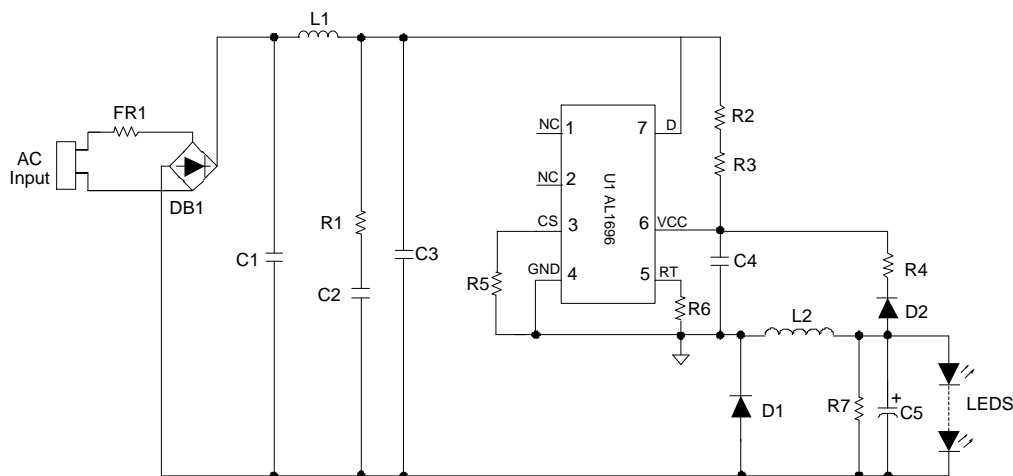
AL1696
Document number: DS38304 Rev. 1 - 2

1 of 14
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Typical Applications Circuit

Buck-Boost Application Circuit

Buck Application Circuit

Pin Descriptions

Pin Number	Pin Name	Function
1, 2	NC	No connection
3	CS	Current sensing
4	GND	Ground
5	RT	Resistor set the system's maximum t_{ON}
6	VCC	Power supply voltage
7	D	Drain of the internal high voltage MOSFET

AL1696
Document number: DS38304 Rev. 1 - 2

2 of 14
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Functional Block Diagram

AL1696
Document number: DS38304 Rev. 1 - 2

3 of 14
www.diodes.com

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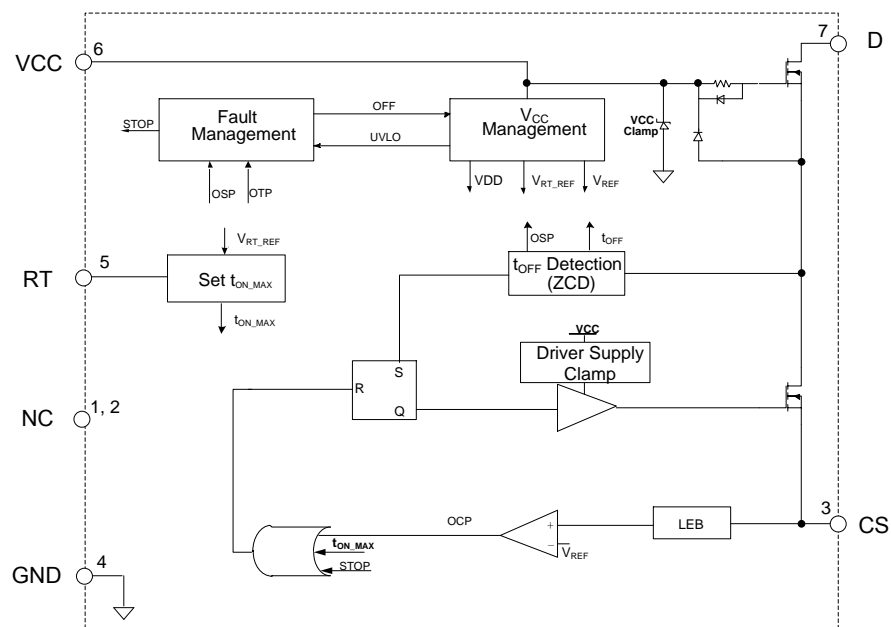
Absolute Maximum Ratings (@T$_A$ = +25°C, unless otherwise specified.) (Note 4)

Symbol	Parameter	Rating	Unit
V$_{CC}$	Power Supply Voltage	18	V
V$_D$	Voltage on D Pin (AL1696-30AS7-13)	300	V
	Voltage on D Pin (AL1696-20BS7-13)	500	V
	Voltage on D Pin (AL1696-20CS7-13)	600	V
I$_D$	Continuous Drain Current T$_C$ = +25°C (AL1696-30AS7-13)	3.0	A
	Continuous Drain Current T$_C$ = +25°C (AL1696-20BS7-13)	2.0	A
	Continuous Drain Current T$_C$ = +25°C (AL1696-20CS7-13)	2.0	A
V$_{CS}$	Voltage on CS Pin	-0.3 to 7	V
V$_{RT}$	Voltage on RT Pin	-0.3 to 7	V
T$_J$	Operating Junction Temperature	-40 to +150	°C
T$_{STG}$	Storage Temperature	-65 to +150	°C
T$_{LEAD}$	Lead Temperature (Soldering, 10 seconds)	+260	°C
P$_D$	Power Dissipation and Thermal Characteristics (T$_A$ = +50°C)	0.8	W
θ$_{JA}$	Thermal Resistance (Junction to Ambient) (Note 5)	123	°C/W
θ$_{JC}$	Thermal Resistance (Junction to Case) (Note 5)	19	°C/W
–	ESD (Human Body Model)	2,000	V
	ESD (Machine Model)	200	V

Notes: 4. Stresses greater than those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "Recommended Operating Conditions" is not implied. Exposure to "Absolute Maximum Ratings" for extended periods may affect device reliability.

5. Device mounted on 1"x1" FR-4 substrate PCB, 2oz copper, with minimum recommended pad layout.

Recommended Operating Conditions (@T$_A$ = +25°C, unless otherwise specified.)

Symbol	Parameter	Min	Max	Unit
T$_A$ (Note 6)	Ambient Temperature	-40	+105	°C

Note: 6. The device can operate normally at +125°C ambient temperature under the condition that the junction temperature is less than +150°C.
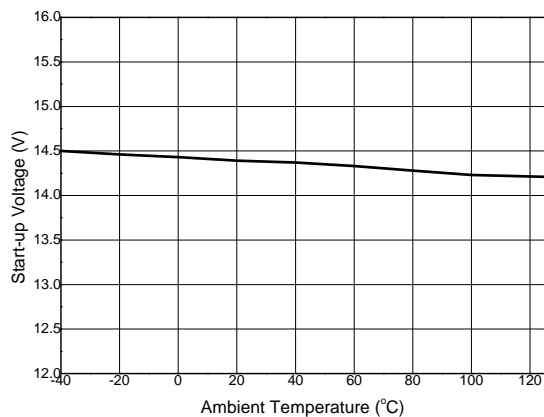
AL1696
Document number: DS38304 Rev. 1 - 2

4 of 14
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October 2015
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Electrical Characteristics (@T$_A$ = +25°C, unless otherwise specified.)

| Symbol	Parameter	Conditions	Min	Typ	Max	Unit
UVLO						
V$_{TH(ST)}$	Startup Voltage	–	–	14.5	–	V
V$_{OPR(min)}$	Minimal Operating Voltage	After Turn On	–	8.5	–	V
V$_{CC\_CLAMP}$	V$_{CC}$ Clamp Voltage	I$_{CC}$ = 1mA	–	16	–	V
Standby Current						
I$_{ST}$	Start-Up Current	V$_{CC}$ = V$_{TH(ST)}$-0.5V, Before Start Up	–	150	–	µA
I$_{CC(OPR)}$	Operating Current	Switching Frequency at 5kHz	–	120	–	µA
Internal High Voltage MOSFET						
R$_{DS(ON)}$	Drain-Source On-State Resistance	AL1696-30A	–	–	3	Ω
		AL1696-20B	–	–	6	Ω
		AL1696-20C	–	–	5.5	Ω
I$_{DS}$	Continuous Drain-Source Current	AL1696-30A	–	–	3.0	A
		AL1696-20B	–	–	2.0	A
		AL1696-20C	–	–	2.0	A
V$_{DS}$	Drain-Source Breakdown Voltage	AL1696-30A	300	–	–	V
		AL1696-20B	500	–	–	V
		AL1696-20C	600	–	–	V
I$_{DSS}$	Drain-Source Leakage Current	AL1696-30A @V$_{DS}$ = 300V, V$_{GS}$ = 0V	–	–	1	µA
		AL1696-20B @V$_{DS}$ = 500V, V$_{GS}$ = 0V	–	–	1	µA
		AL1696-20C @V$_{DS}$ = 600V, V$_{GS}$ = 0V	–	–	1	µA
RT						
V$_{RT}$	Reference Voltage of RT Pin	–	–	0.5	–	V
Current Sense						
V$_{REF}$	Current Sense Reference	–	0.388	0.400	0.412	V
t$_{ON\_MIN}$	Minimum t$_{ON}$	–	–	550	–	ns
t$_{ON\_MAX}$	Maximum t$_{ON}$	R$_T$ = 27kΩ	–	8	–	µs
t$_{OFF\_MAX}$	Maximum t$_{OFF}$	–	–	180	–	µs
t$_{OFF\_MIN}$	Minimum t$_{OFF}$ (Note 7)	–	–	6	–	µs
Over Temperature Protection						
–	Shutdown Temperature (Notes 7 & 8)	–	–	+170	–	°C

Notes: 7. These parameters, although guaranteed by design, are not 100% tested in production.

8. The device will latch when OTP happen and the device won't operate constantly at this temperature.

AL1696
Document number: DS38304 Rev. 1 - 2

5 of 14
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October 2015
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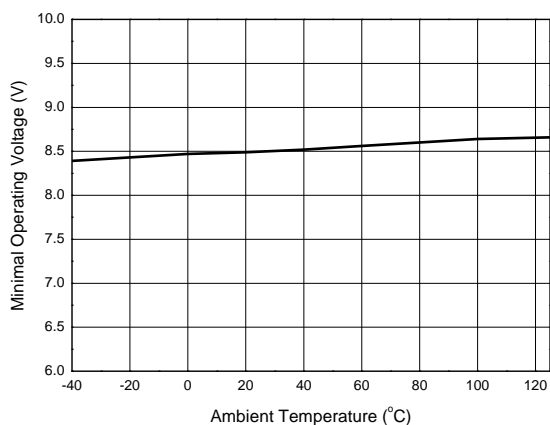## Performance Characteristics (Note 9)

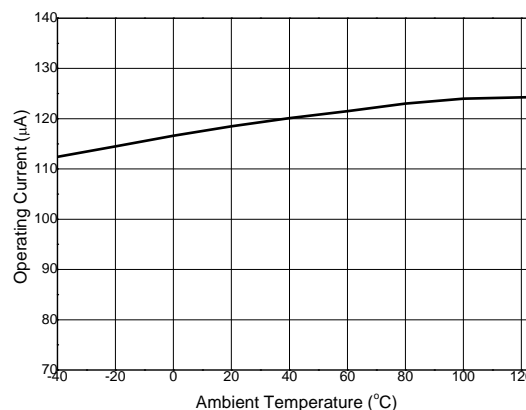**Start-up Voltage vs. Ambient Temperature**

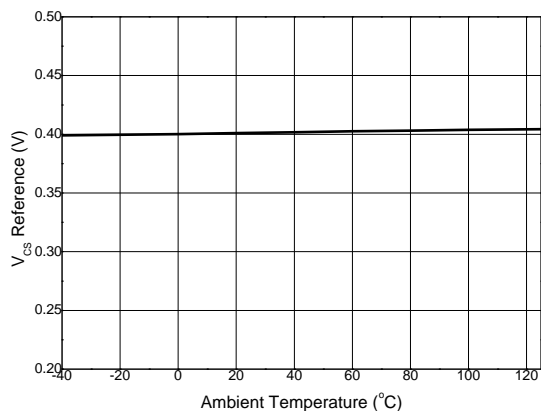**Start-up Current vs. Ambient Temperature**
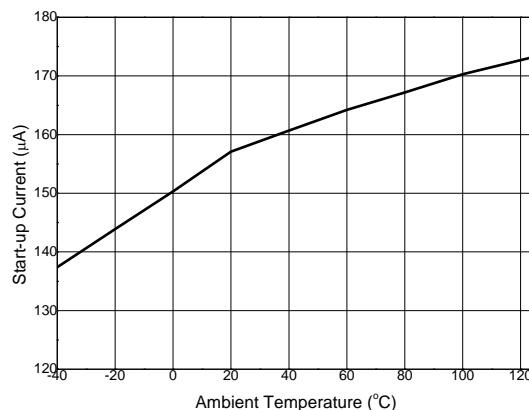
Minimal Operating Voltage vs. Ambient Temperature

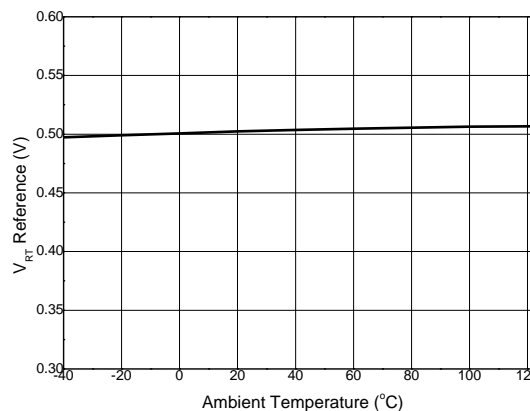**Operating Current vs. Ambient Temperature**

V_{CS} Reference vs. Ambient Temperature

V_{RT} Reference vs. Ambient Temperature

AL1696
Document number: DS38304 Rev. 1 - 2

6 of 14
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Performance Characteristics (Cont.) (Note 9)

VCC Clamp Voltage vs. Ambient Temperature
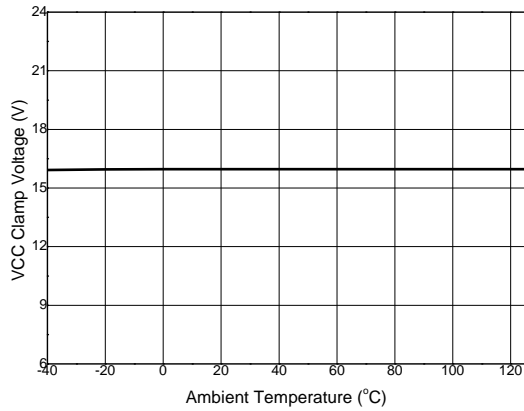
Note: 9. These electrical characteristics are tested under DC condition. The ambient temperature is equal to the junction temperature of the device.

AL1696
Document number: DS38304 Rev. 1 - 2

7 of 14
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Functional Description and Application Information

Convertor Operation

The AL1696 is a single stage, single winding inductor, high efficiency, and high power factor LED driver solution for mains input phase-cutting dimmable application. It is available for three internal MOSFET options (300V/3A, 500V/2A and 600V/2A) which helps to reduce the overall LED driver solution's size and optimize bom cost. The three different MOSFET options can cover most of 3~12W dimmable applications.

The AL1696 internal MOSFET's on time is limited by t_{ON_MAX} which is set through RT pin and internal 0.4V reference. The MOSFET will be turned off either its on time triggers t_{ON_MAX} or input voltage on CS pin triggers internal 0.4V reference voltage. So if the t_{ON_MAX} is set to a very small value, the system will operate in constant on-time (t_{ON_MAX}) mode during the whole rectified mains cycle. It will result in a good power factor, but the line regulation will be worse. Normally, a recommended t_{ON_MAX} should make the system operate in constant on-time (t_{ON_MAX}) mode at valley of input voltage and peak current mode at the crest. A trade off between PF and line regulation need to be done when setting t_{ON_MAX}.

The AL1696 adopts source-driver technology to decrease the system operating current. Besides, it uses a novel t_{OFF} time detection method without auxiliary winding. The AL1696 operates in boundary conduction mode (BCM) which can ease EMI design. All of them help AL1696 to have an extremely low bill of material (BOM).
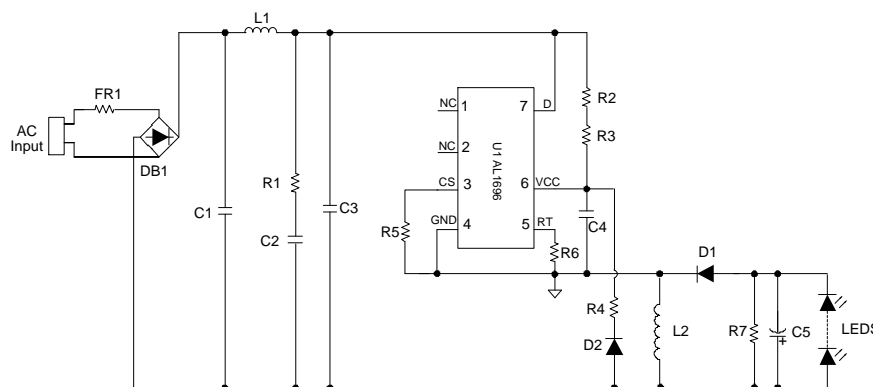

Figure 1. Typical Buck-Boost Application Circuit

Start-up and Supply Voltage

During start-up, the VCC capacitor C4 is charged through startup resistors (R2, R3) from the rectified mains input until the start-up voltage is reached, the AL1696 starts switching. In normal operation, the VCC supply is provided from two paths: one is from start-up resisters (R2, R3) and the other is from output voltage (V_{OUT}) through one diode (D2). In this way the system can provide sufficient VCC supply at low dimming angle.

The AL1696 has internal VCC clamp, the typical voltage is 16V. VCC voltage needs to be between $V_{OPR(min)}$ and V_{CC_CLAMP} during normal operation.

When VCC voltage drops below the under voltage lockout (UVLO), IC will stop switching. The IC will restart once the voltage on VCC pin exceeds the startup voltage ($V_{TH(ST)}$) again.

Protections

Under Voltage Lockout (UVLO)
When the voltage on the VCC pin drops below $V_{OPR(Min)}$, the IC will stop switching. The IC will restart until the VCC voltage exceeds the startup voltage ($V_{TH\,(ST)}$) again.

Leading-Edge Blanking (LEB)
To prevent false detection of the peak current through MOSFET, a blanking time following switch-on is designed. When the internal switch turns on, a short current spike may appear because of the discharge of the parasitic capacitor over MOSFET's drain and source. It will be ignored during the LEB time (t_{ON_MIN}).

Cycle-by-cycle Over Current Protection (OCP)
The AL1696 has a built-in peak current detector. The R5 connected to the CS pin is used to sense the current through MOSFET and will be one of the inputs. The detection circuit is activated after the LEB time. When the voltage on CS pin reaches V_{REF}, the IC will turn off the switch to limit the output current.

AL1696
Document number: DS38304 Rev. 1 - 2

8 of 14
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Functional Description and Application Information (Cont.)

The peak current (I_{PEAK}) of the MOSFET can be set as below:

$$I_{PEAK} = \frac{V_{REF}}{R5}$$

It automatically provides a cycle-by-cycle protection of maximum current through MOSFET during operation. A propagation delay exists between over current detection and actual source-switch off, so the actual peak current will be a little higher than the OCP level set by R5.

Output-Short Protection (OSP)

When LED is shorted, the device can't detect the t_{OFF} time, and the system will work with low switching frequency of 5kHz.

Over-Temperature Protection (OTP)

The AL1696 has OTP protection function. When the junction temperature reaches +170°C typical, it will trigger an over-temperature protection which makes IC shut down and latched. Once OTP is triggered, the system will only restart after the system's power supply powers off and on again.

Design Parameters based on buck-boost topology

Setting the Current Sense Resistor R5

In AL1696 typical application, a t_{ON_MAX} is recommended to make the system operate at constant on-time (t_{ON_MAX}) mode at valley of input voltage and peak current mode at the crest. In most of cases, the t_{ON_MAX} will set to be 1.3 times of the t_{ON} at crest, and the critical angle of two operation modes will be:

$$\theta = a\sin\left(\frac{1}{1.3}\right)$$

Then the output current can be calculated as below:

$$I_{O\_MEAN} = \frac{V_{REF}}{2\pi \cdot R5}(\int_0^{\theta} 1.3 \cdot \sin(\theta)\frac{\sqrt{2}V_{IN\_RMS}\sin(\theta)}{\sqrt{2}V_{IN\_RMS}\sin(\theta)+V_O}dt + \int_{\theta}^{\pi-\theta}\frac{\sqrt{2}V_{IN\_RMS}\sin(\theta)}{\sqrt{2}V_{IN\_RMS}\sin(\theta)+V_O}dt + \int_{\pi-\theta}^{\pi}1.3 \cdot \sin(\theta)\frac{\sqrt{2}V_{IN\_RMS}\sin(\theta)}{\sqrt{2}V_{IN\_RMS}\sin(\theta)+V_O}dt)$$

Where,

V_{REF} is the internal reference, typical 0.4V.

R5 is the current sense resistor.

V_{IN_RMS} is the input voltage's RMS value.

V_O is the system output voltage.

Inductance Selection (L2)

The peak current of the MOSFET is calculated as below:

$$I_{PEAK} = \frac{V_{REF}}{R5}$$

The AL1696 is operating in boundary conduction mode which results in a variable operating frequency. The minimum switching frequency f_{MIN} should be set at the crest of the minimum AC input voltage. Inductance should be calculated according to the chosen f_{MIN}:

$$L2 = \frac{\sqrt{2}V_{IN\_RMS} \cdot V_O}{I_{PEAK} \cdot (\sqrt{2}V_{IN\_RMS}+V_O) \cdot f_{MIN}}$$

According to the Faraday's Law of Induction, the winding number of the inductance can be calculated by:

AL1696
Document number: DS38304 Rev. 1 - 2

9 of 14
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Functional Description and Application information (Cont.)

$$N_{L2} = \frac{L2 \cdot I_{PEAK}}{A_e \cdot B_m}$$

Where,

A_e is the core effective area.

B_m is the maximum magnetic flux density.

t_{ON_MAX} Setting

In order to get a high power factor and good dimmer compatibility, the system should operate in constant on time mode at valley of input voltage. It can be realized by setting the maximum on time, which is set by on external resistor connect to RT pin.

And the t_{ON} time has the below equation:

$$t_{ON\_MAX} = \frac{25 \cdot C_{REF} \cdot R6}{V_{RT\_REF}}$$

Where,

V_{RT_REF} is the internal RT pin 0.5V's reference.

C_{REF} is the internal 6pF capacitor.

Dimmer Compatibility

Passive Bleeder Design

The passive bleeder is designed to supply latching and holding current to get rid of dimmer's misfire and flicker.
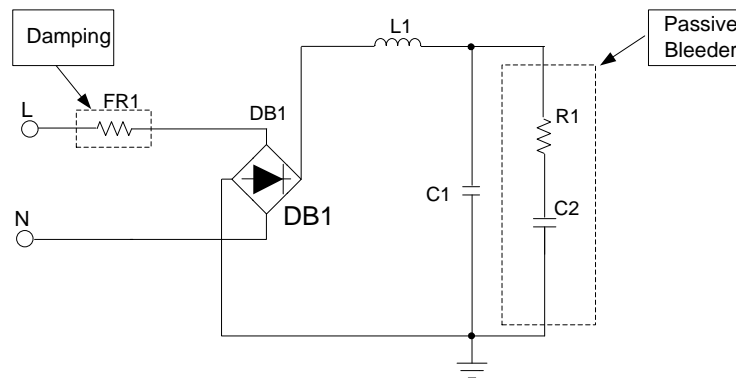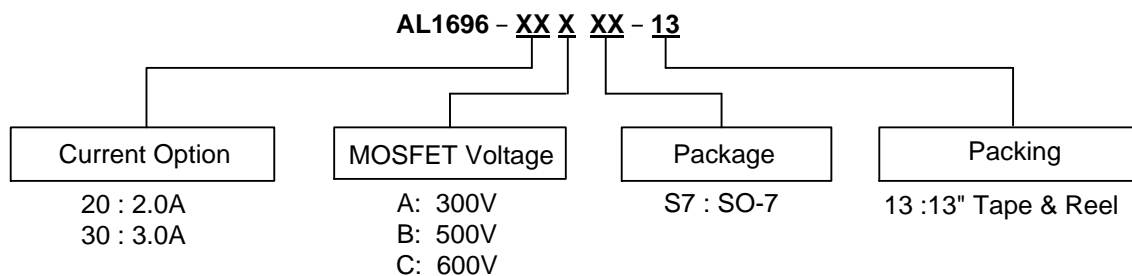

Figure 2. LED Driver Schematic with Passive Bleeder

The passive bleeder includes a capacitor (C2, in hundreds of nF) to provide latching current and a resistor (R1) to limit the current spike. Because a large C2 will affect the PF, THD and efficiency negatively, the value of the capacitor (C2) should be selected carefully. Generally, a capacitance from 100nF/400V to 330nF/400V is recommended. R1 is used to limit the latching current. If R1 is too big, the latching current will be not enough and the TRIAC dimmer will be misfired, resulting in LED flicker. If R1 is too small, it will result in greater power dissipation. Generally, a 200Ω to 2KΩ resistor is recommended for R1.
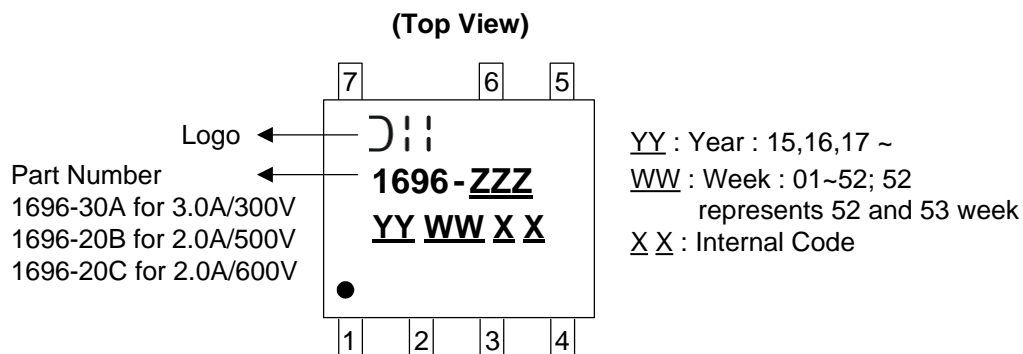
Passive Damping Design

FR1 is to limit the spike current caused by quick charging of C2 when dimmer on. Normally, FR1 will be chosen from 20Ω to 100Ω for low line like 120V$_{AC}$ application, and 51Ω to 200Ω for high line like 230V$_{AC}$ application.

AL1696
Document number: DS38304 Rev. 1 - 2

10 of 14
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October 2015
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Ordering Information

AL1696 – **XX** **X** **XX** – **13**

Current Option	MOSFET Voltage	Package	Packing
20 : 2.0A 30 : 3.0A	A: 300V B: 500V C: 600V	S7 : SO-7	13 :13" Tape & Reel

Part Number	Package Code	Package	13" Tape and Reel	
			Quantity	Part Number Suffix
AL1696-30AS7-13	S7	SO-7	4000/Tape & Reel	-13
AL1696-20BS7-13	S7	SO-7	4000/Tape & Reel	-13
AL1696-20CS7-13	S7	SO-7	4000/Tape & Reel	-13

Marking Information

(Top View)

Logo

Part Number
1696-30A for 3.0A/300V
1696-20B for 2.0A/500V
1696-20C for 2.0A/600V

1696 - **ZZZ**
YY **WW** **X** **X**

<u>YY</u> : Year : 15,16,17 ~
<u>WW</u> : Week : 01~52; 52
    represents 52 and 53 week
<u>X</u> <u>X</u> : Internal Code

AL1696
Document number: DS38304 Rev. 1 - 2

11 of 14
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Package Outline Dimensions (All dimensions in mm.)

(1)   Package Type: SO-7

5.800(0.228)
6.200(0.244)

0.330(0.013)
0.510(0.020)

2.54(0.100)
TYP

1.270(0.050)
TYP

3.800(0.150)
4.000(0.157)

1.350(0.053)
1.750(0.069)

4.700(0.185)
5.100(0.201)

0.080(0.003)
0.250(0.010)

1.250(0.049)
1.500(0.059)

0°
8°

Option 1

0.450(0.017)
0.800(0.031)

0.150(0.006)
0.250(0.010)

0.350(0.014)
TYP

45°
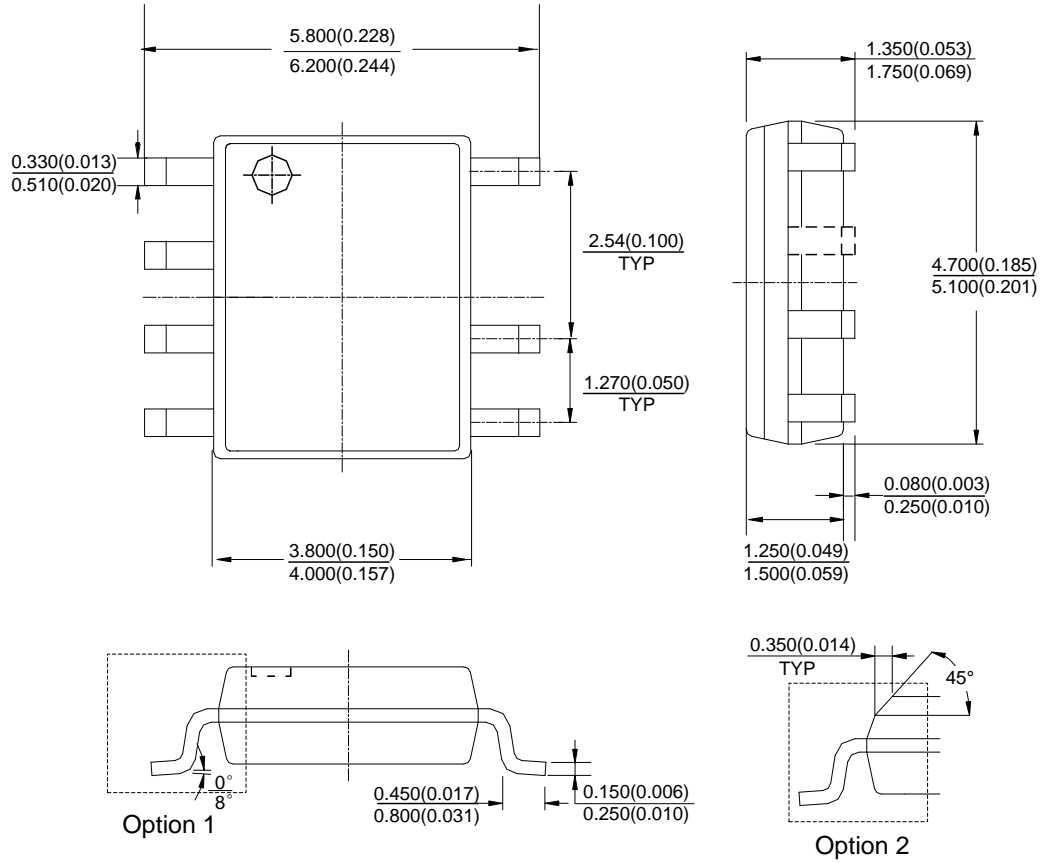
Option 2

Note: Eject hole, oriented hole and mold mark is optional.

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AL1696
Document number: DS38304 Rev. 1 - 2

12 of 14
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Suggested Pad Layout

(1) Package Type: SO-7
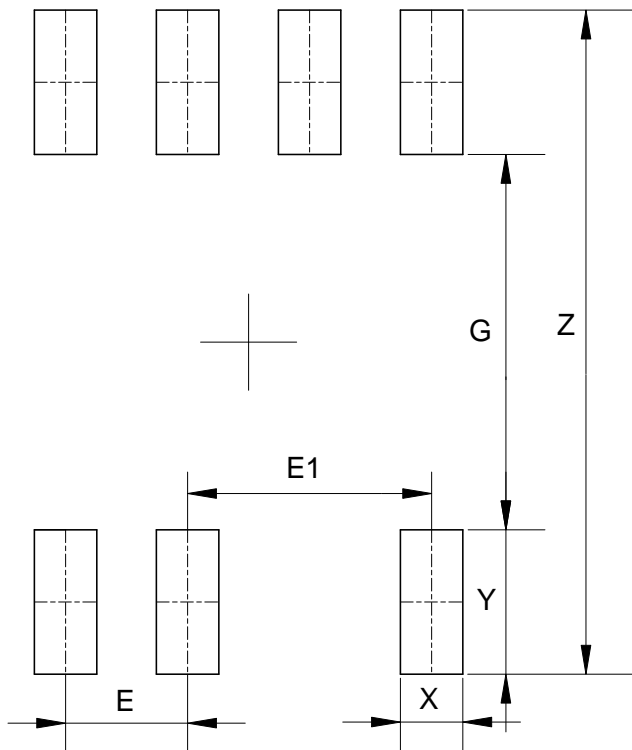
Dimensions	Z (mm)/(inch)	G (mm)/(inch)	X (mm)/(inch)	Y (mm)/(inch)	E (mm)/(inch)	E1 (mm)/(inch)
Value	6.900/0.272	3.900/0.154	0.650/0.026	1.500/0.059	1.270/0.050	2.540/0.100

AL1696
Document number: DS38304 Rev. 1 - 2

13 of 14
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October 2015
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    2. support or sustain life and whose failure to perform when properly used in accordance with instructions for use provided in the labeling can be reasonably expected to result in significant injury to the user.

B. A critical component is any component in a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or to affect its safety or effectiveness.

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AL1696
Document number: DS38304 Rev. 1 - 2

14 of 14
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Наши контакты:

Телефон: +7 812 627 14 35

Электронная почта: sales@st-electron.ru

Адрес: 198099, Санкт-Петербург, Промышленная ул, дом № 19, литера Н, помещение 100-Н Офис 331