

# 4.5V to 18V Input, 3.0A Integrated MOSFET 1ch Synchronous Buck DC/DC Converter

## BD9D321EFJ

#### **General Description**

BD9D321EFJ is a synchronous buck switching regulator with built-in low on-resistance power MOSFETs. It is capable of providing current of up to 3 A. The SLLM<sup>TM</sup> control provides excellent efficiency characteristics in light-load conditions which make the product appropriate for equipment and devices that demand minimal standby power consumption. External phase compensation circuit is not necessary for it is a constant on-time control DC/DC converter with high speed response. •

#### **Features**

- Synchronous Single DC/DC Converter
- Constant On-time Control
- SLLM<sup>TM</sup> (Simple Light Load Mode) Control
- Over Current Protection
- Short Circuit Protection
- Thermal Shutdown Protection
- Under Voltage Lockout Protection
- Adjustable Soft Start
- HTSOP-J8 Package (Backside Heat Dissipation)

## **Applications**

- Step-down Power Supply for DSPs, FPGAs, Microprocessors, etc.
- Set-top Box
- LCD TVs
- DVD / Blu-ray Player / Recorder
- Entertainment Devices

#### **Key Specifications**

- Input Voltage Range: 4.5V to 18.0 V
   Output Voltage Setting Range: 0.765V to 7V
   (V<sub>IN</sub> × 0.07)V to (V<sub>IN</sub> × 0.65)V
- Output Current: 3 A (Max)
- Switching Frequency: 700 kHz (Typ)
- High Side MOSFET On-Resistance:100 m Ω (Typ)
- Low Side MOSFET On-Resistance: 70 m Ω (Typ)
   Standby Current: 2 μA (Typ)

#### Package HTSOP-J8

W (Typ) x D (Typ) x H (Max) 4.90mm x 6.00mm x 1.00mm



#### **Typical Application Circuit**

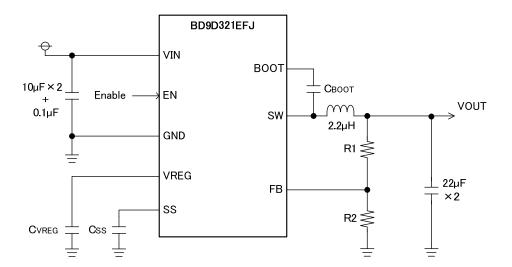


Figure 1. Typical Application Circuit

## **Pin Configuration**

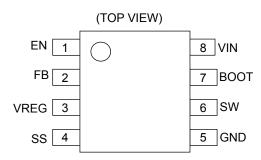


Figure 2. Pin Assignment

## **Pin Descriptions**

Terminal No.	Symbol	Function
1	EN	Turning this terminal signal low level (0.3 V or lower) forces the device to enter the shutdown mode. Turning this terminal signal high level (2.2 V or higher) enables the device. This terminal must be terminated.
2	FB	An inverting input terminal of comparator which compares with reference voltage (V <sub>REF</sub> ). Refer to page 17 for how to calculate the resistance of the output voltage setting.
3	VREG	Power supply voltage terminal inside IC. Voltage of 5.25V (Typ) is outputted with more than 2.2V is impressed to EN terminal. Connect 1µF ceramic capacitor to ground.
4	SS	Terminal for setting the soft start time. The rise time of the output voltage can be specified by connecting a capacitor to this terminal. Refer to page.17 for how to calculate the capacitance.
5	GND	Ground terminal for the output stage of the switching regulator and the control circuit
6	SW	Switch node. This terminal is connected to the source of the high-side MOSFET and drain of the low-side MOSFET. Connect a bootstrap capacitor of 0.1µF between this terminal and BOOT terminal. In addition, connect an inductor considering the direct current superimposition characteristic.
7	воот	Connect a bootstrap capacitor of 0.1µF between this terminal and SW terminal. The voltage of this capacitor is the gate drive voltage of the high-side MOSFET.
8	VIN	Power supply terminal for the switching regulator. Connecting a 20μF(10μF × 2) and 0.1μF ceramic capacitor to ground is recommended.
-	FIN	A backside heat dissipation pad. Connecting to the internal PCB ground plane by using multiple via provides excellent heat dissipation characteristics.

## **Block Diagram**

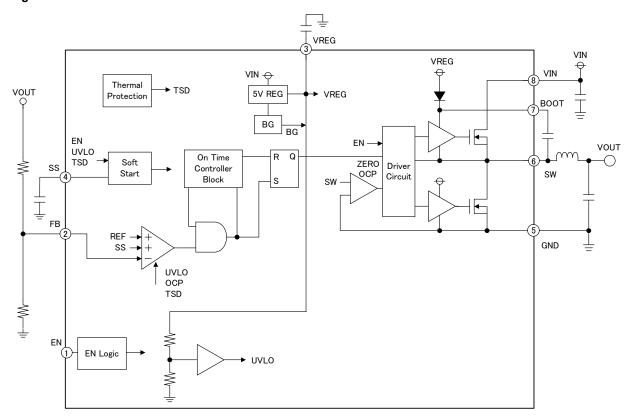


Figure 3. Block Diagram

## Absolute Maximum Ratings (Ta = 25°C)

Parameter	Symbol Rating		Unit
Input Voltage (Note 1)	VIN	20	V
BOOT Voltage (Note 1)	Vвоот	27	V
BOOT-SW Voltage (Note 1)	VBOOT-VSW	7	V
Output Feedback Voltage	VFB	VREG	V
SW Voltage (Note 1)	Vsw	20	V
VREG Voltage (Note 1)	VREG	7	V
SS Voltage (Note 1)	Vss	7	V
Logic Input Voltage (Note 1)	VEN	20	V
Power dissipation (Note 2)	Pd	3.75	W
Operating Temperature Range	Topr	-40 to +85	°C
Storage Temperature Range	Tstg	-55 to +150	°C
Junction Temperature	Tjmax	+150	°C

<sup>(</sup>Note 1) No need to exceed Pd.

(Note 2) Derating in done 30.08 mW/°C for operating above Ta ≥ 25°C (Mount on 4-layer 70.0mm × 70.0mm × 1.6mm board)
Caution1: Operating the IC over the absolute maximum ratings may damage the IC. The damage can either be a short circuit between pins or an open circuit between pins and the internal circuitry. Therefore, it is important to consider circuit protection measures, such as adding a fuse, in case the IC is operated over the absolute maximum ratings.

Caution2: The operating temperature range is intended to guarantee functional operation and does not guarantee the life of the LSI within this range. The life of the LSI is subject to derating depending on usage environment such as the voltage applied, ambient temperature and humidity. Consider derating in the design of equipment and devices.

## **Recommended Operating Conditions**

Parameter	Symbol		Unit			
Farameter	Symbol	Min	Тур	Max	Offic	
Input voltage	VIN	4.5	12	18	٧	
BOOT voltage	Vвоот	4.5	-	24	V	
SW Voltage	Vsw	-0.7	-	+18	٧	
BOOT-SW voltage	VBOOT-VSW	4.5	-	5.5	٧	
Logic Input Voltage	VEN	0	-	18	V	
Output Current	lout	-	-	3	Α	
Output Voltage Range	VRANGE	0.765 <sup>(Note 3)</sup>	-	7 <sup>(Note 4)</sup>	V	

<sup>(</sup>Note 3) Please use under the condition of VOUT ≥ VIN × 0.07 [V].

<sup>(</sup>Note 4) Please use under the condition of VOUT  $\leq$  VIN  $\times$  0.65 [V].

<sup>(</sup>Refer to the page 17 for how to calculate the output voltage setting.)

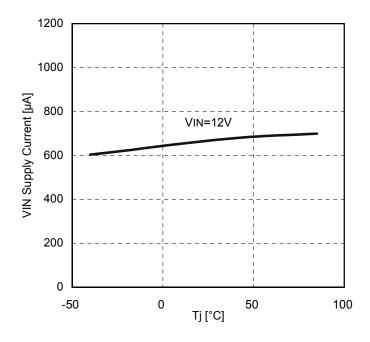
## **Electrical Characteristics**

(Ta = 25°C, VIN = 12V, VEN = 3V unless otherwise specified)

	Symbol	Limit				0 1111
Parameter		Min	Тур	Max	Unit	Conditions
<vin block="" pin=""></vin>		'	<u>'</u>	•		
Standby Circuit Current	I <sub>STB</sub>	-	2	15	μA	VEN=GND
Operating Circuit Current	I <sub>VIN</sub>	-	0.7	2	mA	VEN=3V, IOUT=0mA when no switching
<enable block=""></enable>			1	•	.1	1
EN Low Voltage	V <sub>ENL</sub>	-	-	0.3	V	
EN High Voltage	V <sub>ENH</sub>	2.2	-	VIN	V	
EN Bias Current	I <sub>EN</sub>	-	1.5	5	μA	VEN=3V
<5V Linear Regulator Block >						
VREG Standby Voltage	$V_{VREG\_STB}$	-	-	0.1	V	VEN=GND
VREG Output Voltage	$V_{VREG}$	5	5.25	5.5	V	
Maximum Current	I <sub>REG</sub>	-	10	-	mA	
< Under-Voltage Lock-Out Block >			•	1	1	-
UVLO Threshold Voltage	$V_{VREG\_UVLO}$	3.4	3.8	4.2	V	VREG: Sweep up
UVLO Hysteresis Voltage	dV <sub>VREG_UVLO</sub>	200	300	400	mV	VREG: Sweep down
< Reference Voltage Block >				L	l	1
FB Threshold Voltage1	V <sub>REF1</sub>	0.753	0.765	0.777	V	VIN=12V, VOUT=1.8V PWM Mode Operation
FB Threshold Voltage2	V <sub>REF2</sub>	0.741	0.757	0.773	V	VIN=12V, VOUT=5.0V PWM Mode Operation
FB Input Current	I <sub>FB</sub>	-	-	1	μA	
SS Charge Current	I <sub>SSC</sub>	1.4	2.0	2.6	μA	
SS Discharge Current	I <sub>SSD</sub>	0.1	0.2	-	mA	VREG=5.25V, Vss=0.5V
< On Time Control Block >						
On Time	Ton	-	215	-	nsec	VIN=12V, VOUT=1.8V
Minimum Off Time	Toffmin	100	200	-	nsec	
<sw block=""></sw>						
High Side FET ON Resistance	R <sub>ONH</sub>	-	100	200	mΩ	
Low Side FET ON Resistance	R <sub>ONL</sub>	-	70	140	mΩ	
< Over Current Protection Block >						
Over Current Protection Current Limit	locp	-	5	-	Α	(Note 5)

(Note 5) No tested on outgoing inspection.

## **Typical Performance Curves**



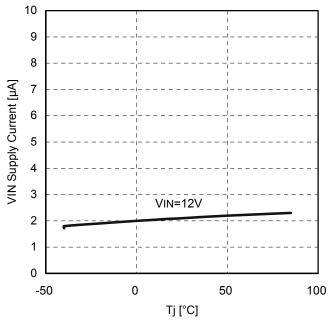


Figure 4. VIN Current vs Junction Temperature

Figure 5. VIN Shutdown Current vs Junction Temperature

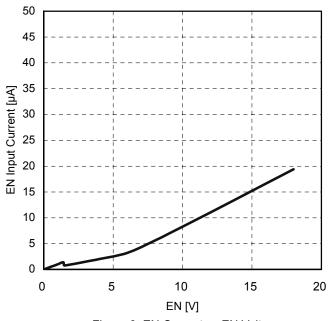


Figure 6. EN Current vs EN Voltage

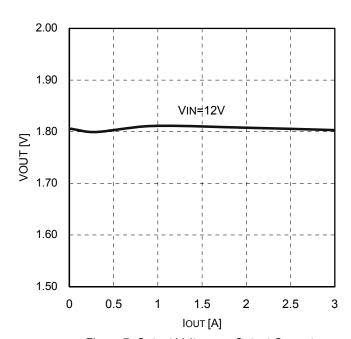
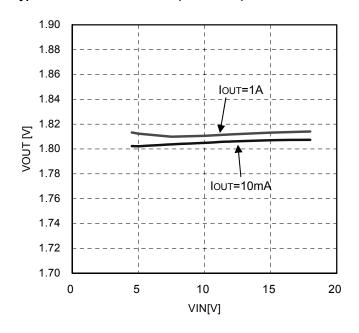


Figure 7. Output Voltage vs Output Current



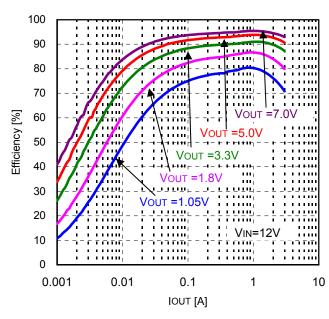


Figure 8. Output Voltage vs Input Voltage

Figure 9. Efficiency vs Output Current

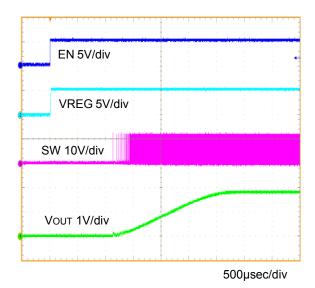


Figure 10. Start-up Waveform (EN=0V $\rightarrow$ 5V) (VIN=12V, VOUT=1.8V, IOUT=3A, Css=3300pF)

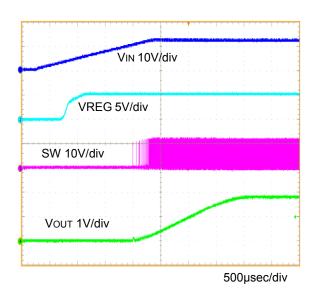
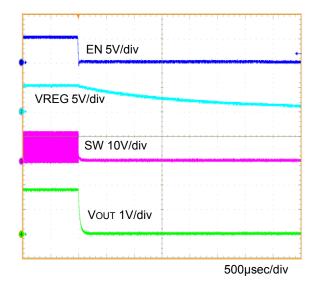


Figure 11. Start-up Waveform (VIN=EN) (VIN=12V, VOUT=1.8V, IOUT=3A, Css=3300pF)



VIN 10V/div

VREG 5V/div

SW 10V/div

Vout 1V/div

500µsec/div

Figure 12. Shutdown Waveform (EN=5V→0V) (VIN=12V, VOUT=1.8V, IOUT=3A, Css=3300pF)

Figure 13. Shutdown Waveform (VIN=EN) (VIN=12V, VOUT=1.8V, IOUT=3A, Css=3300pF)

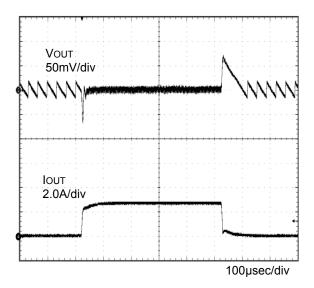


Figure 14. Load Transient Response (VIN=12V, VOUT=1.8V, IOUT=50mA to 3A)

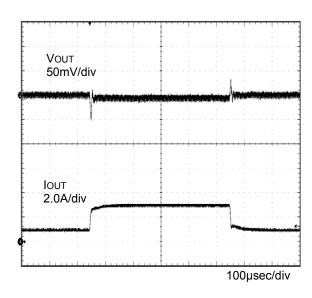


Figure 15. Load Transient Response (VIN=12V, VOUT=1.8V, IOUT=1A to 3A)

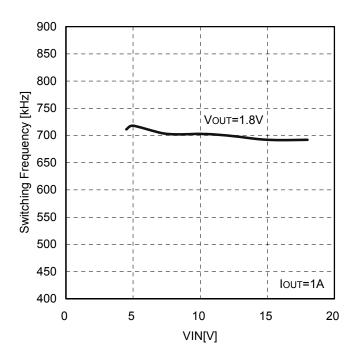


Figure 16. Switching Frequency vs Input Voltage

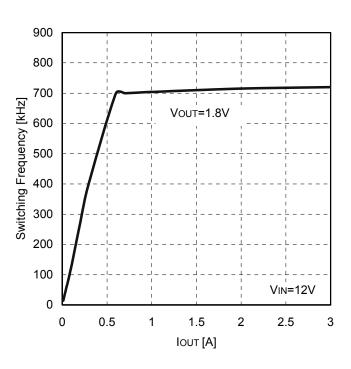


Figure 17. Switching Frequency vs Output Current

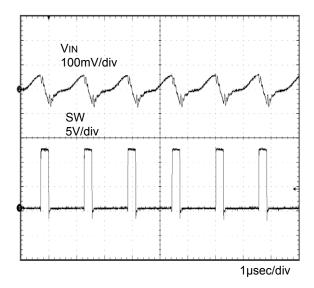


Figure 18. Voltage Ripple at Input (VIN=12V, VOUT=1.8V, IOUT=3A, L=2.2 $\mu$ H, CIN=10 $\mu$ F x 2)

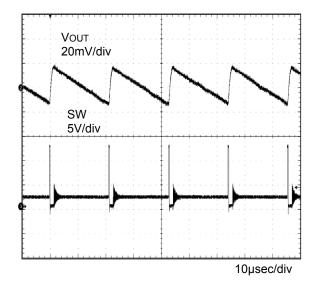


Figure 19. Voltage Ripple at Output (VIN=12V, VOUT=1.8V, IOUT=30mA, L=2.2 $\mu$ H, COUT=22 $\mu$ F x 2)

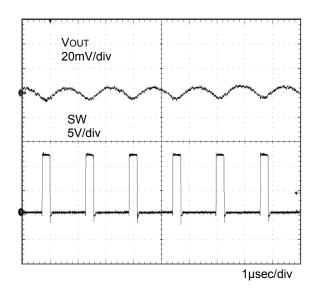


Figure 20. Voltage Ripple at Output (VIN=12V, VOUT=1.8V, IOUT=3A, L=2.2 $\mu$ H, COUT=22 $\mu$ F x 2)

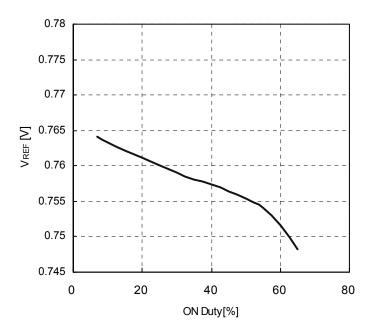


Figure 21. Reference Voltage vs ON Duty (PWM operation)

#### **Function Explanations**

## 1 Basic Operation

## 1-1 Constant On Time Control

BD9D321EFJ is a single synchronous buck switching regulator employing a constant on-time control system. It controls the on-time by using the duty ratio of Vout /VIN inside IC so that a switching frequency becomes 700 kHz. Therefore it runs with the frequency of 700 kHz under the constant on-time decided with Vout / VIN.

## 1-2 SLLM<sup>™</sup> Control

BD9D321EFJ utilizes switching operation in PWM (Pulse Width Modulation) mode for heavier load, while it utilizes SLLM (Simple Light Load Mode) control for lighter load to improve efficiency.

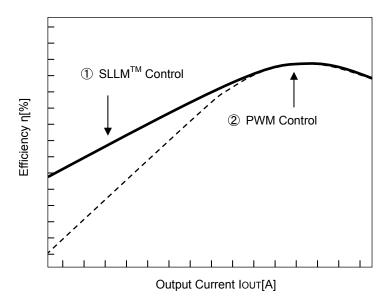


Figure 22. Efficiency (SLLM<sup>TM</sup> Control and PWM Control)

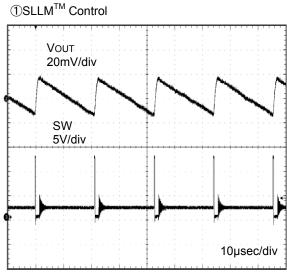


Figure 23. SW Waveform (①SLLM<sup>TM</sup> control) (VIN = 12V, VOUT = 1.8V, IOUT = 30mA)

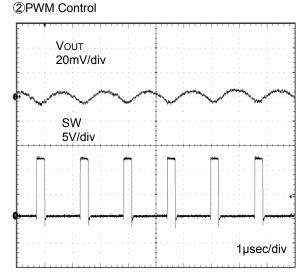


Figure 24. SW Waveform (②PWM control) (VIN = 12V, VOUT = 1.8V, IOUT = 3A)

## 1-3 Enable Control

The IC shutdown can be controlled by the voltage applied to the EN terminal. When VEN reaches 2.2 V (Typ), the internal circuit is activated and the IC starts up.

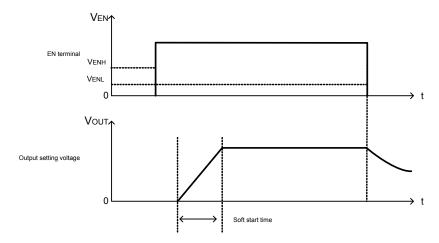


Figure 25. Start-up with EN pin

#### 1-4 Soft Start Function

By turning EN terminal to High, the soft start function operates and it gradually starts output voltage by controlling the current at start-up. Also soft start function prevents sudden current and over shoot of output voltage. Rising time can be set by connecting capacitor to SS terminal. For setting the rising time, please refer to page.17.

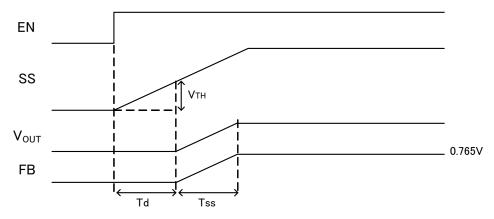


Figure 26. Soft Start Timing chart

#### 2 Protective Functions

The protective circuits are intended for prevention of damage caused by unexpected accidents. Do not use them for continuous protective operation.

## 2-1 Over Current Protection (OCP)

Over current protection function is effective by controlling current which flows in low side MOSFET by 1 cycle each of switching period. With inductor current exceeding the current restriction setting value  $I_{OCP}$  when LG is ON, the HG pulse cannot be hit even with FB voltage under REF voltage and LG continues to be ON until it is below  $I_{OCP}$ . It hits HG when it goes below  $I_{OCP}$ . As a result both frequency and duty fluctuates and output voltage may decrease. In a case where output is decreased because of OCP, output may rise after OCP is released due to the action at high speed load response. This is non-latch protection and after over current situation is released the output voltage will recover.

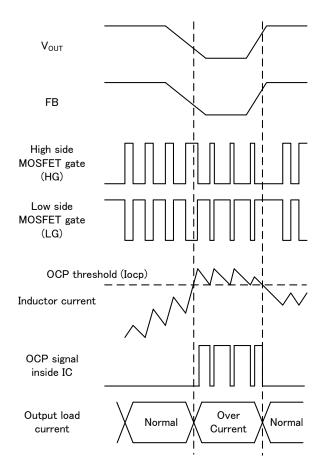


Figure 27. Over current protection timing chart

#### 2-2 Under Voltage Lockout Protection (UVLO)

The Under Voltage Lockout Protection circuit monitors the VREG terminal voltage. The operation enters standby when the VREG terminal voltage is 3.5 V (Typ) or lower. The operation starts when the VREG terminal voltage is 3.8 V (Typ) or higher.

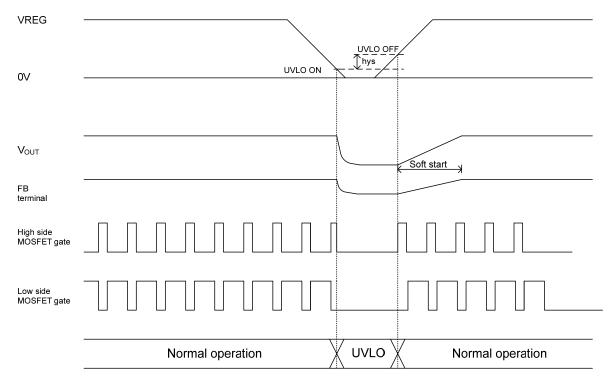


Figure 28. UVLO Timing Chart

#### **%Load at Startup**

Ensure that the respective output has light load at startup of this IC. Also, restrain the power supply line noise at startup and voltage drop generated by operating current within the hysteresis width of UVLO. Noise exceeding the hysteresis noise width may cause the IC to malfunction.

#### 2-3 Thermal Shutdown Function

When the chip temperature exceeds  $Tj = 175^{\circ}C$ , the DC/DC converter is stopped. The thermal shutdown circuit is intended for shutting down the IC from thermal runaway in an abnormal state with the temperature exceeding  $Tjmax = 150^{\circ}C$ . Do not use this function for application protection design. This is non-latch protection.

#### **Application Example**

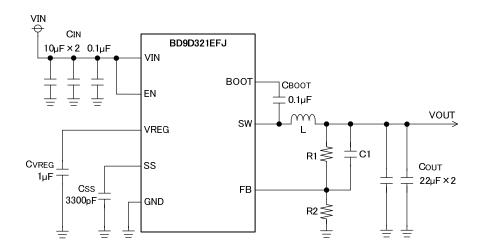


Figure 29. Application Circuit

Table 1. Recommended Component values

VIN [V]	Vout [V]	R1 [kΩ]	R2 [kΩ]	C1 [pF]	L [μH] <sup>(Note 7)</sup>
12	1.0	6.8	22	_ (Note 6)	1.5
12	1.05	8.2	22	_ (Note 6)	1.5
12	1.2	12+0.51	22	_ (Note 6)	1.5
12	1.8	30	22	_ (Note 6)	2.2
12	3.3	68+5.6	22	_ (Note 6)	2.2
12	5.0	120+3.3	22	_ (Note 6)	3.3
12	7.0	180+3.3	22	_ (Note 6)	3.3

(Note 6) C1 is a feed forward capacitor.

Additional phase boost can be achieved by adding the 5pF to 100pF capacitor (C1) in parallel with R1.

(Note 7) Recommended Inductor · ALPS GLMC series

• TDK SPM6530 series

## **Selection of Components Externally Connected**

## (1) Output LC Filter Constant

The DC/DC converter requires an LC filter for smoothing the output voltage in order to supply a continuous current to the load. Selecting an inductor with a large inductance causes the ripple current  $\Delta I_L$  that flows into the inductor to be small. However, decreasing the ripple voltage generated in the output is not advantageous in terms of the load transient response characteristic. An inductor with a small inductance improves the transient response characteristic but causes the inductor ripple current to be large which increases the ripple voltage in the output voltage, showing a trade-off relationship. The recommended inductor values are shown in Table 1.

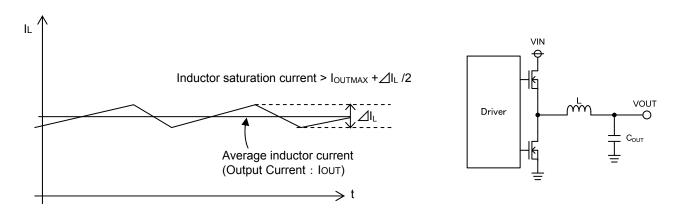


Figure 30. Waveform of current through inductor

Figure 31. Output LC filter circuit

The inductor peak to peak ripple current ∠IL is calculated using the following equation.

$$\Delta I_{L} = V_{OUT} \times (V_{IN} - V_{OUT}) \times \frac{1}{V_{IN} \times F_{OSC} \times L} [A]$$

For example, with VIN = 12 V, VOUT = 1.8 V, L =  $2.2\mu H$  and the switching frequency  $F_{OSC}$  = 700 kHz, the calculated peak current  $\Delta IL$  is 1.0A.

Then, the inductor saturation current must be larger than the sum of the maximum output current (IOUTMAX) and 1/2 of the inductor ripple current ( $\Delta L/2$ ).

The output capacitor  $C_{\text{OUT}}$  affects the output ripple voltage characteristics. The output capacitor  $C_{\text{OUT}}$  must satisfy the required ripple voltage characteristics.

The output ripple voltage can be represented by the following equation.

$$\Delta V_{RPL} = \Delta I_L \times (R_{ESR} + \frac{1}{8 \times C_{OUT} \times F_{OSC}}) [V]$$

R<sub>ESR</sub> is the Equivalent Series Resistance (ESR) of the output capacitor.

%The capacitor rating must allow a sufficient margin with respect to the output voltage. The output ripple voltage can be decreased with a smaller ESR. A ceramic capacitor of about 22 μF to 100 μF is recommended.

\*Pay attention to total capacitance value, when additional capacitor CLOAD is connected in addition to output capacitor COUT. Then, please determine CLOAD and soft start time Tss (Refer to (3) Soft Start Setting) as satisfying the following equation.

$$C_{OUT} + C_{LOAD} \le \frac{(I_{OCP} - I_{OUT}) \times T_{SS}}{V_{OUT}} [\mu F]$$

IOCP is Over Current Protection Current limit value.

#### (2) Output Voltage Setting

The output voltage value can be set by the feedback resistance ratio.

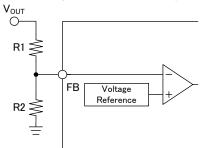


Figure 32. Feedback Resistor Circuit

$$V_{OUT} = \frac{R_1 + R_2}{R_2} \times V_{REF} \text{ [V]}$$

The VREF can be represented by the following equation defining Vout\_t as the target output voltage.

$$\begin{split} &\text{In case} \quad 0.07 \leq \frac{V_{OUT\_T}}{V_{IN}} \leq 0.5, \ \ V_{REF} = -0.02 \times \frac{V_{OUT\_T}}{V_{IN}} + 0.765 \ \ \text{[V]} \\ &\text{In case} \quad 0.5 < \frac{V_{OUT\_T}}{V_{IN}} \leq 0.65, \ \ V_{REF} = -0.22 \times \left(\frac{V_{OUT\_T}}{V_{IN}}\right)^2 + 0.2 \times \frac{V_{OUT\_T}}{V_{IN}} + 0.7105 \ \ \text{[V]} \end{split}$$

BD9D321EFJ can operate under the condition which satisfies the following equation.

$$0.07 \le \frac{V_{OUT}}{V_{IN}} \le 0.65$$

## 3) Soft Start Setting

Turning the EN terminal signal High activates the soft start function. This causes the output voltage to rise gradually while the current at startup is placed under control. This allows the prevention of output voltage overshoot and inrush current. The rise time depends on the value of the capacitor connected to the SS terminal.

$$T_d = (C_{SS} \times V_{TH})/I_{SS}$$
  
$$T_{SS} = (C_{SS} \times V_{FB} \times 1.15)/I_{SS}$$

where

 $T_d$  is Soft Start Delay Time

 $T_{\rm ss}$  is Soft Start Time

 $\mathcal{C}_{ss}$  is Capacitor connected to Soft Start Time Terminal

 $V_{FB}$  is FB Terminal Voltage(0.765V Typ)

 $V_{TH}$  is Internal MOS threshold voltage(0.7V Typ)

 $I_{SS}$  is Soft Start Terminal Source Current(2.0 $\mu$ A Typ)

with Css = 3300pF,

$$T_d = (3300 \, [pF] \times 0.7 \, [V]) / 2.0 \, [\mu A]$$

= 1.16 [msec]

 $T_{SS} = (3300 [pF] \times 0.765 [V] \times 1.15) / 2.0 [\mu A]$ 

=1.45[msec]

#### **PCB Layout Design**

In the step-down DC/DC converter, a large pulse current flows into two loops. The first loop is the one into which the current flows when the high side FET is turned ON. The flow starts from the input capacitor  $C_{\text{IN}}$ , runs through the FET, inductor L and output capacitor  $C_{\text{OUT}}$  and back to ground of  $C_{\text{IN}}$  via ground of  $C_{\text{OUT}}$ . The second loop is the one into which the current flows when the low side FET is turned on. The flow starts from the low side FET, runs through the inductor L and output capacitor  $C_{\text{OUT}}$  and back to ground of the low side FET via ground of  $C_{\text{OUT}}$ . Route these two loops as thick and as short as possible to allow noise to be reduced for improved efficiency. It is recommended to connect the input and output capacitors directly to the ground plane. The PCB layout has a great influence on the DC/DC converter in terms of all of the heat generation, noise and efficiency characteristics.

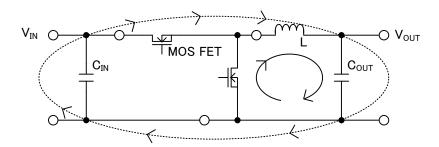
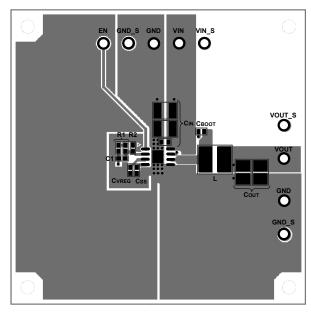
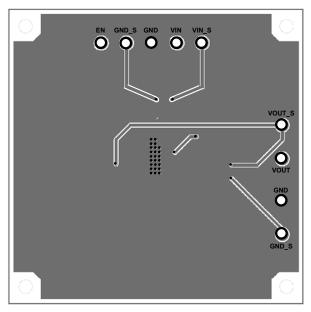


Figure 33. Current Loop of Buck Converter

Accordingly, design the PCB layout considering the following points.

- Connect an input capacitor as close as possible to the IC VIN terminal on the same plane as the IC.
- If there is any unused area on the PCB, provide a copper foil plane for the ground node to assist heat dissipation from the IC and the surrounding components.
- Switching nodes such as SW are susceptible to noise due to AC coupling with other nodes. Route the coil pattern as thick and as short as possible.
- Provide lines connected to FB and SS far from the SW nodes.
- Place the output capacitor away from the input capacitor in order to avoid the effect of harmonic noise from the input.





**TOP Layer** 

**Bottom Layer** 

Figure 34. Example of PCB layout

## **Power Dissipation**

When designing the PCB layout and peripheral circuitry, sufficient consideration must be given to ensure that the power dissipation is within the allowable dissipation curve.

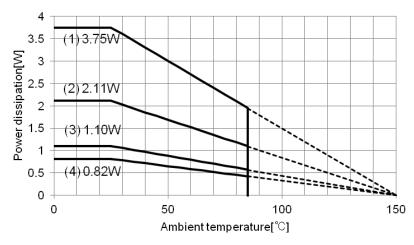


Figure 35. Power dissipation (HTSOP-J8)

## HTSOP-J8 Package

 $70\times70\times1.6$  mm assembled glass epoxide board

- (1) 4-layer board (Copper foil area 70 mm × 70 mm)
- (2) 2-layer board (Copper foil area 70 mm × 70 mm)
- (3) 2-layer board (Copper foil area 15 mm × 15 mm)
- (4) 1-layer board (Copper foil area 0 mm× 0 mm)

## I/O Equivalent Circuit

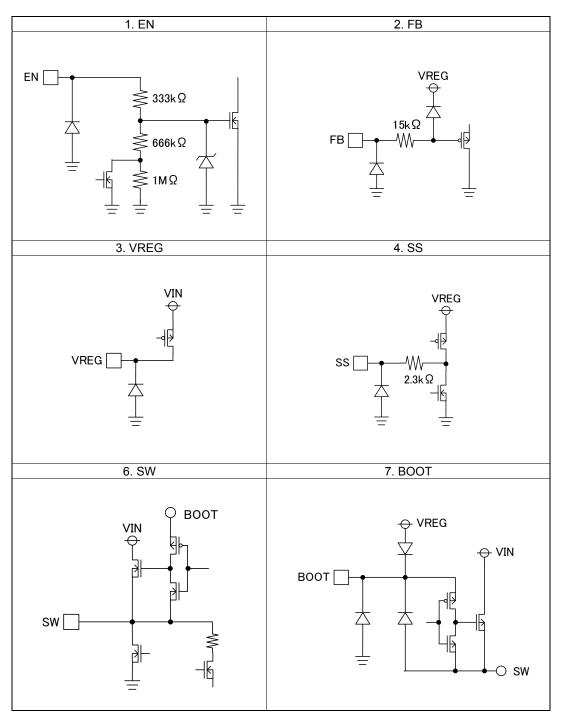


Figure 36. I/O equivalence circuit

#### **Operational Notes**

#### 1. Reverse Connection of Power Supply

Connecting the power supply in reverse polarity can damage the IC. Take precautions against reverse polarity when connecting the power supply, such as mounting an external diode between the power supply and the IC's power supply terminals.

#### 2. Power Supply Lines

Design the PCB layout pattern to provide low impedance supply lines. Separate the ground and supply lines of the digital and analog blocks to prevent noise in the ground and supply lines of the digital block from affecting the analog block. Furthermore, connect a capacitor to ground at all power supply pins. Consider the effect of temperature and aging on the capacitance value when using electrolytic capacitors.

#### 3. Ground Voltage

Ensure that no pins are at a voltage below that of the ground pin at any time, even during transient condition.

#### 4. Ground Wiring Pattern

When using both small-signal and large-current ground traces, the two ground traces should be routed separately but connected to a single ground at the reference point of the application board to avoid fluctuations in the small-signal ground caused by large currents. Also ensure that the ground traces of external components do not cause variations on the ground voltage. The ground lines must be as short and thick as possible to reduce line impedance.

#### 5. Thermal Consideration

Should by any chance the power dissipation rating be exceeded the rise in temperature of the chip may result in deterioration of the properties of the chip. The absolute maximum rating of the Pd stated in this specification is when the IC is mounted on 4 - layer 70mm x 70mm x 1.6mm glass epoxy board. In case of exceeding this absolute maximum rating, increase the board size and copper area to prevent exceeding the Pd rating.

#### 6. Recommended Operating Conditions

These conditions represent a range within which the expected characteristics of the IC can be approximately obtained. The electrical characteristics are guaranteed under the conditions of each parameter.

#### 7. Rush Current

When power is first supplied to the IC, it is possible that the internal logic may be unstable and inrush current may flow instantaneously due to the internal powering sequence and delays, especially if the IC has more than one power supply. Therefore, give special consideration to power coupling capacitance, power wiring, width of ground wiring, and routing of connections.

## 8. Operation Under Strong Electromagnetic Field

Operating the IC in the presence of a strong electromagnetic field may cause the IC to malfunction.

## 9. Testing on Application Boards

When testing the IC on an application board, connecting a capacitor directly to a low-impedance output pin may subject the IC to stress. Always discharge capacitors completely after each process or step. The IC's power supply should always be turned off completely before connecting or removing it from the test setup during the inspection process. To prevent damage from static discharge, ground the IC during assembly and use similar precautions during transport and storage.

## 10. Inter-pin Short and Mounting Errors

Ensure that the direction and position are correct when mounting the IC on the PCB. Incorrect mounting may result in damaging the IC. Avoid nearby pins being shorted to each other especially to ground, power supply and output pin. Inter-pin shorts could be due to many reasons such as metal particles, water droplets (in very humid environment) and unintentional solder bridge deposited in between pins during assembly to name a few.

## Operational Notes - continued

#### 11. Regarding the Input Pin of the IC

This monolithic IC contains P+ isolation and P substrate layers between adjacent elements in order to keep them isolated. P-N junctions are formed at the intersection of the P layers with the N layers of other elements, creating a parasitic diode or transistor. For example (refer to figure below):

When GND > Pin A and GND > Pin B, the P-N junction operates as a parasitic diode.

When GND > Pin B, the P-N junction operates as a parasitic transistor.

Parasitic diodes inevitably occur in the structure of the IC. The operation of parasitic diodes can result in mutual interference among circuits, operational faults, or physical damage. Therefore, conditions that cause these diodes to operate, such as applying a voltage lower than the GND voltage to an input pin (and thus to the P substrate) should be

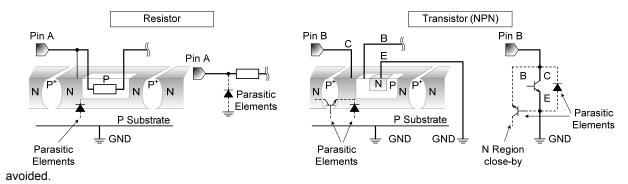


Figure 37. Example of monolithic IC structure

#### 12. Ceramic Capacitor

When using a ceramic capacitor, determine the dielectric constant considering the change of capacitance with temperature and the decrease in nominal capacitance due to DC bias and others.

#### 13. Thermal Shutdown Circuit(TSD)

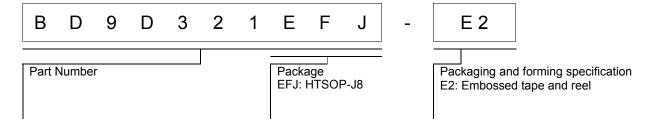
This IC has a built-in thermal shutdown circuit that prevents heat damage to the IC. Normal operation should always be within the IC's power dissipation rating. If however the rating is exceeded for a continued period, the junction temperature (Tj) will rise which will activate the TSD circuit that will turn OFF all output pins. When the Tj falls below the TSD threshold, the circuits are automatically restored to normal operation.

Note that the TSD circuit operates in a situation that exceeds the absolute maximum ratings and therefore, under no circumstances, should the TSD circuit be used in a set design or for any purpose other than protecting the IC from heat damage.

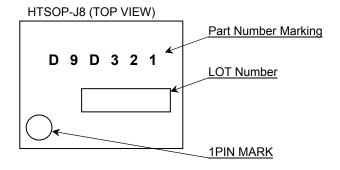
## 14. Over Current Protection Circuit (OCP)

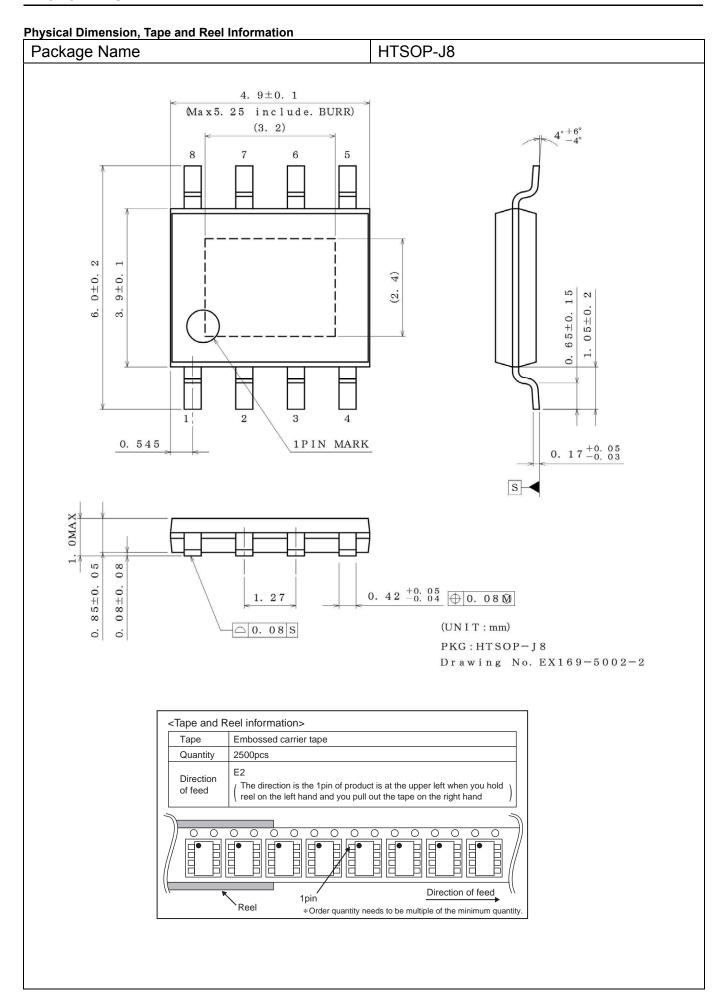
This IC incorporates an integrated overcurrent protection circuit that is activated when the load is shorted. This protection circuit is effective in preventing damage due to sudden and unexpected incidents. However, the IC should not be used in applications characterized by continuous operation or transitioning of the protection circuit.

## **Ordering Information**



## **Marking Diagram**





## **Revision History**

Date	Revision	Changes
07.Aug.2013	001	Created
29.Jan.2015	002	Revised the Electrical Characteristics and Table1. Added Figure 21.

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CLASSIV	CLASSⅢ	CLASSⅢ	- CLASSIII

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