

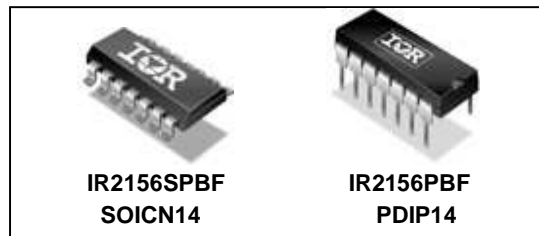
Features

- Ballast control and half bridge driver in one IC
- Programmable preheat frequency
- Programmable preheat time
- Internal ignition ramp
- Programmable over-current threshold
- Programmable run frequency
- Programmable dead time
- DC bus under-voltage reset
- Shutdown pin with hysteresis
- Internal 15.6V zener clamp diode on Vcc
- Micropower startup (150 μ A)
- Latch immunity and ESD protection

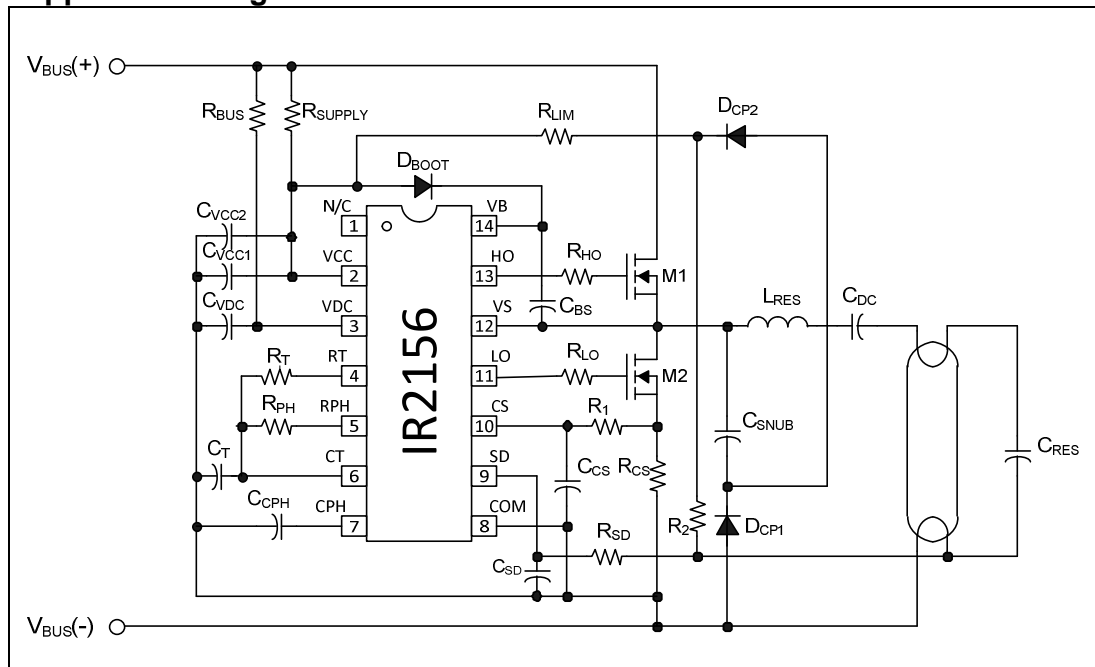
Description

The IR2156 incorporates a high voltage half-bridge gate driver with a programmable oscillator and state diagram to form a complete ballast control IC. The IR2156 features include programmable preheat and run frequencies, programmable preheat time, programmable dead-time, and programmable over-current protection. Comprehensive protection features such as protection from failure of a lamp to strike, filament failures, as well as an automatic restart function, have been included in the design.

Packages



Application Diagram



Absolute Maximum Ratings

Absolute Maximum Ratings indicate sustained limits beyond which damage to the device may occur. All voltage parameters are absolute voltages referenced to COM, all currents are defined positive into any lead. The Thermal Resistance and Power Dissipation ratings are measured under board mounted and still air conditions.

| Symbol | Definition | Min. | Max. | Units | |
|---------------|---|---------------|-----------|------------------|--------------------|
| VB | High side floating supply voltage | -0.3 | 625 | V | |
| VS | High side floating supply offset voltage | VB - 25 | VB + 0.3 | | |
| VHO | High side floating output voltage | VS - 0.3 | VB + 0.3 | | |
| VLO | Low side output voltage | -0.3 | VCC + 0.3 | | |
| IOMAX | Maximum allowable output current (HO, LO) due to external power transistor miller effect | -500 | 500 | mA | |
| VDC | VDC pin voltage | -0.3 | VCC+0.3 | V | |
| VCT | CT pin voltage | -0.3 | VCC+0.3 | | |
| VCPH | CPH pin voltage | -0.3 | VCC+0.3 | | |
| ICPH | CPH pin current | -5 | 5 | mA | |
| IRPH | RPH pin current | -5 | 5 | | |
| VRPH | RPH pin voltage | -0.3 | VCC+0.3 | V | |
| IRT | RT pin current | -5 | 5 | mA | |
| VRT | RT pin voltage | -0.3 | VCC+0.3 | V | |
| VCS | Current sense pin voltage | -0.3 | 5.5 | | |
| ICS | Current sense pin current | -5 | 5 | mA | |
| ISD | Shutdown pin current | -5 | 5 | | |
| ICC | Supply current (Note 1) | -20 | 20 | | |
| dV/dt | Allowable offset voltage slew rate | -50 | 50 | V/ns | |
| PD | Package power dissipation @ $T_A \leq +25^\circ\text{C}$ $PD = (T_{JMAX} - T_A) / R_{\theta JA}$ | (14-Pin DIP) | --- | 1.80 | W |
| | | (14-Pin SOIC) | --- | 1.40 | |
| R θ JA | Thermal resistance, junction to ambient | (14-Pin DIP) | --- | 70 | $^\circ\text{C/W}$ |
| | | (14-Pin SOIC) | --- | 82 | |
| TJ | Junction temperature | -55 | 150 | $^\circ\text{C}$ | |
| TS | Storage temperature | -55 | 150 | | |
| TL | Lead temperature (soldering, 10 seconds) | --- | 300 | | |

Note 1: This IC contains a zener clamp structure between the chip VCC and COM which has a nominal breakdown voltage of 15.6V. Please note that this supply pin should not be driven by a DC, low impedance power source greater than the VCLAMP specified in the Electrical Characteristics section.

Recommended Operating Conditions

For proper operation the device should be used within the recommended conditions.

| Symbol | Definition | Min. | Max. | Units |
|--------|---|--------|--------|-------|
| VBS | High side floating supply voltage | VBSUV+ | VCLAMP | V |
| VS | Steady state high side floating supply offset voltage | -1 | 600 | |
| VCC | Supply voltage | VCCUV+ | VCLAMP | |
| ICC | Supply current | Note 2 | 10 | mA |
| CT | CT lead capacitance | 220 | --- | pF |
| ISD | Shutdown lead current | -1 | 1 | mA |
| ICS | Current sense pin current | -1 | 1 | |
| TJ | Junction temperature | -25 | 125 | °C |

Note 2: Enough current should be supplied into the VCC pin to keep the internal 15.6V zener clamp diode on this pin regulating at its voltage, VCLAMP.

Electrical Characteristics

VCC = VBS = VBIAS = 14V +/- 0.25V, VVDC=Open, RT=40KΩ, RPH=100KΩ, CT=470 pF, VCPH=0.0V, VSD=0.0V, VCS=0.0V, CLO=CHO=1000 pF, TA=25C unless otherwise specified.

| Symbol | Definition | Min | Typ | Max | Units | Test Conditions |
|--|--|------|------|------|-------|--|
| Supply Characteristics | | | | | | |
| VCCUV+ | VCC supply undervoltage positive going threshold | 10.5 | 11.5 | 12.5 | V | VCC rising from 0V |
| VCCUV- | VCC supply undervoltage negative going threshold | 8.5 | 9.5 | 10.5 | | VCC falling from 14V |
| VUVHYS | VCC supply undervoltage lockout hysteresis | 1.5 | 2.0 | 3.0 | | |
| IQCCUV | UVLO mode quiescent current | 50 | 120 | 200 | μA | VCC=11V |
| IQCCFLT | Fault-mode quiescent current | --- | 200 | 470 | | SD = 5.1V, or CS = 1.3V |
| IQCC | Quiescent VCC supply current | --- | 1.0 | 1.5 | mA | CT connected to COM, VCC = 14V, RT = 15kΩ |
| ICC40k | VCC supply current, f = 40kHz | 1.3 | 1.5 | 1.7 | | VCPH=12V, VVDC=12V |
| VCLAMP | VCC zener clamp voltage | 14.5 | 15.6 | 16.5 | V | ICC = 5mA |
| Floating Supply Characteristics | | | | | | |
| IQBS0 | Quiescent VBS supply current | -5 | 0 | 5 | μA | VHO = VS (CT=0V) |
| IQBS1 | Quiescent VBS supply current | --- | 30 | 50 | | VHO = VB (CT=14V) |
| ILK | Offset supply leakage current | --- | --- | 50 | μA | VB = VS = 600V |

Electrical Characteristics

VCC = VBS = VBIAS = 14V +/- 0.25V, VVDC=Open, RT=40KΩ, RPH=100KΩ, CT=470 pF, VCPH=0.0V, VSD=0.0V, VCS=0.0V, CLO=CHO=1000 pF, TA=25C unless otherwise specified.

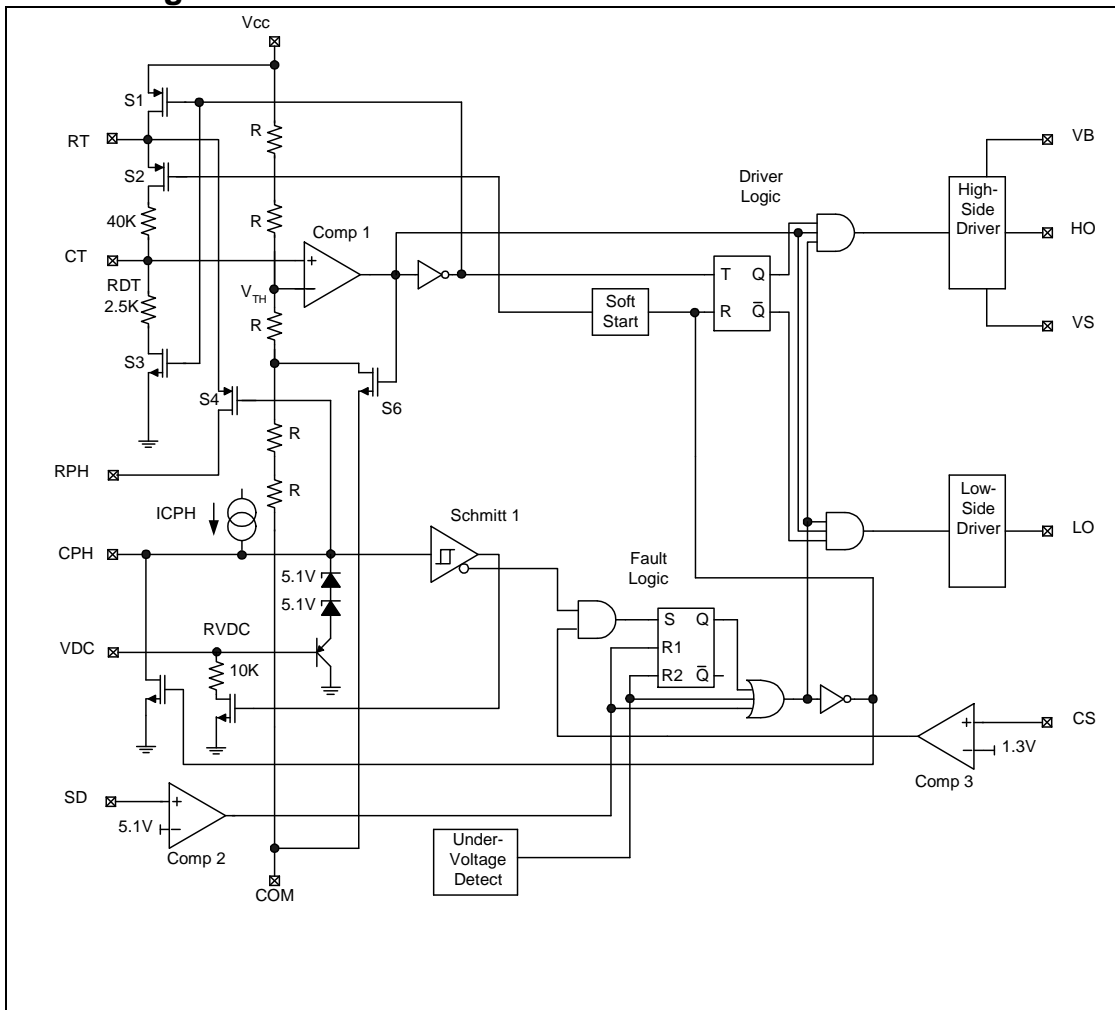
| Symbol | Definition | Min | Typ | Max | Units | Test Conditions |
|---|--|------|------|------|-------|--|
| Oscillator, Ballast Control, I/O Characteristics | | | | | | |
| f _{OSCRUN} | Oscillator frequency during RUN mode | 36.0 | 40.0 | 44.0 | kHz | V _{VDC} =14V, V _{CPH} =Open |
| f _{OSCPH} | Oscillator frequency during PH mode | 49.0 | 55.0 | 60.0 | | V _{VDC} =14V, V _{CPH} =COM |
| d | Oscillator duty cycle | --- | 50 | --- | % | |
| VCT+ | Upper CT ramp voltage threshold | --- | 8.3 | --- | V | V _{CC} =14V SD>5.1V or CS>1.3V |
| VCT- | Lower CT ramp voltage threshold | --- | 4.8 | --- | | |
| VCTFLT | Fault-mode CT pin voltage | --- | 0 | --- | | |
| t _{DLO} | LO output deadtime | --- | 2.0 | --- | μsec | |
| t _{DHO} | HO output deadtime | --- | 2.0 | --- | | |
| RDT | Internal deadtime resistor | --- | 3 | --- | kΩ | |
| Preheat Characteristics | | | | | | |
| I _{CPH} | CPH pin charging current | 3.6 | 4.3 | 5.2 | μA | CT=10V, VDC=5V, VCPH=0V |
| V _{CPHFLT} | Fault-mode CPH pin voltage | --- | 0 | --- | mV | SD>5.1V or CS>1.3V |
| RPH Characteristics | | | | | | |
| I _{RPHLK} | Open circuit RPH pin leakage current | --- | 0.1 | --- | μA | CT=10V |
| V _{RPHFLT} | Fault-mode RPH pin voltage | --- | 0 | --- | mV | SD>5.1V or CS>1.3V |
| RT Characteristics | | | | | | |
| I _{RTLK} | Open circuit RT pin leakage current | --- | 0.1 | --- | μA | CT=10V |
| V _{RTFLT} | Fault-mode RT pin voltage | --- | 0 | --- | mV | SD>5.1V or CS>1.3V |
| Protection Circuitry Characteristics | | | | | | |
| V _{SDTH+} | Rising shutdown pin threshold voltage | --- | 5.1 | --- | V | |
| V _{SDHYS} | SD pin Reset threshold voltage | --- | 450 | --- | mV | |
| V _{CSTH+} | Over-current sense threshold voltage | 1.1 | 1.25 | 1.44 | V | |
| t _{CS} | Over-current sense propagation delay | --- | 160 | --- | nsec | Delay from CS to LO |
| V _{CSPW} | Over-current sense minimum pulse width | --- | 135 | --- | | VCS pulse amplitude = V _{CSTH} + 100mV |
| R _{VDC} | DC bus sensing resistor | 7.5 | 10.0 | 14.0 | kΩ | V _{CPH} >12V, VDC=7V, CT=0 |
| V _{CPH-VDC} | CPH to VDC offset voltage | 10.3 | 10.9 | 11.4 | V | V _{CPH} open, VDC=0V |

Electrical Characteristics

VCC = VBS = VBIAS = 14V +/- 0.25V, VVDC=Open, RT=39KΩ, RPH=100KΩ, CT=470 pF, VCPH=0.0V, VSD=0.0V, VCS=0.0V, CLO=CHO=1000 pF, TA=25C unless otherwise specified.

| Symbol | Definition | Min | Typ | Max | Units | Test Conditions |
|---|---------------------------|-----|-----|-----|-------|-----------------|
| Gate Driver Output Characteristics | | | | | | |
| VOL | Low-level output voltage | --- | COM | --- | V | IO = 0 |
| VOH | High-level output voltage | --- | VCC | --- | | IO = 0 |
| t _r | Turn-on rise time | --- | 110 | 150 | nsec | |
| t _f | Turn-off fall time | --- | 55 | 100 | | |

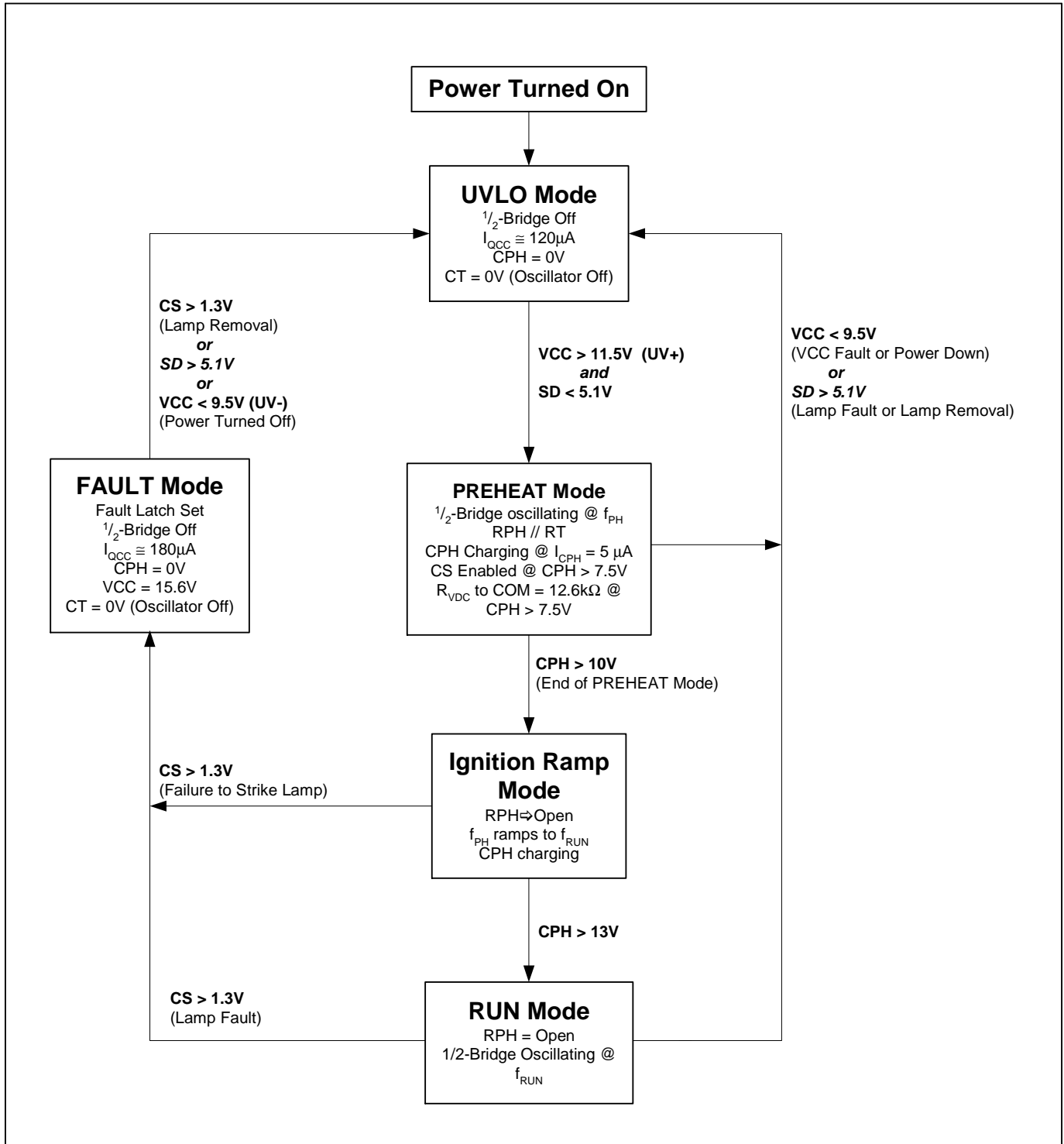
Block Diagram



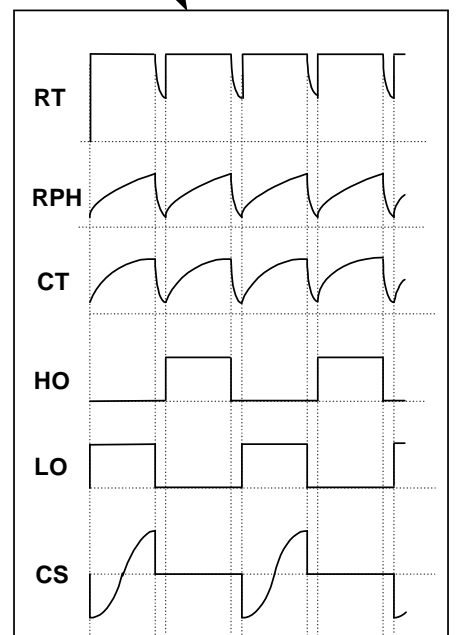
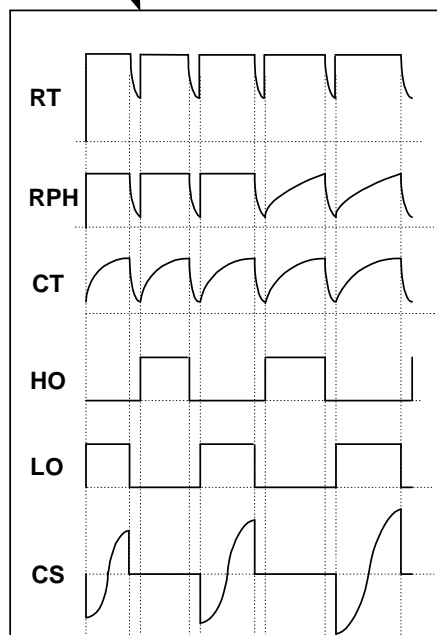
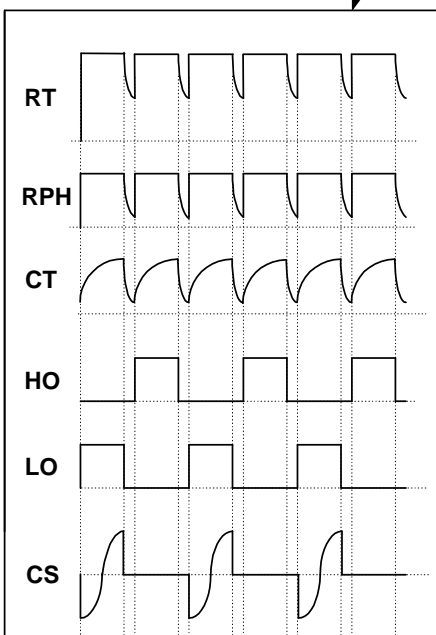
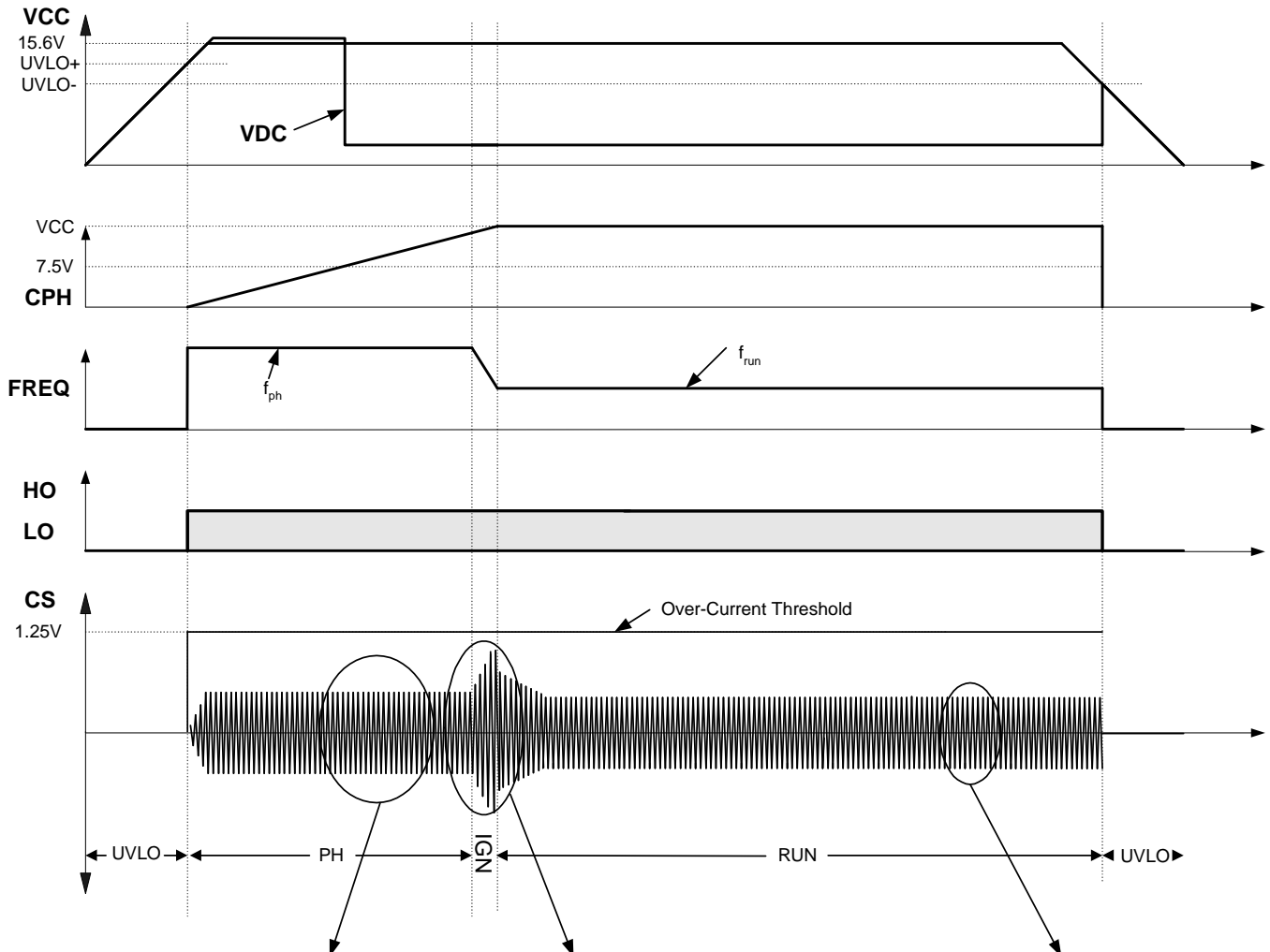
Pin Assignments & Definitions

| Pin Assignments | | Pin # | Symbol | Description |
|-----------------|---|-------|--------|---------------------------------------|
| NC | 1 | 14 | VB | No Connect |
| VCC | 2 | 13 | HO | Logic & Low-Side Gate Driver Supply |
| VDC | 3 | 12 | VS | IC Start-up and DC Bus Sensing Input |
| RT | 4 | 11 | LO | Minimum Frequency Timing Resistor |
| RPH | 5 | 10 | CS | Preheat Frequency Timing Resistor |
| CT | 6 | 9 | SD | Oscillator Timing Capacitor |
| CPH | 7 | 8 | COM | Preheat Timing Capacitor |
| | | | | IC Power & Signal Ground |
| | | | | Shutdown Input |
| | | | | Current Sensing Input |
| | | | | Low-Side Gate Driver Output |
| | | | | High-Side Floating Return |
| | | | | High-Side Gate Driver Output |
| | | | | High-Side Gate Driver Floating Supply |

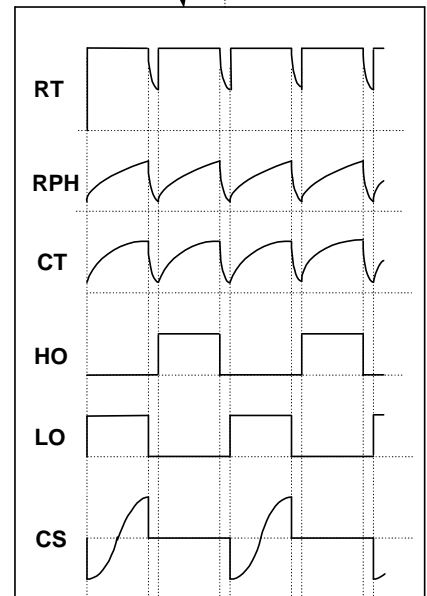
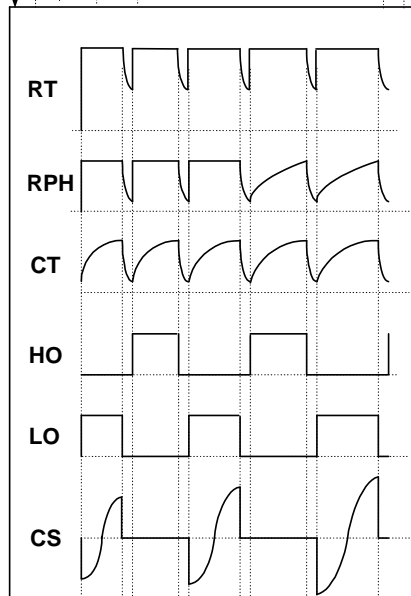
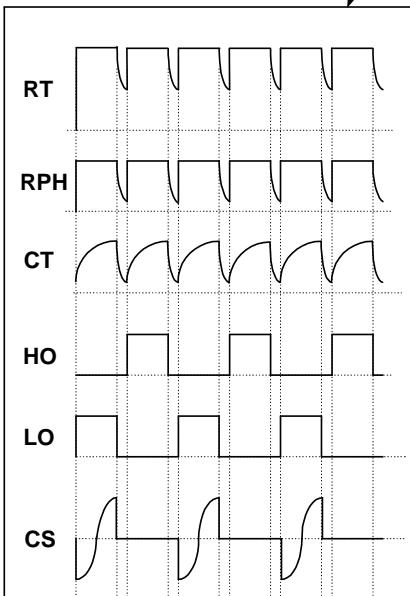
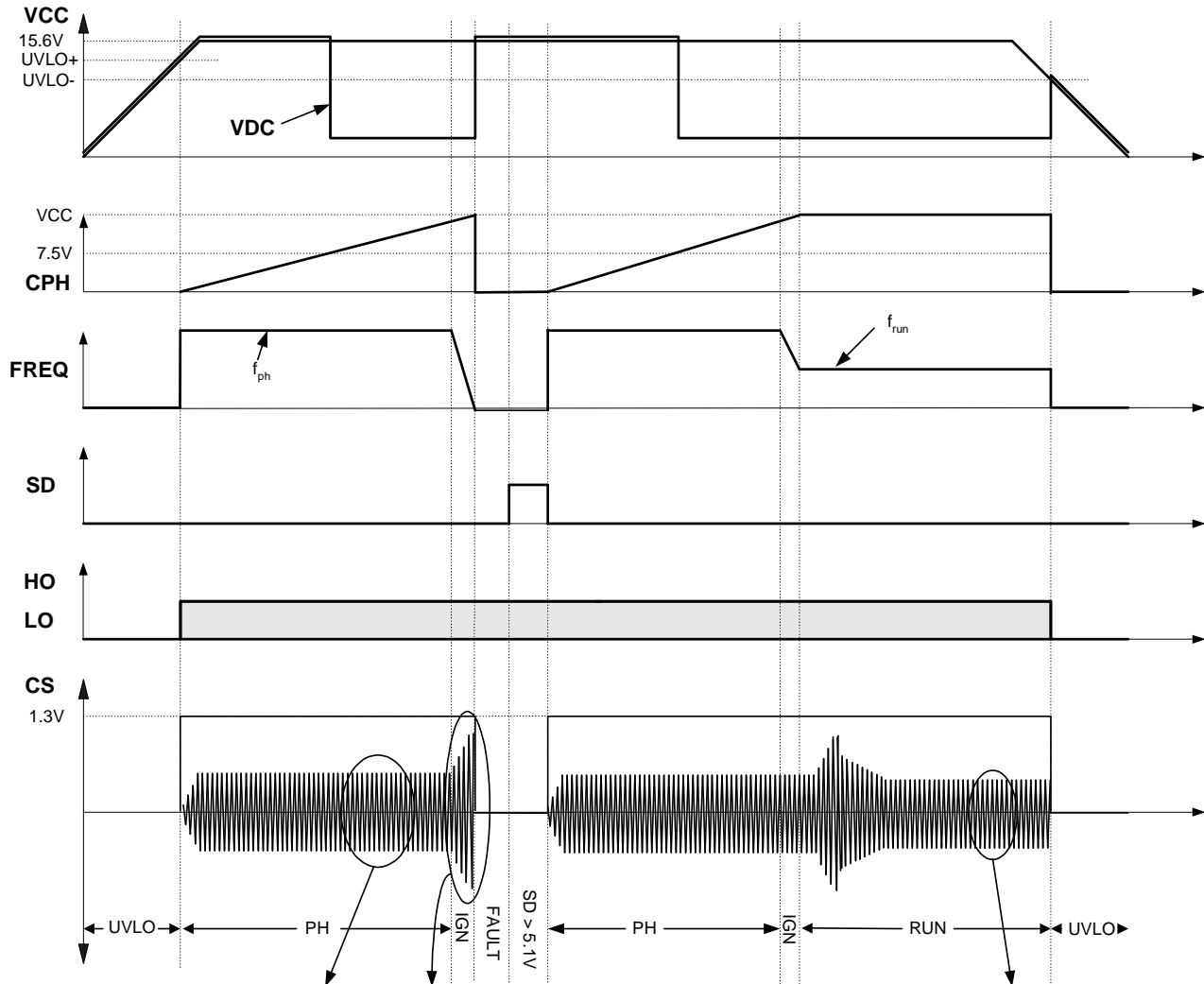
State Diagram



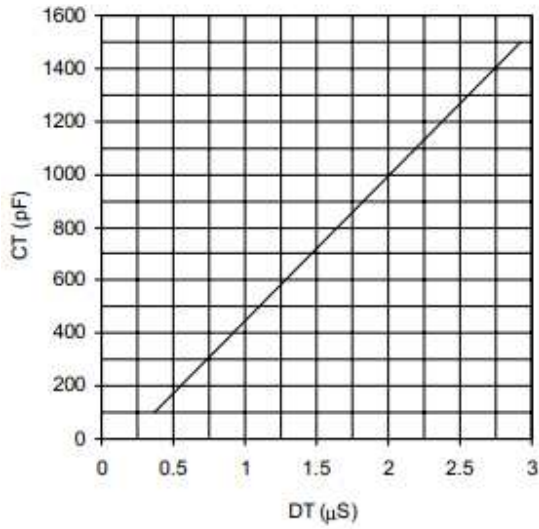
TIMING DIAGRAMS NORMAL OPERATION



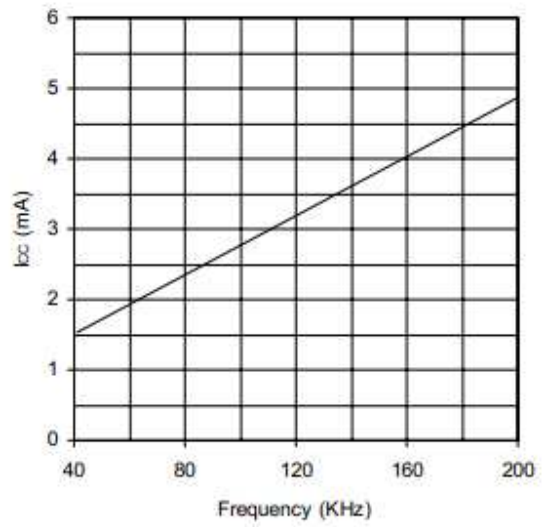
TIMING DIAGRAMS FAULT CONDITION



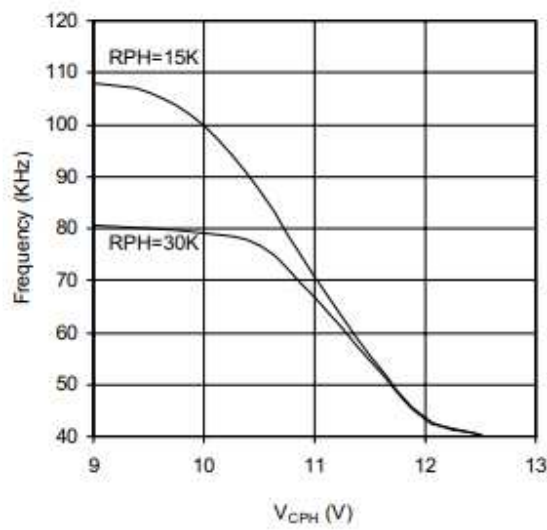
Characterization Data



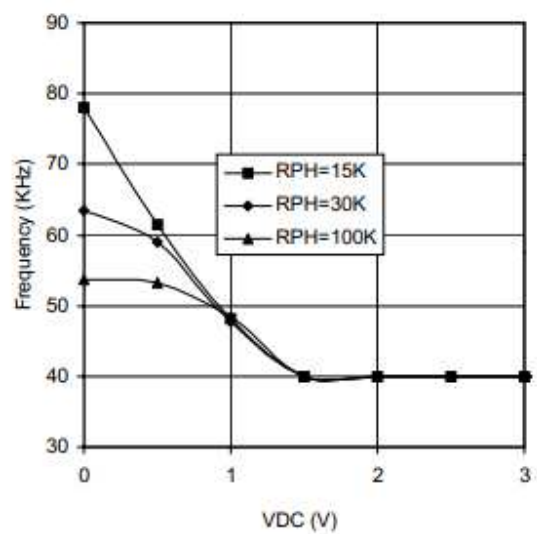
Graph 1. CT vs Dead Time (IR2156)



Graph 2. I_{cc} vs Frequency (IR2156)

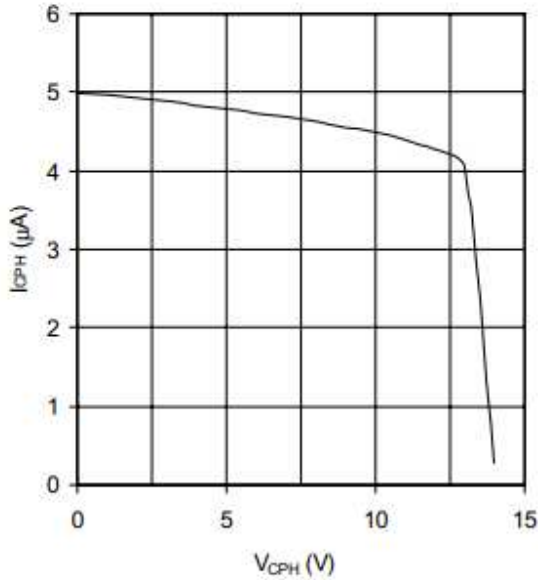


Graph 3. Frequency vs V_{CPH} (IR2156)

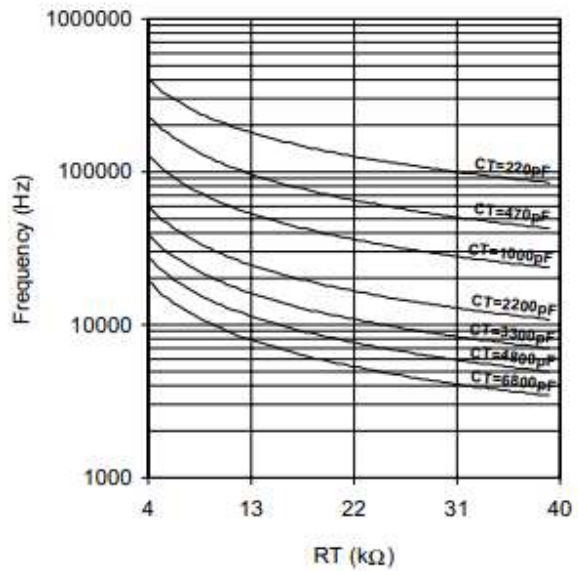


Graph 4. Frequency vs V_{DC} (IR2156)

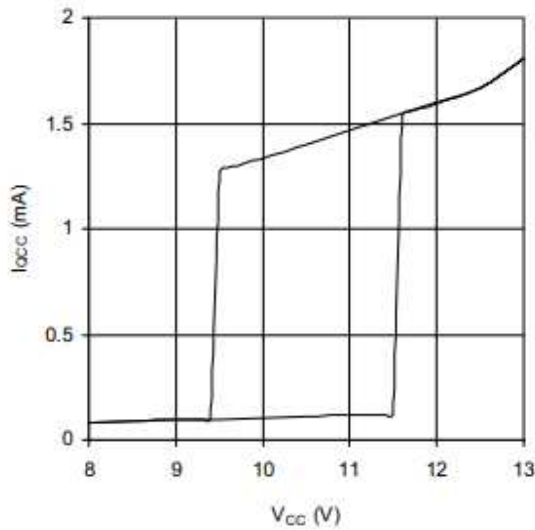
Characterization Data



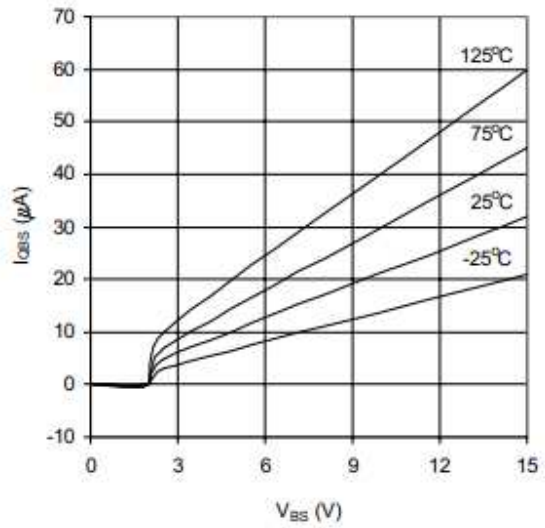
Graph 5. I_{CPH} vs V_{CPH} (IR2156)



Graph 6. Frequency vs RT (IR2156)

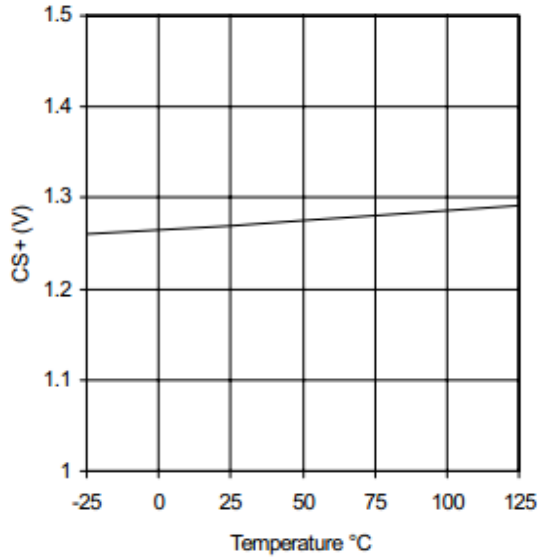


Graph 7. I_{QCC} vs V_{CC} (IR2156)
 UVLO Hysteresis

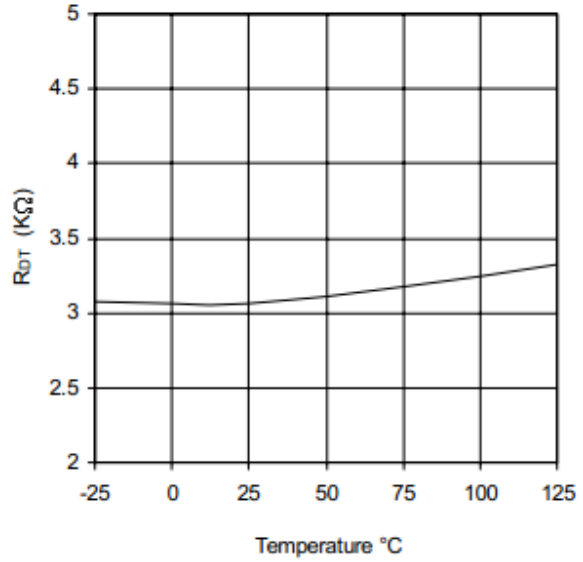


Graph 8. I_{QBS} vs V_{CC} vs Temp (IR2156)

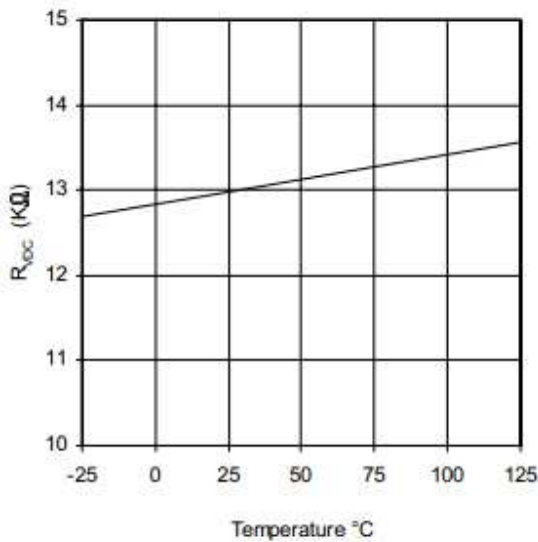
Characterization Data



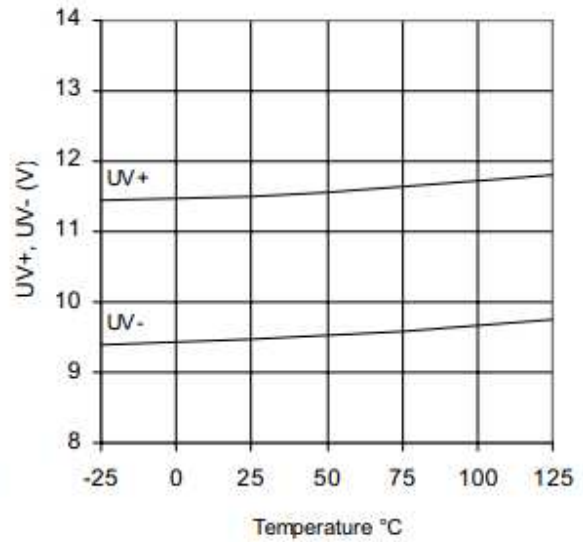
Graph 9. VCSTH+ vs Temperature (IR2156)



Graph 10. RDT vs Temperature (IR2156)

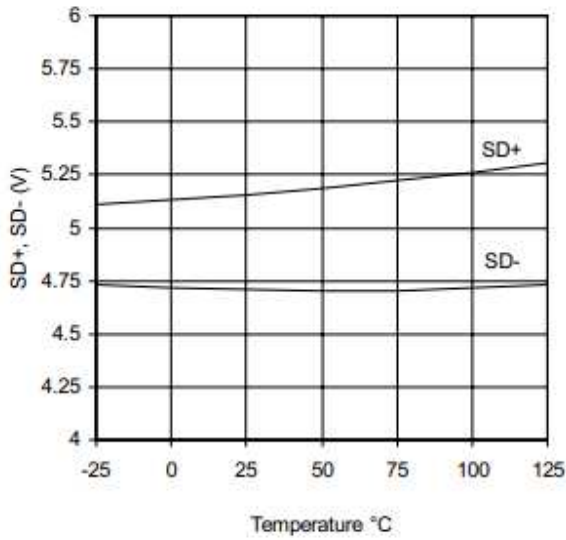


Graph 11. RvDC+ vs Temperature (IR2156)

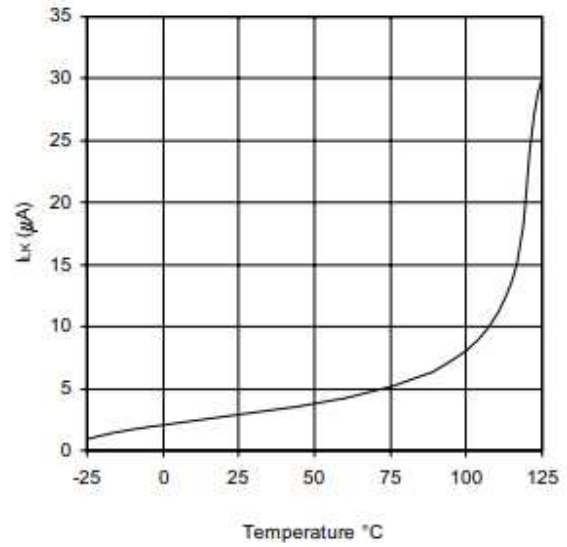


Graph 12. UV+, UV- vs Temperature (IR2156)

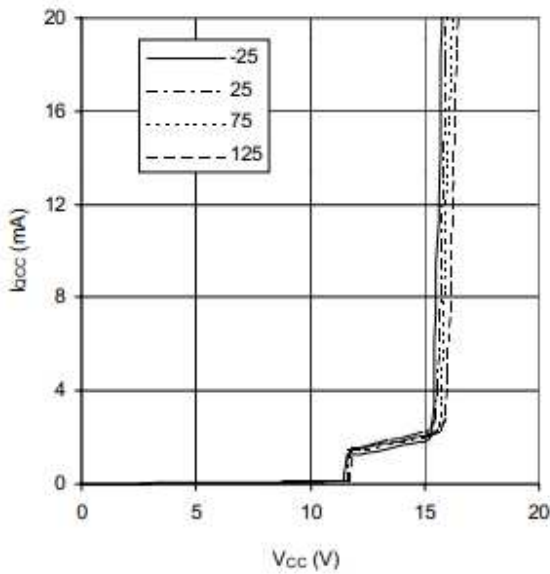
Characterization Data



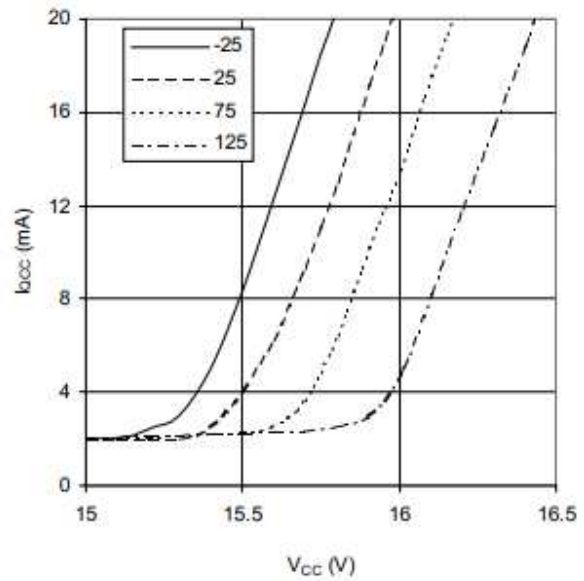
Graph 13. SD+, SD- vs Temperature (IR2156)



Graph 14. I_{LK} vs Temperature (IR2156)

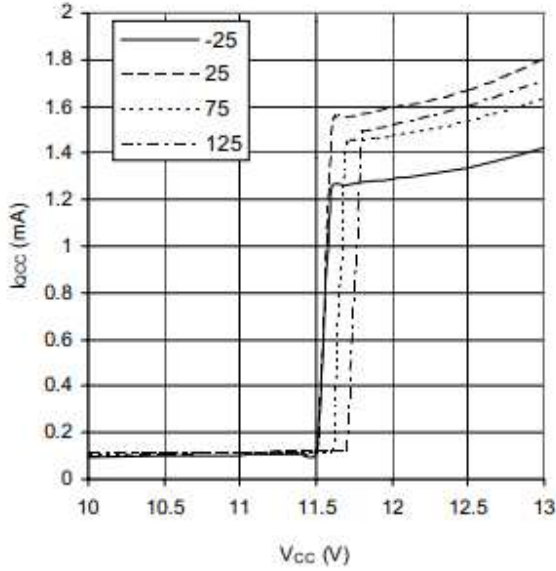


Graph 15. I_{QCC} vs V_{CC} vs Temperature (IR2156)

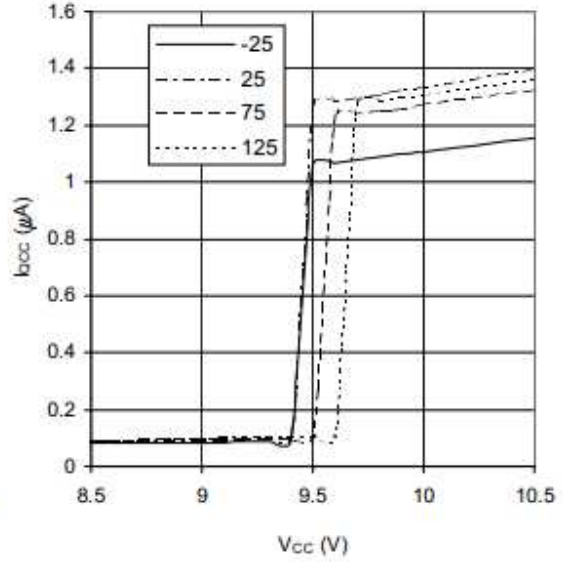


Graph 16. I_{QCC} vs V_{CC} vs Temperature (IR2156)
 Internal Zener Diode Curve

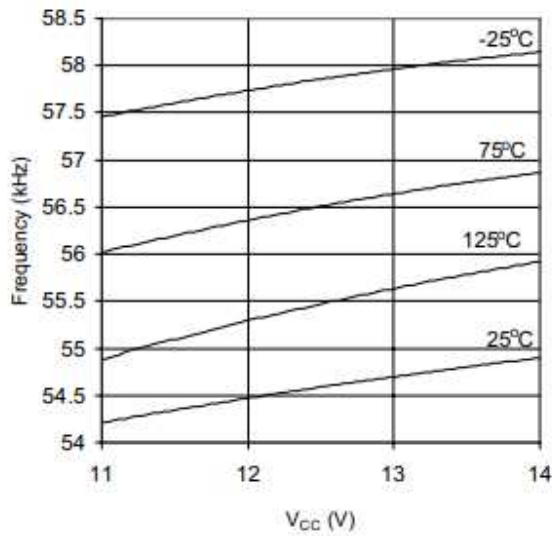
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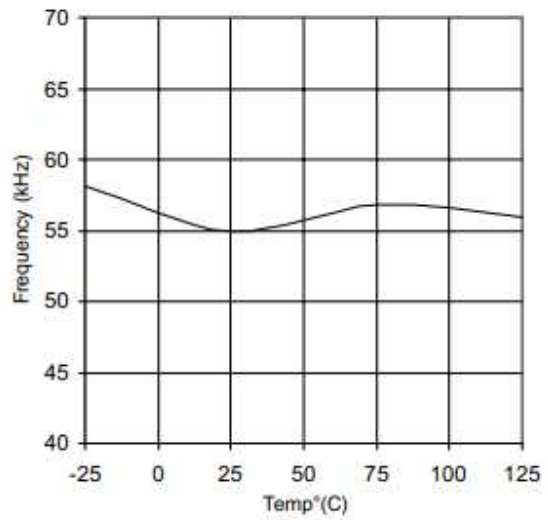
Graph 17. I_{QCC} vs V_{CC} vs Temperature (IR2156)
 V_{CCUV+}



Graph 18. I_{QCC} vs V_{CC} vs Temperature (IR2156)
 V_{CCUV-}

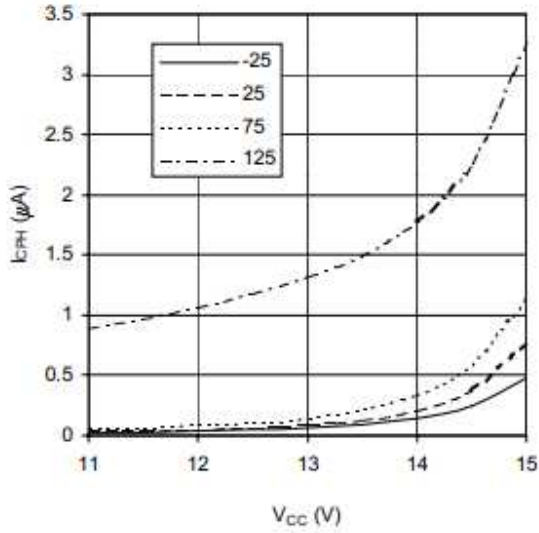


Graph 19. F_{OSC} vs V_{CC} vs Temperature (IR2156)
 $V_{CPH} = 0V$

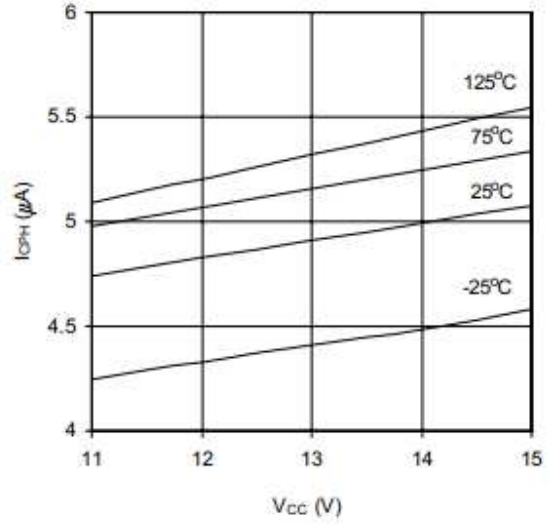


Graph 20. F_{OSC} vs Temperature (IR2156)
 $V_{CPH} = 0V$

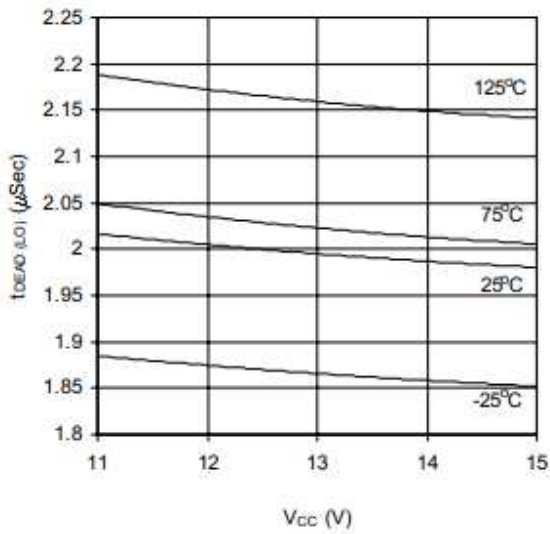
Characterization Data



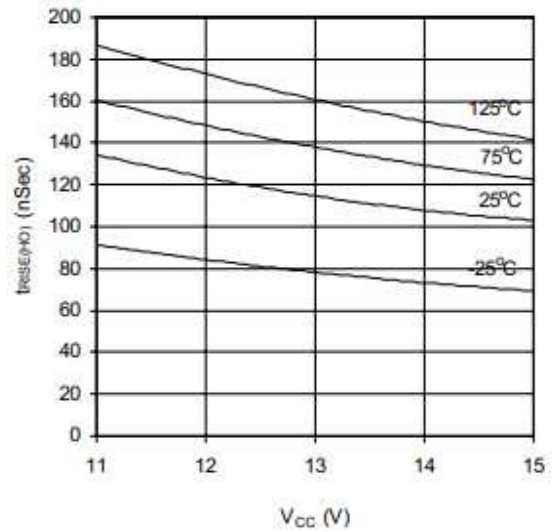
Graph 21. I_{CPH} vs V_{CC} vs Temperature (IR2156)
 $V_{CPH} = V_{CC}$



Graph 22. I_{CPH} vs V_{CC} vs Temperature (IR2156)
 $V_{CPH} = 0V$

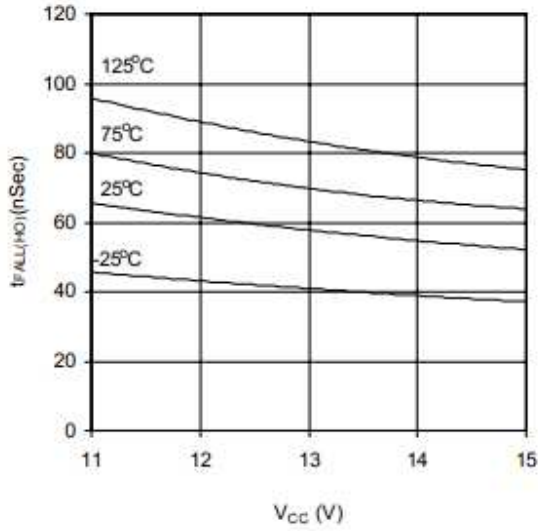


Graph 23. $t_{DEAD} (LO)$ vs V_{CC} vs Temperature (IR2156)
 $C_T = 1nF$

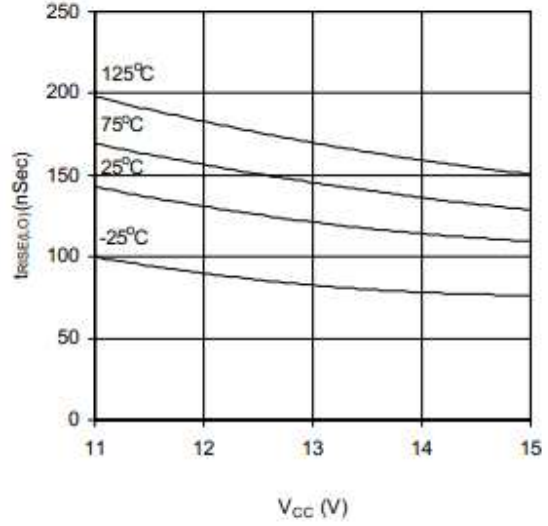


Graph 24. $t_{RISE} (HO)$ vs V_{CC} vs Temperature (IR2156)

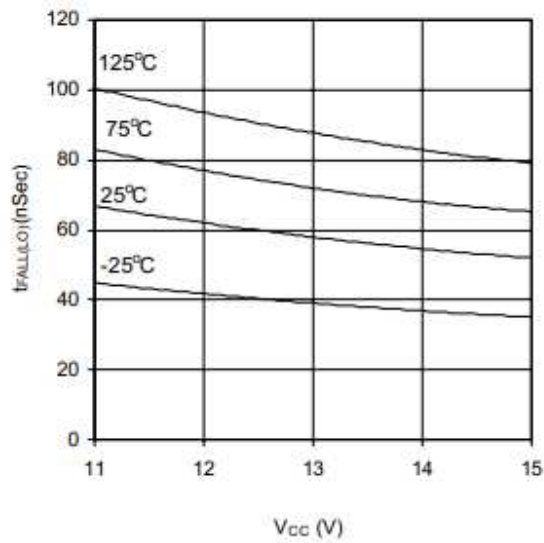
Characterization Data



Graph 25. tFALL(HO) vs V_{CC} vs Temperature (IR2156)



Graph 26. tRISE(LO) vs V_{CC} vs Temperature (IR2156)



Graph 27. tFALL(LO) vs V_{CC} vs Temperature (IR2156)

Functional Description

Under-voltage Lock-out Mode (UVLO)

The under-voltage lock-out mode (UVLO) is defined as the state the IC is in when VCC is below the turn-on threshold of the IC. To identify the different modes of the IC, refer to the State Diagram shown on page 2 of this document. The IR2156 undervoltage lock-out is designed to maintain an ultra low supply current of less than 200uA, and to guarantee the IC is fully functional before the high and low side output drivers are activated. Figure 1 shows an efficient supply voltage using the start-up current of the IRS2156 together with a charge pump from the ballast output stage (R_{SUPPLY} , C_{VCC} , D_{CP1} and D_{CP2}).

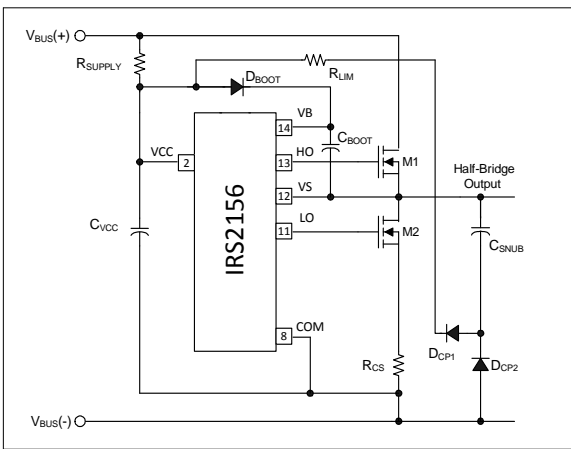


Figure 1, Start-up and supply circuitry.

The start-up capacitor (C_{VCC}) is charged by current through supply resistor (R_{SUPPLY}) minus the start-up current drawn by the IC. This resistor is chosen to provide 2X the maximum start-up current to guarantee ballast start-up at low line input voltage. Once the capacitor voltage on VCC reaches the start-up threshold V_{CCUV+} , and the SD pin is below V_{SDTH-} , the IC turns on and HO and LO begin to oscillate. The capacitor begins to discharge due to the increase in IC operating current (Figure 2).

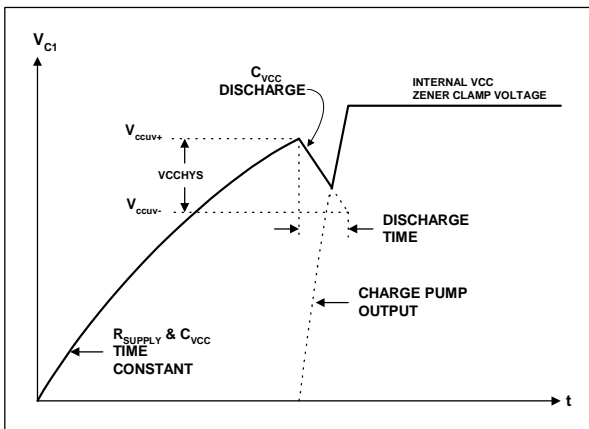


Figure 2, Supply capacitor (C_{VCC}) voltage.

During the discharge cycle, the rectified current from the charge pump charges the capacitor above the IC turn-off threshold. The charge pump and the internal 15.6V zener clamp of the IC take over as the supply voltage. The start-up capacitor and snubber capacitor must be selected such that enough supply

current is available over all ballast operating conditions. An external bootstrap diode (D_{BOOT}) and the supply capacitor (C_{BOOT}) comprise the supply voltage for the high side driver circuitry. To guarantee that the high-side supply is charged up before the first pulse on pin HO, the first pulse from the output drivers comes from the LO pin. During undervoltage lock-out mode, the high- and low-side driver outputs HO and LO are both low, pin CT is connected internally to COM to disable the oscillator, and pin CPH is connected internally to COM for resetting the preheat time.

Preheat Mode (PH)

The preheat mode is defined as the state the IC is in when the lamp filaments are being heated to their correct emission temperature. This is necessary for maximizing lamp life and reducing the required ignition voltage. The IR2156 enters preheat mode when VCC exceeds the V_{CCUV+} positive-going threshold. HO and LO begin to oscillate at the preheat frequency with 50% duty cycle and with a dead-time which is set by the value of the external timing capacitor, CT, and internal deadtime resistor, RDT. Pin CPH is disconnected from COM and an internal 4uA current source (Figure 3) charges the external preheat timing capacitor on CPH linearly. The over-current protection on pin CS is disabled during preheat.

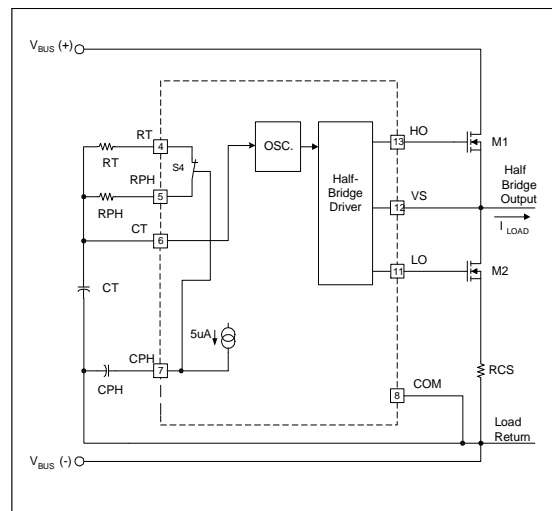


Figure 3, Preheat circuitry

The preheat frequency is determined by the parallel combination of resistors R_T and R_{PH} , together with timing capacitor C_T . C_T charges and discharges between 1/3 and 3/5 of VCC (see Timing Diagram, page 9). C_T is charged exponentially through the parallel combination of R_T and R_{PH} connected internally to VCC through MOSFET S1. The charge time of C_T from 1/3 to 3/5 VCC is the on-time of the respective output gate driver, HO or LO. Once C_T exceeds 3/5 VCC, MOSFET S1 is turned off, disconnecting R_T and R_{PH} from VCC. C_T is then discharged exponentially through an internal resistor, RDT, through MOSFET S3 to COM. The discharge time of C_T from 3/5 to 1/3 VCC is the dead-time (both off) of the output gate drivers, HO and LO. The selected value of C_T together with RDT therefore program the desired dead-time (see Design Equations, page 12, Equations 1 and 2). Once C_T discharges below 1/3 VCC, MOSFET S3 is turned off, disconnecting RDT from COM, and MOSFET S1 is turned on, connecting R_T and R_{PH} again to VCC. The frequency remains at the preheat frequency until the voltage on pin CPH exceeds 13V and the IC enters Ignition Mode. During the preheat mode, both the over-current protection and the

Design Equations

Note: The results from the following design equations can differ slightly from experimental measurements due to IC tolerances, component tolerances, and oscillator over- and under-shoot due to internal comparator response time.

For additional design support for different lamp types and AC line input configurations, including component calculations, schematics, bill of materials and inductor specifications, please download IR's Ballast Design Assistant (BDA) software at www.irf.com.

Step 1: Program Dead-time

The dead-time between the gate driver outputs HO and LO is programmed with timing capacitor CT and an internal dead-time resistor RDT. The dead-time is the discharge time of capacitor CT from 3/5VCC to 1/3VCC and is given as:

$$t_{DT} = C_T \cdot 2000 \quad [\text{Seconds}] \quad (1)$$

Or,

$$C_T = \frac{t_{DT}}{2000} \quad [\text{Farads}] \quad (2)$$

Step 2: Program Run Frequency

The final run frequency is programmed with timing resistor RT and timing capacitor CT. The charge time of capacitor CT from 1/3VCC to 3/5VCC determines the on-time of HO and LO gate driver outputs. The run frequency is therefore given as:

$$f_{RUN} = \frac{1}{2 \cdot C_T (0.6 \cdot R_T + 2000)} \quad [\text{Hertz}] \quad (3)$$

Or,

$$R_T = \frac{1}{1.12 \cdot C_T \cdot f_{RUN}} - 3333 \quad [\text{Ohms}] \quad (4)$$

Step 3: Program Preheat Frequency

The preheat frequency is programmed with timing resistors RT and RPH, and timing capacitor CT. The timing resistors are

connected in parallel internally for the duration of the preheat time. The preheat frequency is therefore given as:

$$f_{PH} = \frac{1}{2 \cdot C_T \cdot \left(\frac{0.51 \cdot R_T \cdot R_{PH}}{R_T + R_{PH}} + 2000 \right)} \quad [\text{Hertz}] \quad (5)$$

Or,

$$R_{PH} = \frac{\left(\frac{1}{1.12 \cdot C_T \cdot f_{PH}} - 3333 \right) \cdot R_T}{R_T - \left(\frac{1}{1.12 \cdot C_T \cdot f_{PH}} - 3333 \right)} \quad [\text{Ohms}] \quad (6)$$

Step 4: Program Preheat Time

The preheat time is defined by the time it takes for the capacitor on pin CPH to charge up to 13 volts. An internal current source of 5uA flows out of pin CPH. The preheat time is therefore given as:

$$t_{PH} = C_{PH} \cdot 3.02e6 \quad [\text{Seconds}] \quad (7)$$

Or,

$$C_{PH} = t_{PH} \cdot 0.33e-6 \quad [\text{Farads}] \quad (8)$$

Step 5: Program Maximum Ignition Current

The maximum ignition current is programmed with the external resistor RCS and an internal threshold of 1.25 volts (VCSTH+). This threshold determines the over-current limit of the ballast, which can be exceeded when the frequency ramps down towards resonance during ignition and the lamp does not ignite. The maximum ignition current is given as:

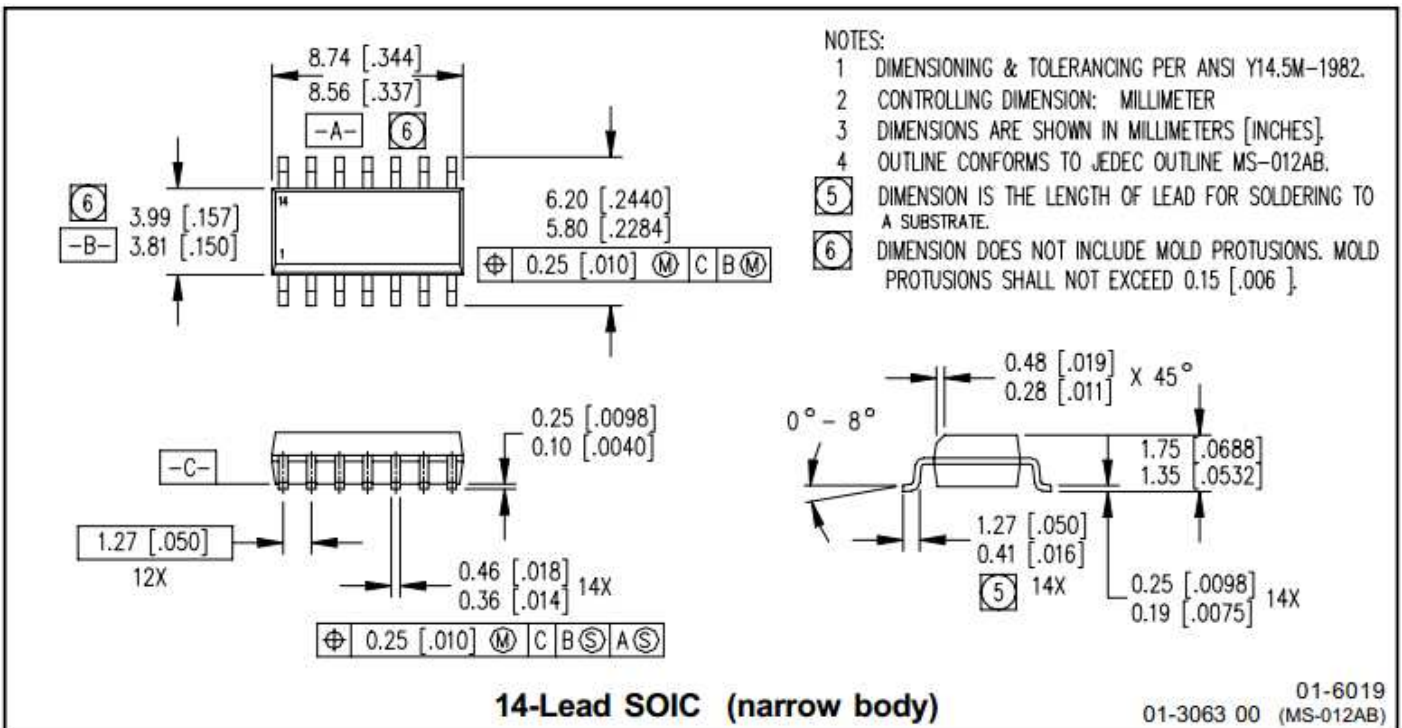
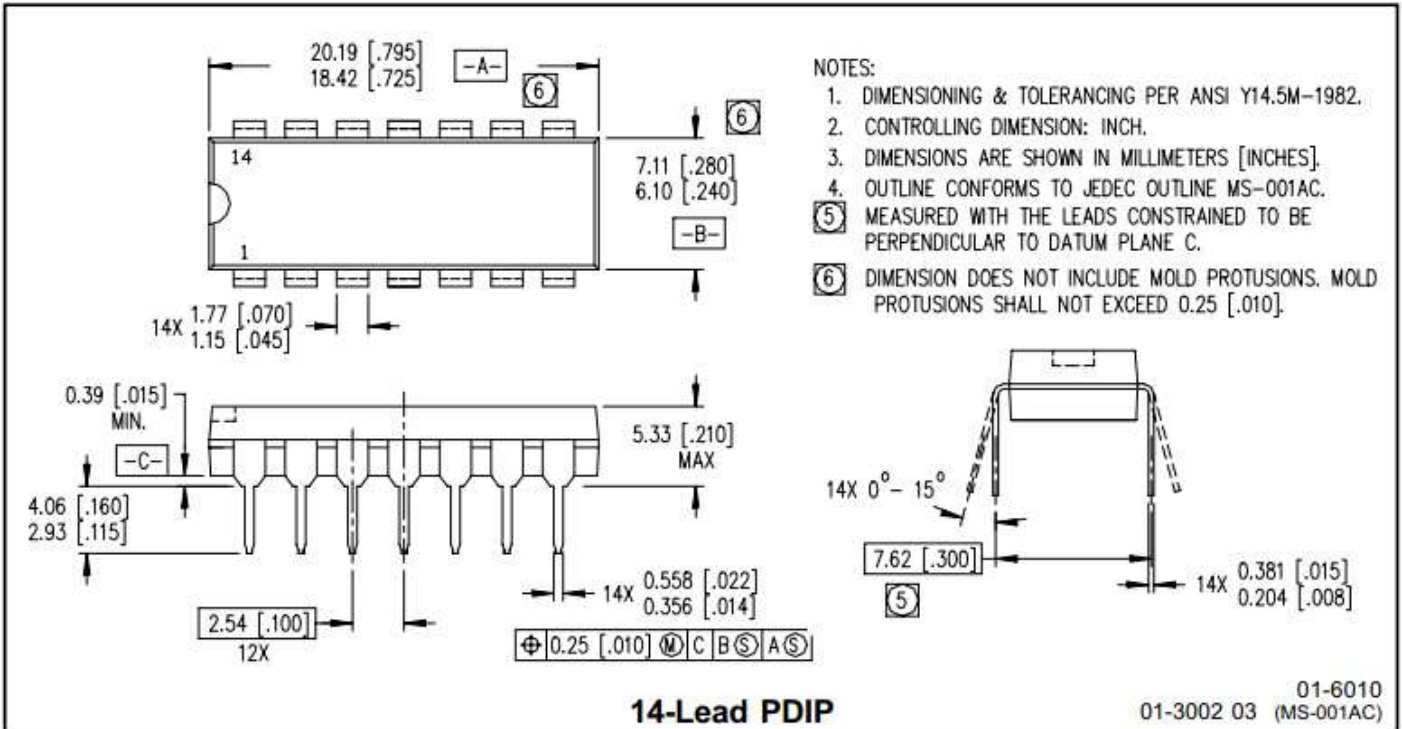
$$I_{IGN} = \frac{1.25}{R_{CS}} \quad [\text{Amps Peak}] \quad (9)$$

Or,

$$R_{CS} = \frac{1.25}{I_{IGN}} \quad [\text{Ohms}] \quad (10)$$

IR2156(S)PbF

Case Outline



IR2156(S)PbF

Qualification: Lead-free MSL3, industrial

International
IOR Rectifier

<http://www.irf.com>

Data and specifications subject to change without notice.
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