

Middle Power Class-D Speaker Amplifier Series

17W+17W

Class D Speaker Amplifier for Digital Input

BD28623MUV

General Description

BD28623MUV is a Class D Speaker Amplifier designed for Flat-panel TVs in particular for space-saving and low-power consumption. This IC delivers an output power of 20W+20W. This IC employs state-of-the-art Bipolar, CMOS, and DMOS (BCD) process technology. With this technology, the IC can achieve high efficiency. In addition, the IC is packaged in a compact back-surface heat-sink type power package to achieve low power consumption and low heat generation and to eliminate need for external heat-sink. With this package, total output power is only 34W as compared to 40W total output power of package with external heat-sink. This product satisfies all needs for drastic downsizing, low-profile structures and powerful high quality playback of sound systems.

Features

- 1 Digital Audio Interface
I²S format
SDATA: 16 / 20 / 24bit
LRCLK (f_s): 32 kHz / 44.1kHz / 48kHz
BCLK: 64f_s (fixed)
MCLK: 256f_s / 512f_s Automatic Identification)
- Low supply current at RESET mode.
- Slew rate controller
; No need snubber circuit (V_{cc}≤22V)
- Output Feedback Circuitry which prevents decrease of sound quality caused by change of power supply voltage, achieves low noise and low distortion, So the large electrolytic-capacitors for V_{cc} bypass is able to be eliminated.
- Variable Gain (17dB / 20dB / 26dB)
- Wide power supply voltage range (8.5V to 24V)
- High efficiency, low heat
- Pop noise prevention at power supply on / off
- Soft Muting Technology
- High reliability design by built-in protection circuits
 - Overheat protection
 - Under voltage protection
 - Output short protection
 - Output DC voltage protection
 - Clock stop protection (MCLK, BCLK, LRCLK)
- Small package (VQFN024V4040)

Applications

- Flat Panel TVs (LCD, Plasma)
- Home Audio (Sound Bar)
- Amusement Equipment
- Electronic Music Equipment
- Desktop PC, etc.

Key Specifications

- Supply Voltage: 8.5V to 24V
- Speaker Output Power: 17W+17W (Typ)
(V_{cc}=18V, R_L=8Ω, Gain=26dB)
- Total Harmonic Distortion: 0.08% (Typ) @P_O=1W
(V_{cc}=12V, R_L=8Ω, Gain=20dB)
- Crosstalk: 90dB (Typ)
- PSRR: 60dB (Typ)
- Output Noise Voltage: 150μV_{rms} (Typ)
- Standby Current: 33μA (Typ)
- Operating Temperature Range: -25°C to +85°C

Package

VQFN024V4040

 W(Typ) x D(Typ) x H(Max)
4.00mm x 4.00mm x 1.00mm


Typical Application Circuit

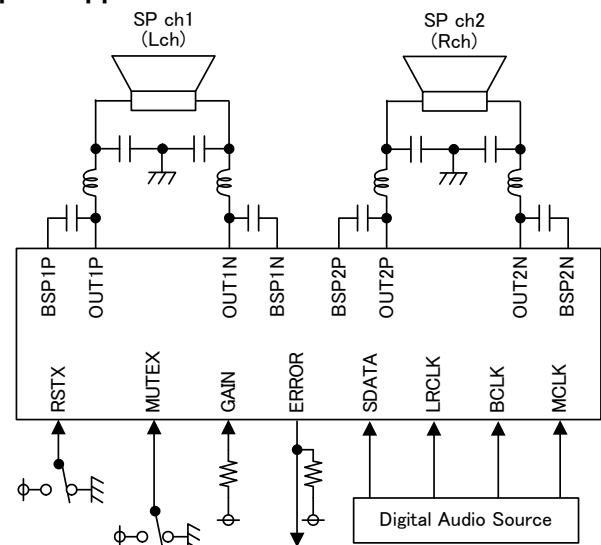


Figure 1. Typical Application Circuit

Pin Configuration

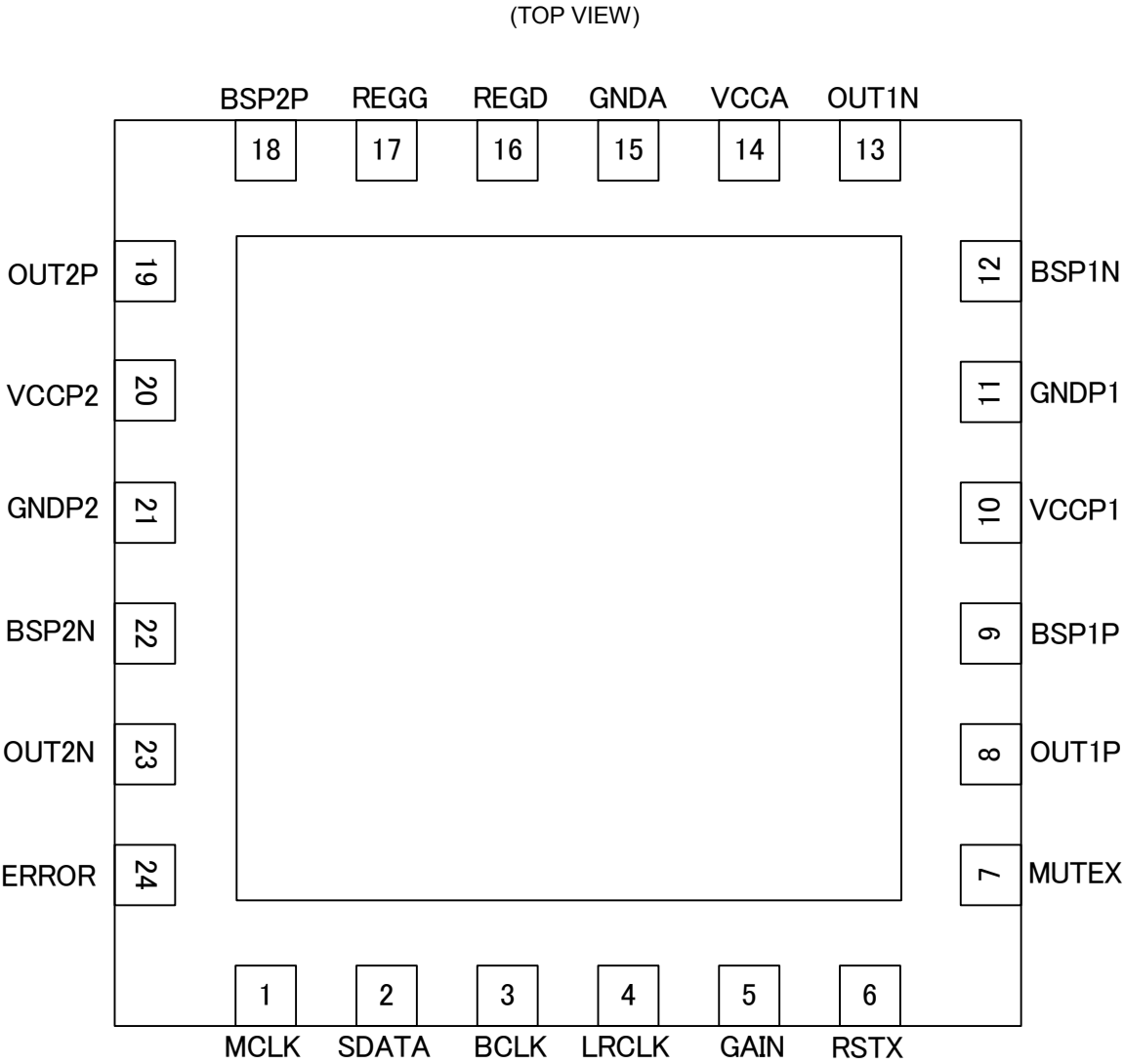


Figure 2. Pin Configuration

Pin Descriptions, I/O Equivalent Circuits (Provided pin voltages are typical values)

Pin No.	Pin Name	Pin Voltage	Pin Descriptions	Internal Equivalent Circuit
1 2 3 4	MCLK SDATA BCLK LRCLK	0V	Digital sound signal input pin	
7	MUTEX		Speaker output mute control pin H: Mute OFF L: Mute ON	
5	PLIMIT	0V	Gain setting pin	
6	RSTX	0V	Reset pin H: Reset OFF L: Reset ON	
8	OUT1P	V _{CC} to 0V	Output pin of Ch1 positive PWM signal Please connect to output LPF. *If this pin is shorted to GND, the IC may be broken.	
9	BSP1P	-	Boot-strap pin of Ch1 positive PWM signal Please connect a capacitor to OUT1P.	
10	VCCP1	-	Power supply pin for Ch1 PWM signal Please connect a capacitor.	
11	GNDP1	0V	GND pin for Ch1 PWM signal	
12	BSP1N	-	Boot-strap pin of Ch1 negative PWM signal Please connect a capacitor to OUT1N.	
13	OUT1N	V _{CC} to 0V	Output pin of Ch1 negative PWM signal Please connect to output LPF. *If this pin is shorted to GND, the IC may be broken.	
14	VCCA	V _{CC}	Power supply pin for Analog signal Please connect a capacitor to GND.	—
15	GNDA	0V	GND pin for Analog signal	—
16	REGD	5.0V	Internal power supply pin for Digital circuit Please connect a capacitor to GND. *The REGD terminal of BD28623MUV should not be used as external supply. Therefore, don't connect anything except for the capacitor for stabilization.	

Pin Descriptions, I/O Equivalent Circuits – continued (Provided pin voltages are typical values)

Pin No.	Pin Name	Pin Voltage	Pin Descriptions	Internal Equivalent Circuit
17	REGG	5.7V	Internal power supply pin for Gate driver Please connect a capacitor to GND. *The REGG terminal of BD28623MUV should not be used as external supply. Therefore, don't connect anything except for the capacitor for stabilization.	
18	BSP2P	-	Boot-strap pin of Ch2 positive PWM signal Please connect a capacitor to OUT2P.	
19	OUT2P	V _{CC} to 0V	Output pin of Ch2 positive PWM signal Please connect to output LPF. *If this pin is shorted to GND, the IC may be broken.	
20	VCCP2	V _{CC}	Power supply pin for Ch2 PWM signal Please connect a capacitor to GND.	
21	GNDP2	0V	GND pin for Ch2 PWM signal	
22	BSP2N	-	Boot-strap pin of Ch2 negative PWM signal Please connect a capacitor to OUT2N.	
23	OUT2N	V _{CC} to 0V	Output pin of Ch2 negative PWM signal Please connect to output LPF. *If this pin is shorted to GND, the IC may be broken.	
24	ERROR	-	Error flag pin Please connect pull-up resistor. H: Normal L: Error *An error flag is outputted when Output Short Protection, DC Voltage Protection in the speaker, and High Temperature Protection are operated. This flag shows IC condition during operation.	

The numerical value of internal equivalent circuit is typical value, not guaranteed value.

Block Diagram

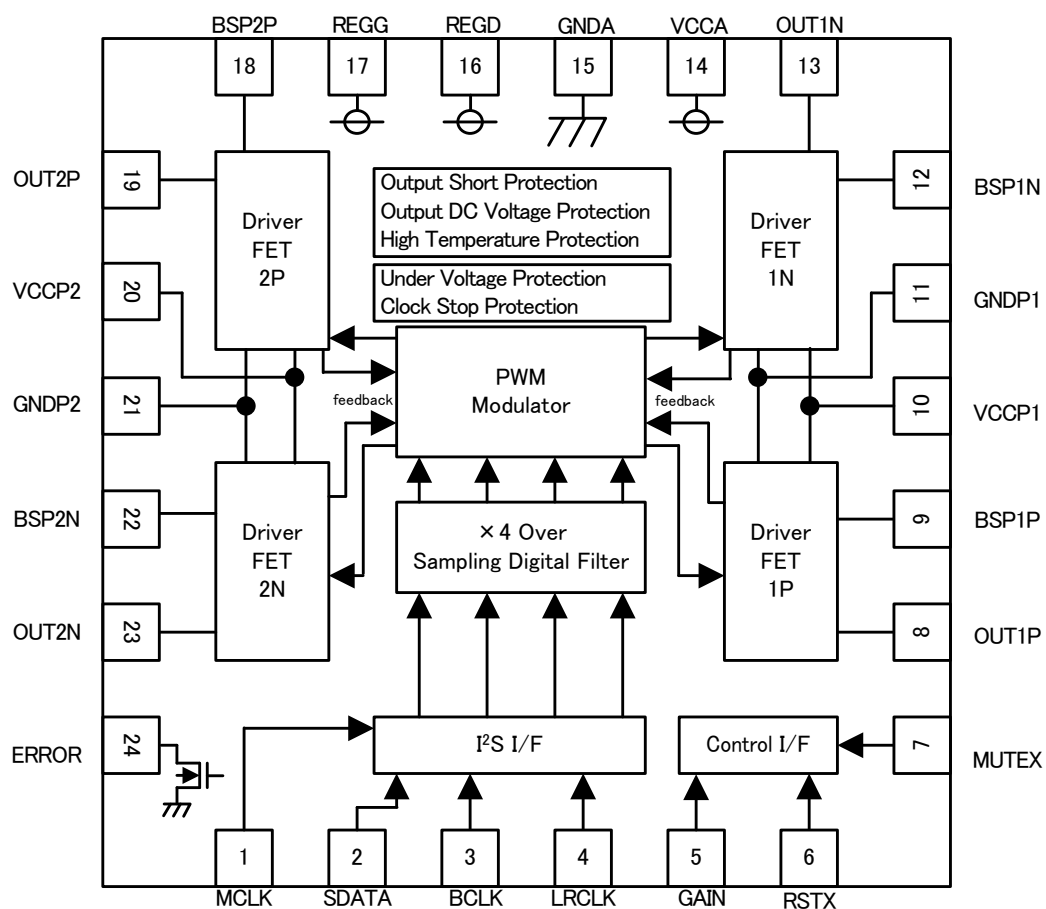


Figure 3. Block Diagram

Absolute Maximum Ratings

Parameter	Symbol	Limit	Unit	Conditions
Supply Voltage ^{(Note 1) (Note 2)}	V_{CCMAX}	-0.3 to +30	V	Pin10, 14, 20
Power Dissipation	P_d	2.21 ^(Note 3)	W	Please refer to Power Dissipation for details.
		3.56 ^(Note 4)		
Input Voltage1 ^(Note 1)	V_{IN1}	-0.3 to +3.7	V	Pin1-7
Terminal Voltage 1 ^(Note 1)	V_{PIN1}	-0.3 to +7	V	Pin16, 17
Terminal Voltage 2 ^{(Note 1) (Note 5-1)}	V_{PIN2}	-0.3 to V_{CC}	V	Pin8, 13, 19, 23
Terminal Voltage 3 ^{(Note 1) (Note 5-2)}	V_{PIN3}	-0.3 to $OUT_{xx}+7$	V	Pin9, 12, 18, 22
Open-drain Terminal Voltage ^(Note 1)	V_{ERR}	-0.3 to V_{CCMAX}	V	Pin24
Operating Temperature Range	T_{opr}	-25 to +85	°C	
Storage Temperature Range	T_{stg}	-55 to +150	°C	
Maximum Junction Temperature	T_{jmax}	+150	°C	

(Note 1) Voltage that can be applied with reference to GND (Pin11, 15, 21).

(Note 2) P_d and $T_{jmax}=150^{\circ}\text{C}$ must not be exceeded.

(Note 3) 74.2mm×74.2mm×1.6mm, FR4, 4-layer glass epoxy board

(Top and bottom layer back copper foil size: 20.2mm², 2nd and 3rd layer back copper foil size: 5505mm²)

Derate by 17.7mW/°C when operating above $T_a=25^{\circ}\text{C}$. The board is provided with thermal via.

(Note 4) 74.2mm×74.2mm×1.6mm, FR4, 4-layer glass epoxy board

(Top and bottom layer back copper foil size: 5505mm²)

Derate by 28.5mW/°C when operating above $T_a=25^{\circ}\text{C}$. The board is provided with thermal via.

(Note 5-1) The chip should be used within AC peak limits at all conditions. Overshoot should be $\leq 30\text{V}$ with reference to GND.

Undershoot should be $\leq 10\text{nsec}$ and $\leq 30\text{V}$ with reference to V_{CC} . (Please refer to figure 4-1.)

(Note 5-2) The chip should be used within AC peak limits at all conditions. Overshoot should be $\leq OUT_{xx}+7\text{V}$ with reference to OUT_{xx} .

Undershoot should be $\leq 10\text{nsec}$ and $\leq OUT_{xx}+7\text{V}$ with reference to OUT_{xx} . (Please refer to figure 4-2.)

Caution: Operating the IC over the absolute maximum ratings may damage the IC. The damage can either be a short circuit between pins or an open circuit between pins and the internal circuitry. Therefore, it is important to consider circuit protection measures, such as adding a fuse, in case the IC is operated over the absolute maximum ratings.

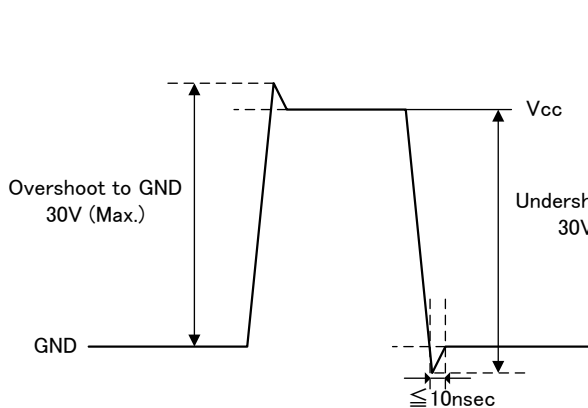


Figure 4-1

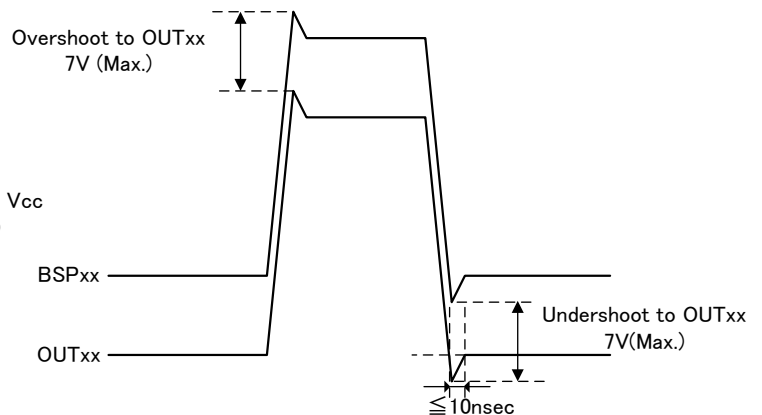


Figure 4-2

Recommended Operating Conditions

Parameter	Symbol	Limit	Unit	Conditions
Supply Voltage ^{(Note 1) (Note 2)}	V_{CC}	8.5 to 24	V	Pin10, 14, 20
Minimum Load Impedance ^(Note 6)	R_L	6.4	Ω	$21\text{V} < V_{CC} \leq 24\text{V}$
		4.8		$14\text{V} < V_{CC} \leq 21\text{V}$
		3.6		$V_{CC} \leq 14\text{V}$

(Note 6) P_d should not be exceeded.

Electrical Characteristics

(Unless otherwise specified, Ta=25°C, V_{CC}=18V, f=1kHz, R_L=8Ω, RSTX=3.3V, MUTEX=3.3V, Gain= 20dB, f_S=48kHz, MCLK=256f_S, Output LC filter: L=10μH, C=0.68μF, Without Snubber circuit)

Parameter	Symbol	Limit			Unit	Conditions
		Min	Typ	Max		
Total Circuit						
Circuit Current (Reset Mode)	I _{CC1}	-	33	200	μA	No load, RSTX=0V, MUTEX=0V
Circuit Current (Mute Mode)	I _{CC2}	-	15	25	mA	No load, RSTX=3.3V, MUTEX=0V
Circuit Current (Active Mode)	I _{CC3}	-	40	80	mA	No load, RSTX=3.3V, MUTEX=3.3V
Open-drain Terminal Low Level Voltage	V _{ERR}	-	-	0.8	V	I _o =0.5mA
Regulator Output Voltage 1	V _{REGG}	4.6	5.7	6.5	V	RSTX=3.3V, MUTEX=3.3V
Regulator Output Voltage 2	V _{REGD}	4.2	5.0	5.7	V	RSTX=3.3V, MUTEX=3.3V
High level Input Voltage 1	V _{IH1}	2.2	-	3.3	V	Pin1-4,6-7
Low level Input Voltage 1	V _{IL1}	0	-	0.8	V	Pin1-4,6-7
High level Input Voltage 2	V _{IH2}	2.6	-	3.3	V	Pin5
Low level Input Voltage 2	V _{IL2}	0	-	0.45	V	Pin5
Input Current1 (Input Pull-down Terminal)	I _{IH}	27.5	33	42	μA	V _{IN} = 3.3V, Pin1-4,6-7
Input Current2 (Input Pull-down Terminal)	I _{IH2}	65	100	135	μA	V _{IN} = 3.3V, Pin5
Speaker Parts						
Maximum Output Power 1 ^(Note 7)	P _{O1}	-	15	-	W	V _{CC} =16V, THD+N=10%, GAIN=26dB
Maximum Output Power 2 ^(Note 7)	P _{O2}	10	12.5	-	W	V _{CC} =16V, THD+N<10%, GAIN=20dB
Maximum Output Power 3 ^(Note 7)	P _{O3}	5	6.3	-	W	V _{CC} =16V, THD+N<10%, GAIN=17dB
Voltage Gain1 ^(Note 7)	G _{V26}	25	26	27	dB	P _O =1W, GAIN=H
Voltage Gain2 ^(Note 7)	G _{V20}	19	20	21	dB	P _O =1W , GAIN=Pull up(47kΩ)
Voltage Gain3 ^(Note 7)	G _{V17}	16	17	18	dB	P _O =1W, GAIN=L
Total Harmonic Distortion1 ^(Note 7)	THD ₁	-	0.08	-	%	V _{CC} =12V, P _O =1W BW=20 to 20kHz (AES17) GAIN=20dB, With snubber circuit
Crosstalk ^(Note 7)	CT	60	90	-	dB	P _O =1W, 1kHz BPF, GAIN=20dB
PSRR ^(Note 7)	PSRR	-	60	-	dB	V _{ripple} =1V _{rms} , f=1kHz, GAIN=20dB
Output Noise Voltage ^(Note 7)	V _{NO}	-	150	250	μV _{rms}	Input=-∞dBFS, BW=IHF-A, GAIN=20dB
PWM (Pulse Width Modulation) Frequency	f _{PWM}	-	512	-	kHz	f _S =32kHz
		-	705.6	-	kHz	f _S =44.1kHz
		-	768	-	kHz	f _S =48kHz

(Note 7) The rated values of items above indicate average performances of the device, which largely depend on circuit layouts, components, and power supplies. The reference values are those applicable to the device and components directly installed on a board specified by ROHM during testing.

Typical Performance Curves (1/11)

(Unless otherwise specified, $T_a=25^\circ\text{C}$, $V_{CC}=18\text{V}$, $f=1\text{kHz}$, $R_L=8\Omega$, $R_{STX}=3.3\text{V}$, $MUTEX=3.3\text{V}$, $f_s=48\text{kHz}$, $MCLK=256f_s$, $\text{Gain}=26\text{dB}$, ROHM 4-layer Board)

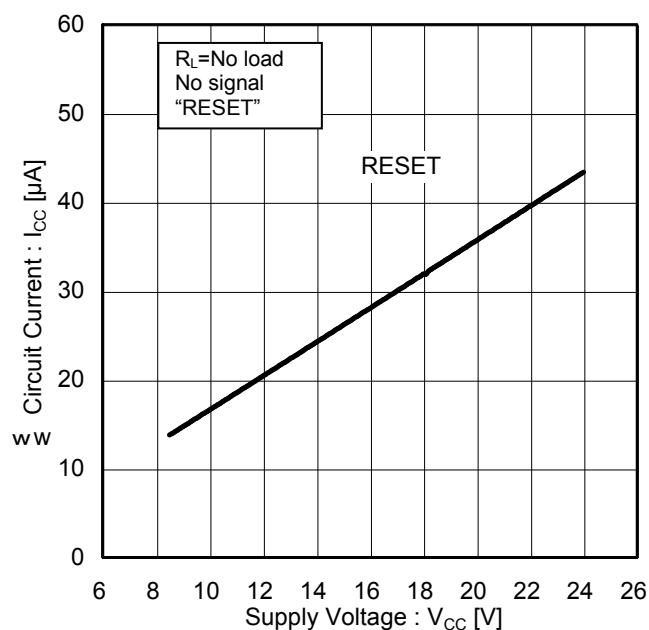


Figure 5. Circuit Current vs Supply Voltage (RESET)

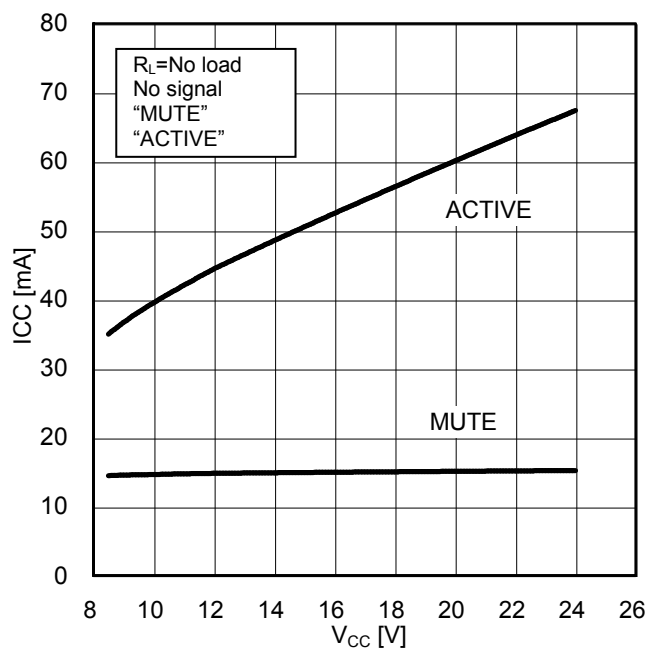


Figure 6. Circuit Current vs Supply Voltage (MUTE, ACTIVE)

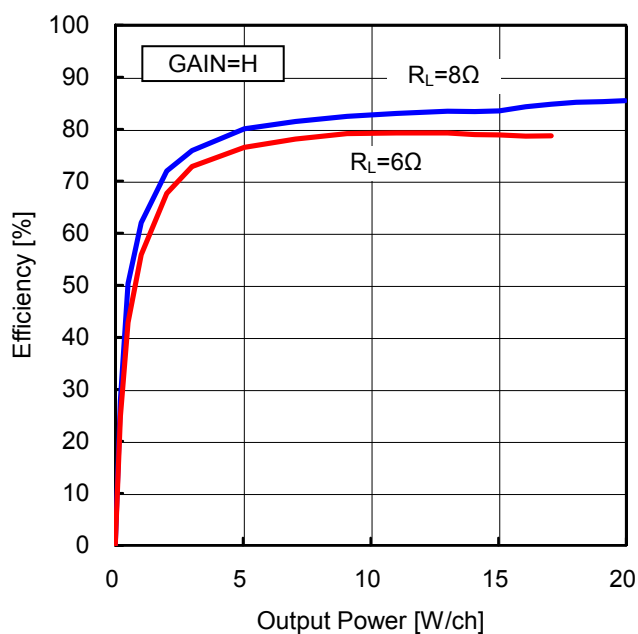


Figure 7. Efficiency vs Output Power (8Ω, 6Ω)

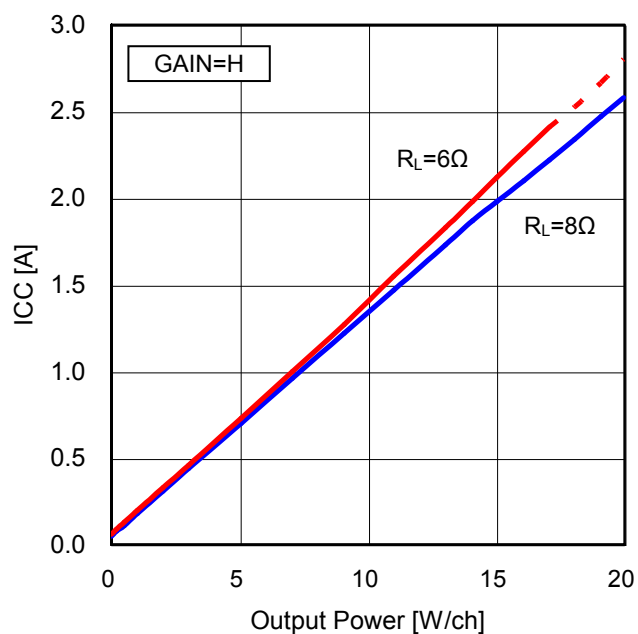


Figure 8. Circuit Current vs Output Power (8Ω, 6Ω)

※ Dotted line means power dissipation is exceeded.

Typical Performance Curves – continued (2/11)

(Unless otherwise specified, $T_a=25^\circ\text{C}$, $V_{CC}=18\text{V}$, $f=1\text{kHz}$, $R_L=8\Omega$, $R_{STX}=3.3\text{V}$, $MUTEX=3.3\text{V}$, $f_s=48\text{kHz}$, $MCLK=256f_s$, $\text{Gain}=20\text{dB}$, ROHM 4-layer Board)

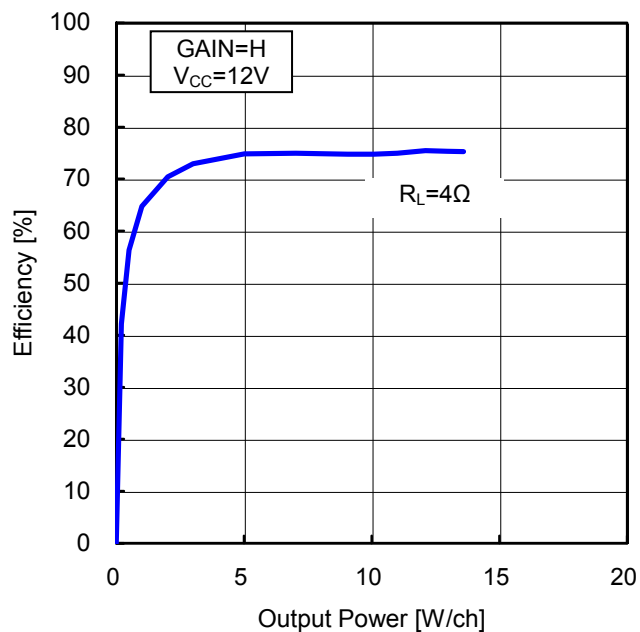


Figure 9. Efficiency vs Output Power (4Ω)

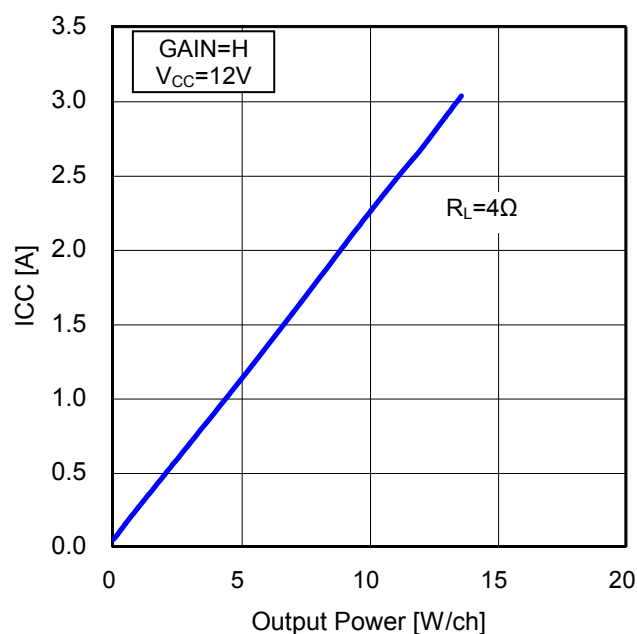


Figure 10. Circuit Current vs Output Power (4Ω)

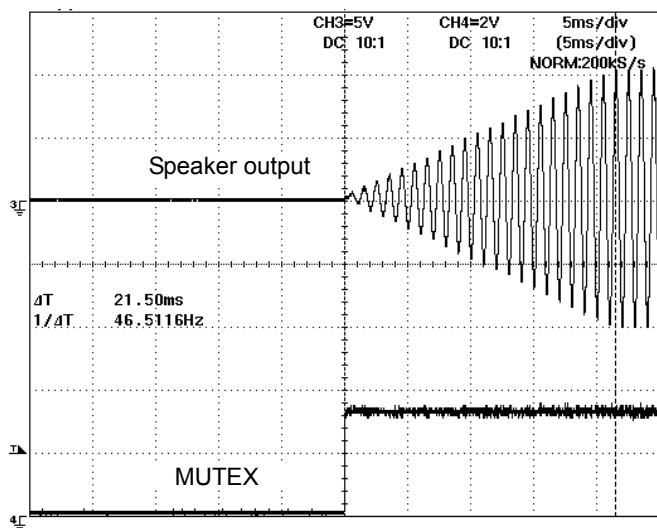


Figure 11. Waveform of Soft Start

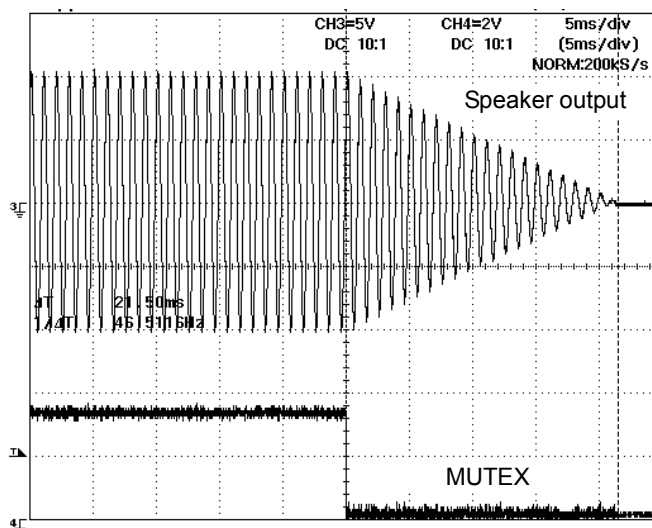


Figure 12. Waveform of Soft Mute

Typical Performance Curves - continued (3/11)

(Unless otherwise specified, $T_a=25^\circ\text{C}$, $V_{CC}=18\text{V}$, $f=1\text{kHz}$, $R_L=8\Omega$, $R_{STX}=3.3\Omega$, $MUTEX=3.3\text{V}$, $f_s=48\text{kHz}$, $MCLK=256f_s$, Gain=26dB, ROHM 4-layer Board)

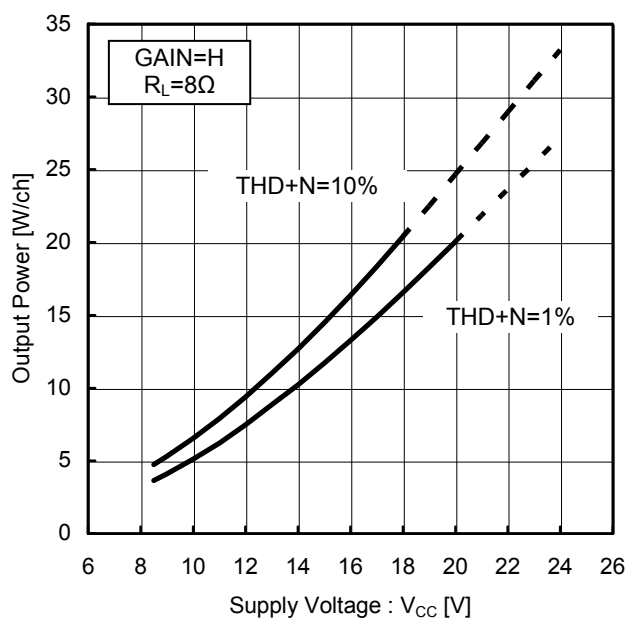


Figure 13. Output Power vs Supply Voltage (8Ω)

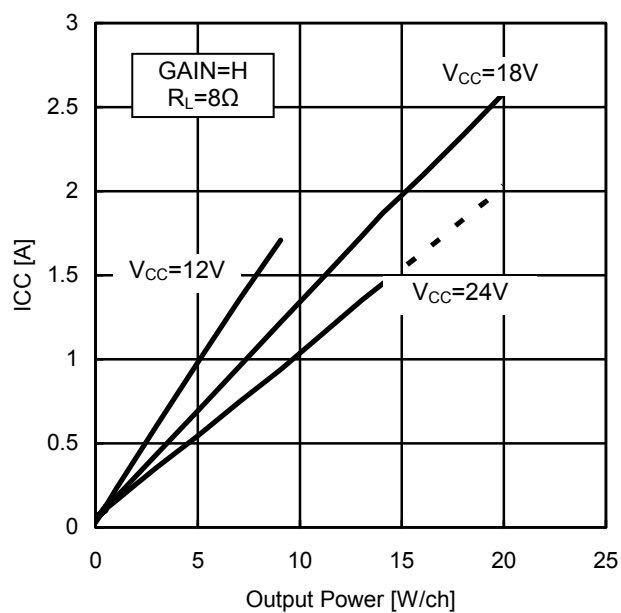


Figure 14. Circuit Current vs Output Power (8Ω)

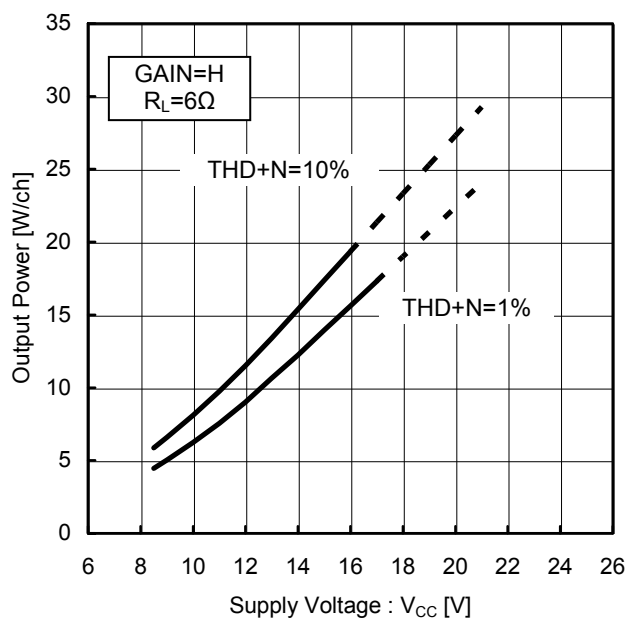


Figure 15. Output Power vs Supply Voltage (6Ω)

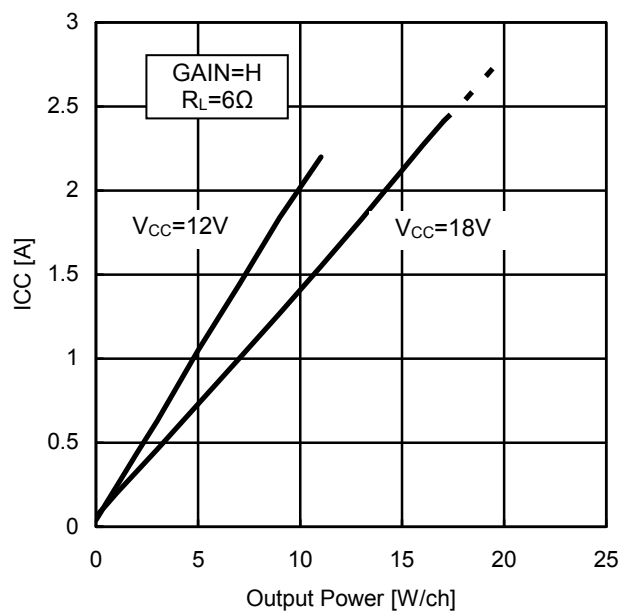


Figure 16. Circuit Current vs Output Power (6Ω)

※ Dotted line means power dissipation is exceeded.

Typical Performance Curves - continued (4/11)

(Unless otherwise specified, $T_a=25^\circ\text{C}$, $V_{CC}=18\text{V}$, $f=1\text{kHz}$, $R_L=8\Omega$, $R_{STX}=3.3\text{V}$, $MUTEX=3.3\text{V}$, $f_s=48\text{kHz}$, $MCLK=256f_s$, $\text{Gain}=20\text{dB}$, ROHM 4-layer Board)

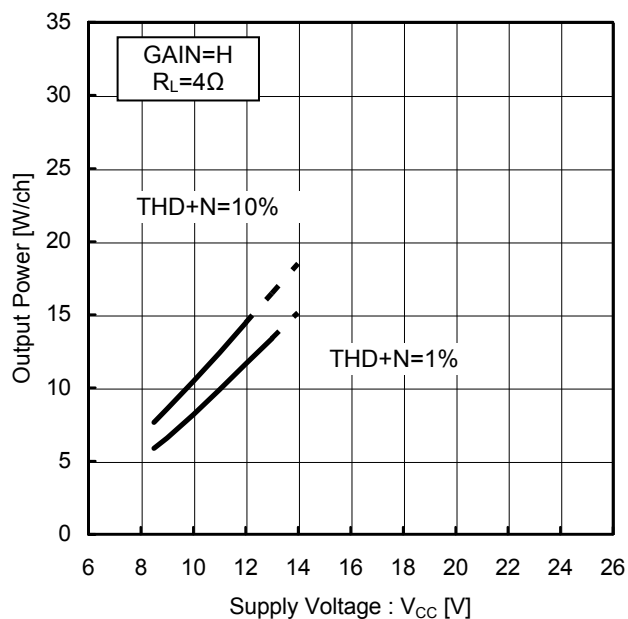


Figure 17. Output Power vs Supply Voltage (4Ω)

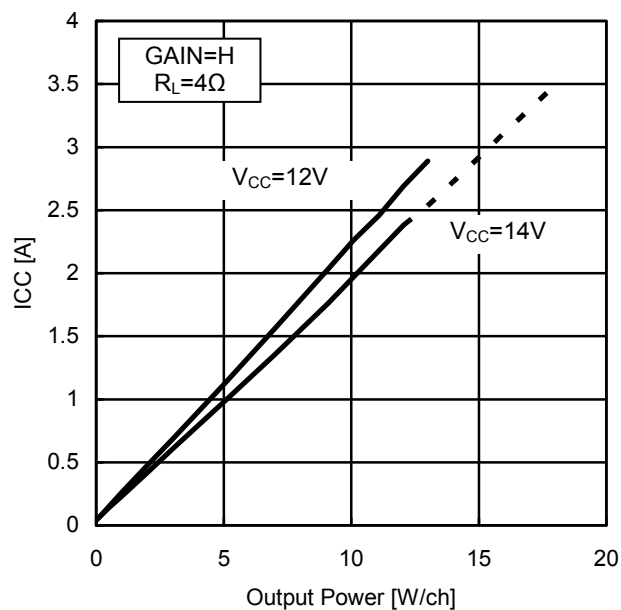


Figure 18. Circuit Current vs Output Power (4Ω)

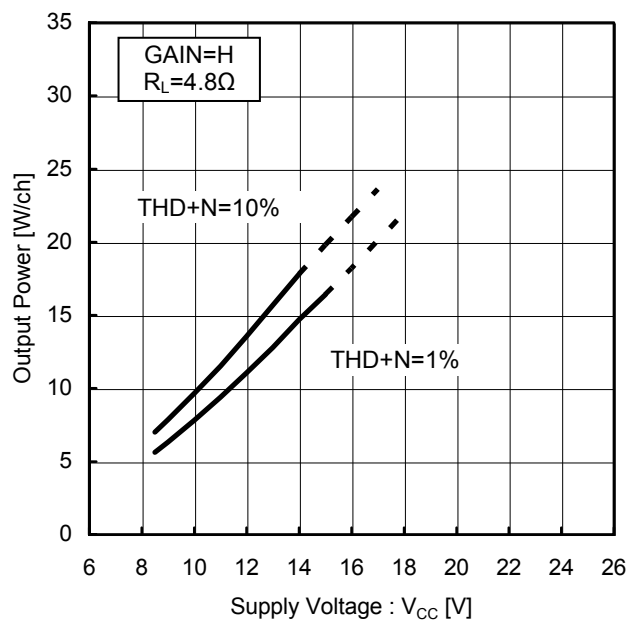


Figure 19. Output Power vs Supply Voltage (4.8Ω)

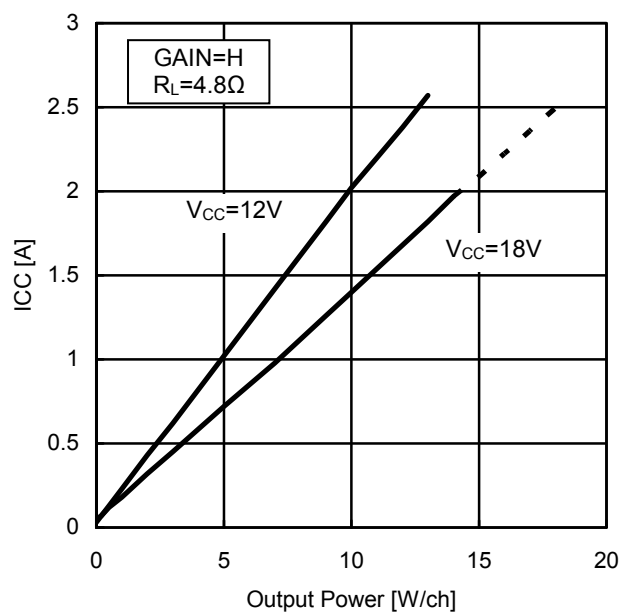


Figure 20. Circuit Current vs Output Power (4.8Ω)

※ Dotted line means power dissipation is exceeded.

Typical Performance Curves - continued (5/11)

(Unless otherwise specified, $T_a=25^\circ\text{C}$, $V_{CC}=18\text{V}$, $f=1\text{kHz}$, $R_L=8\Omega$, $R_{STX}=3.3\Omega$, $MUTEX=3.3\text{V}$, $f_s=48\text{kHz}$, $MCLK=256f_s$, $\text{Gain}=20\text{dB}$, ROHM 4-layer Board)

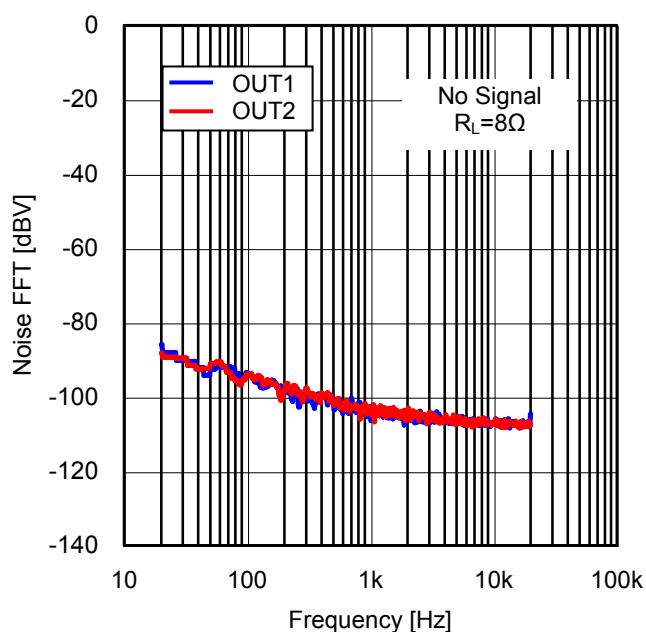


Figure 21. FFT of output noise voltage (8Ω)

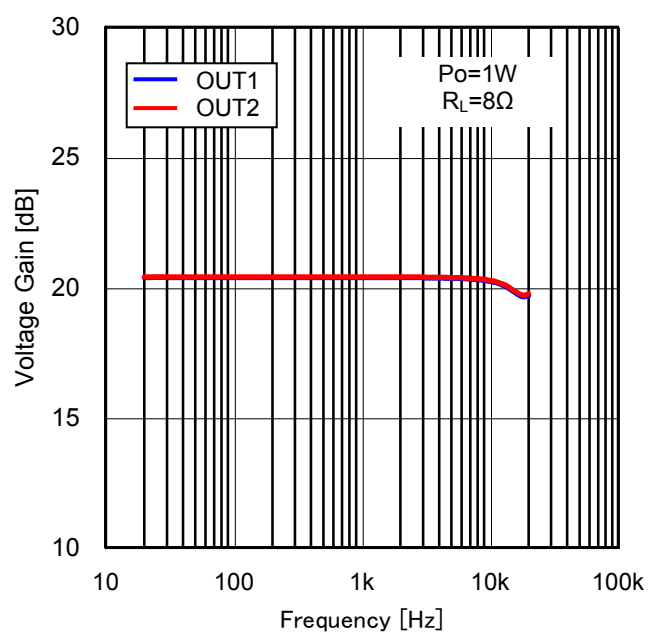


Figure 22. Voltage Gain vs Frequency (8Ω)

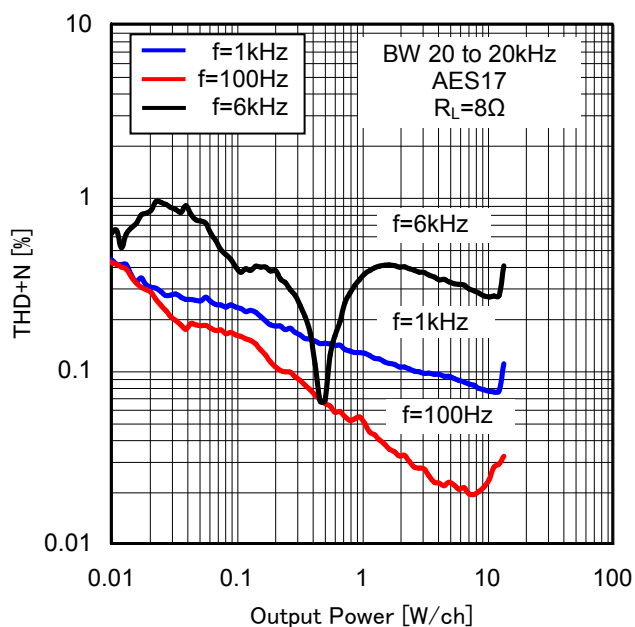


Figure 23. THD+N vs Output Power (8Ω)

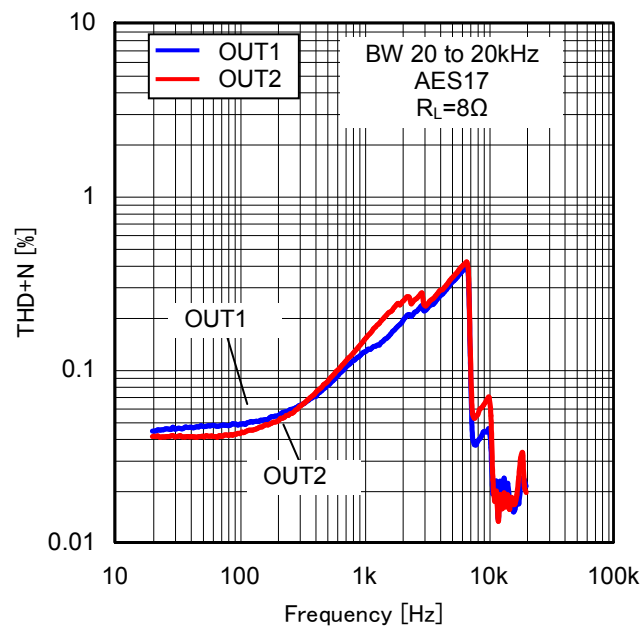


Figure 24. THD+N vs Frequency (8Ω)

Typical Performance Curves - continued (6/11)

(Unless otherwise specified, $T_a=25^\circ\text{C}$, $V_{CC}=18\text{V}$, $f=1\text{kHz}$, $R_L=8\Omega/6\Omega$, $R_{STX}=3.3\Omega$, $MUTEX=3.3\text{V}$, $f_s=48\text{kHz}$, $MCLK=256f_s$, $\text{Gain}=20\text{dB}$, ROHM 4-layer Board)

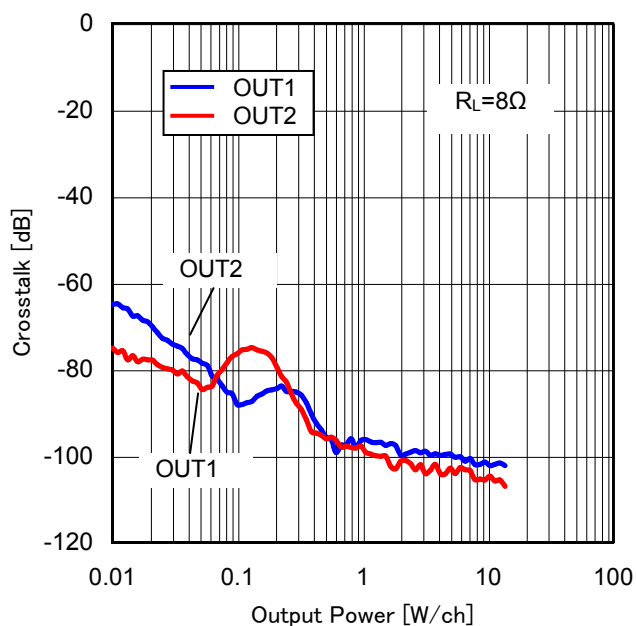


Figure 25. Crosstalk vs Output Power (8Ω)

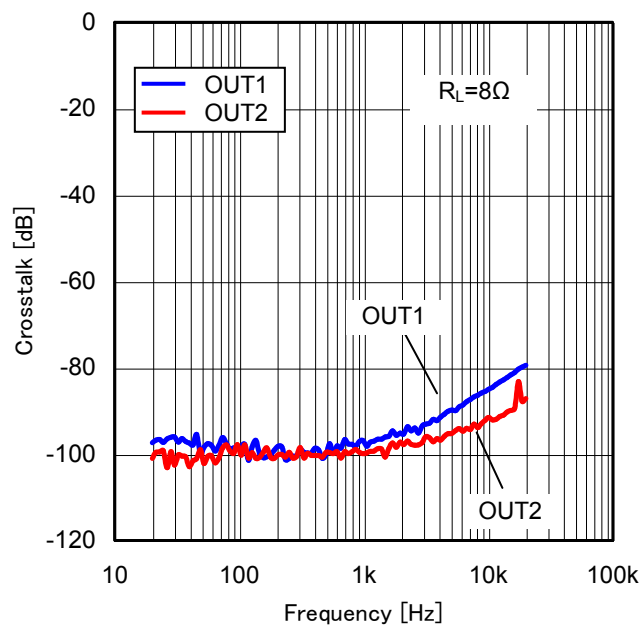


Figure 26. Crosstalk vs Frequency (8Ω)

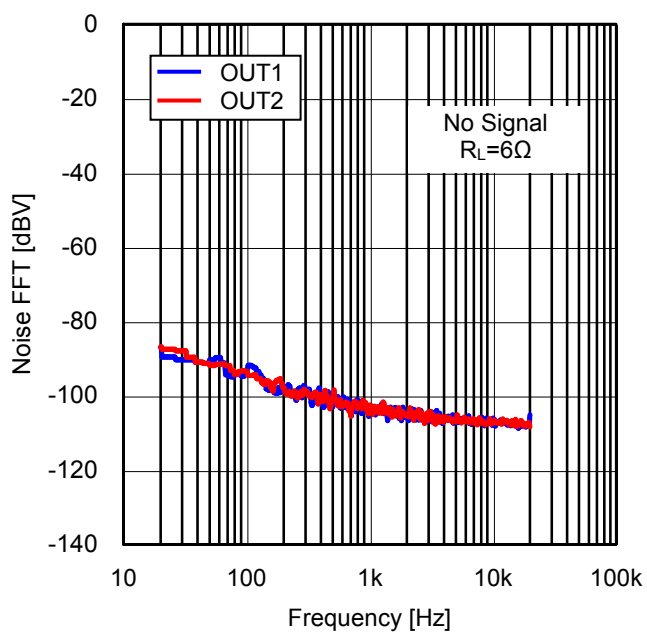


Figure 27. FFT of output noise voltage (6Ω)

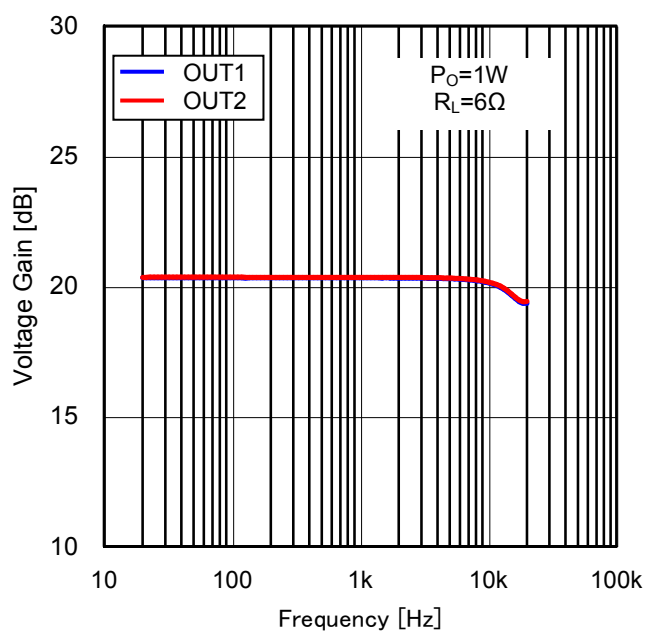


Figure 28. Voltage Gain vs Frequency (6Ω)

Typical Performance Curves – continued (7/11)

(Unless otherwise specified, $T_a=25^\circ\text{C}$, $V_{CC}=18\text{V}$, $f=1\text{kHz}$, $R_L=6\Omega$, $R_{STX}=3.3\text{V}$, $MUTEX=3.3\text{V}$, $f_s=48\text{kHz}$, $MCLK=256f_s$, $\text{Gain}=20\text{dB}$, ROHM 4-layer Board)

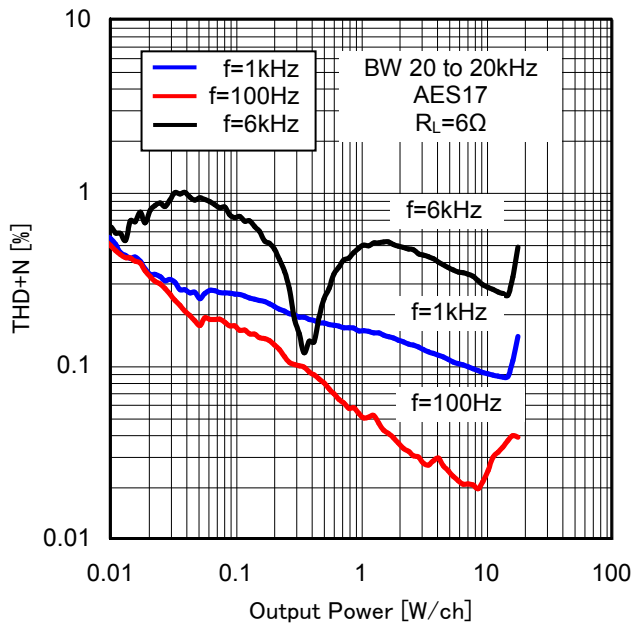


Figure 29. THD+N vs Output Power (6Ω)

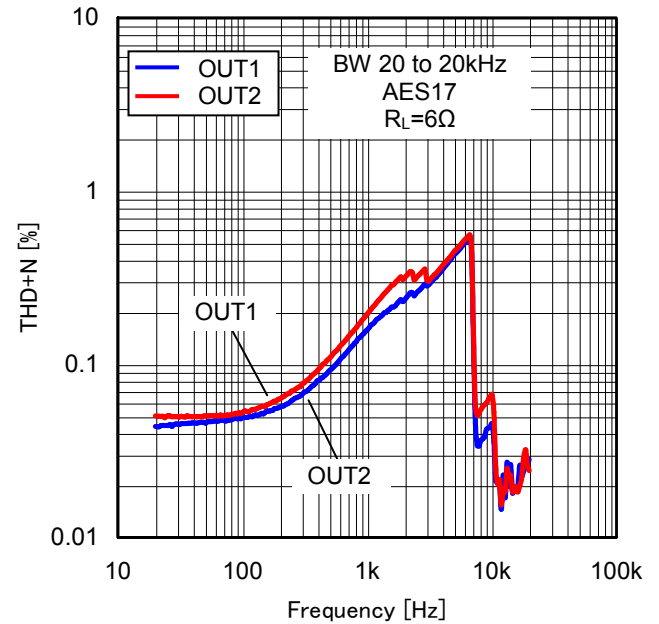


Figure 30. THD+N vs Frequency (6Ω)

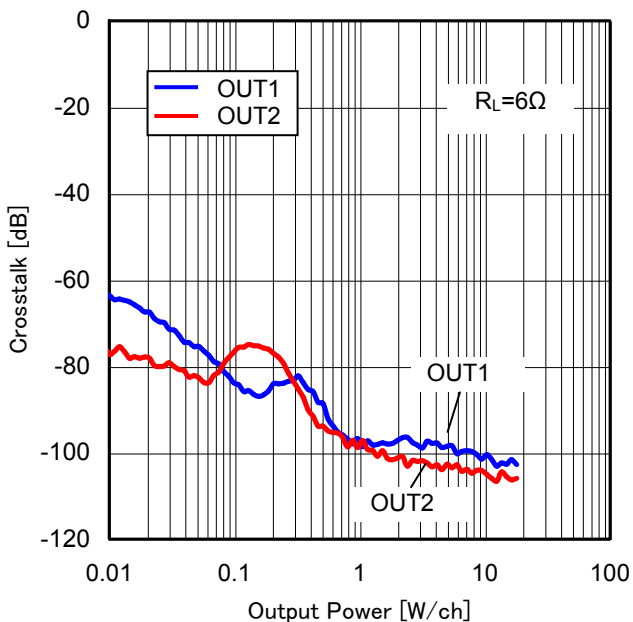


Figure 31. Crosstalk vs Output Power (6Ω)

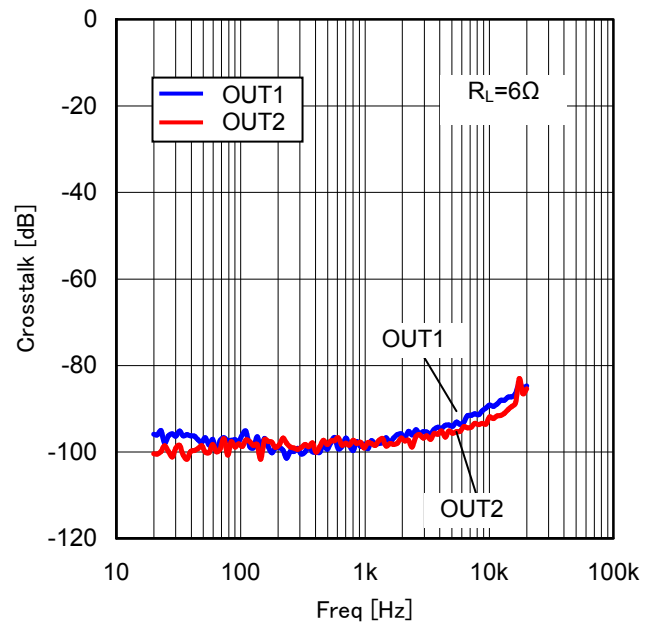


Figure 32. Crosstalk vs Frequency (6Ω)

Typical Performance Curves – continued (8/11)

(Unless otherwise specified, $T_a=25^\circ\text{C}$, $V_{CC}=12\text{V}$, $f=1\text{kHz}$, $R_L=4\Omega$, $R_{STX}=3.3\Omega$, $MUTEX=3.3\text{V}$, $f_s=48\text{kHz}$, $MCLK=256f_s$, $\text{Gain}=20\text{dB}$, ROHM 4-layer Board)

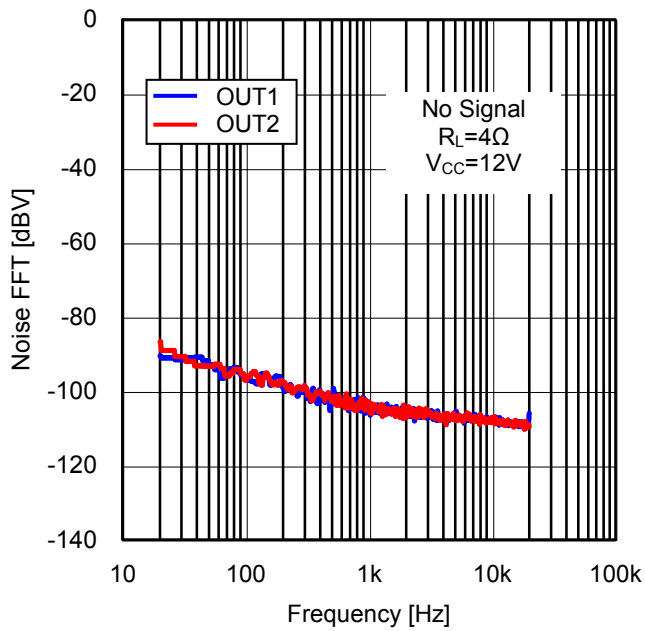


Figure 33. FFT of output noise voltage (4Ω)

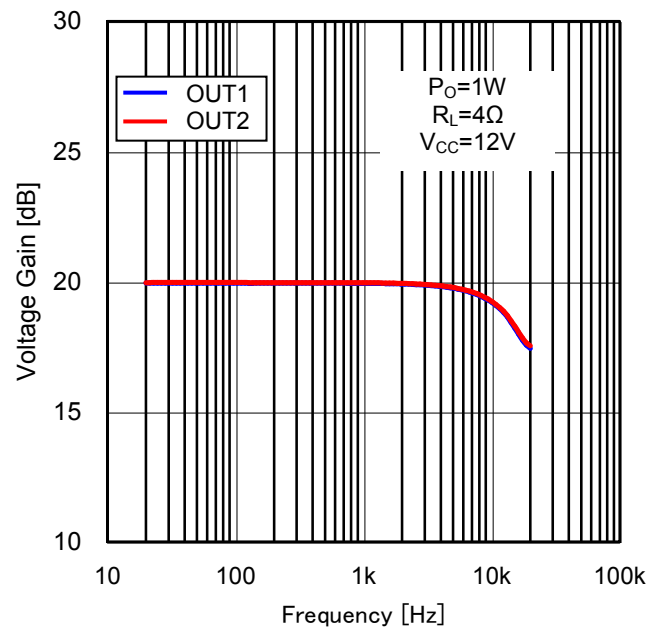


Figure 34. Voltage Gain vs Frequency (4Ω)

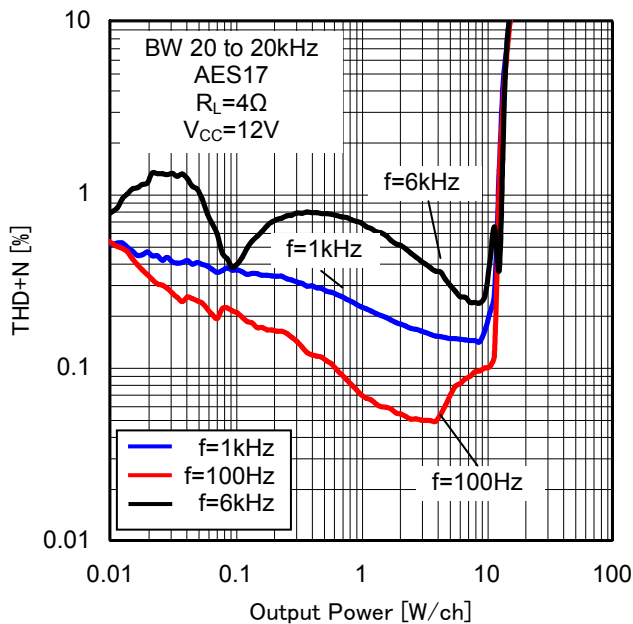


Figure 35. THD+N vs Output Power (4Ω)

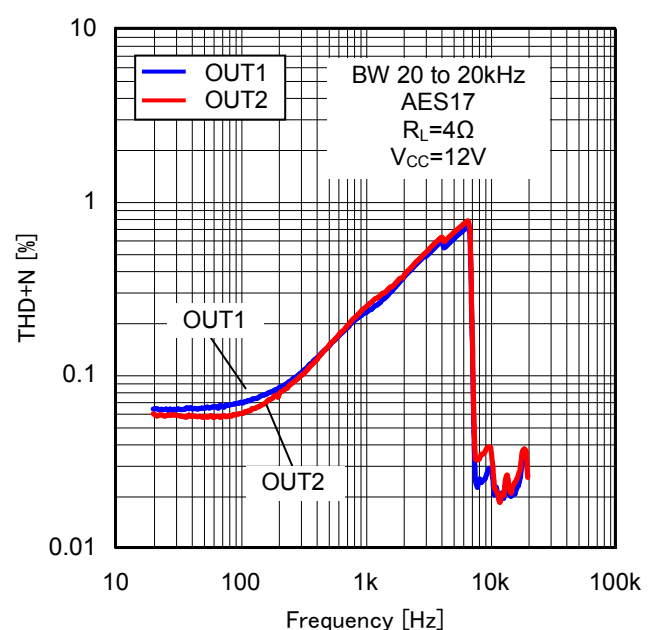


Figure 36. THD+N vs Frequency (4Ω)

Typical Performance Curves - continued (9/11)

(Unless otherwise specified, $T_a=25^\circ\text{C}$, $V_{CC}=18\text{V}$, $f=1\text{kHz}$, $R_L=4\Omega/4.8\Omega$, $R_{STX}=3.3\Omega$, $MUTEX=3.3\text{V}$, $f_s=48\text{kHz}$, $MCLK=256f_s$, $\text{Gain}=20\text{dB}$, ROHM 4-layer Board)

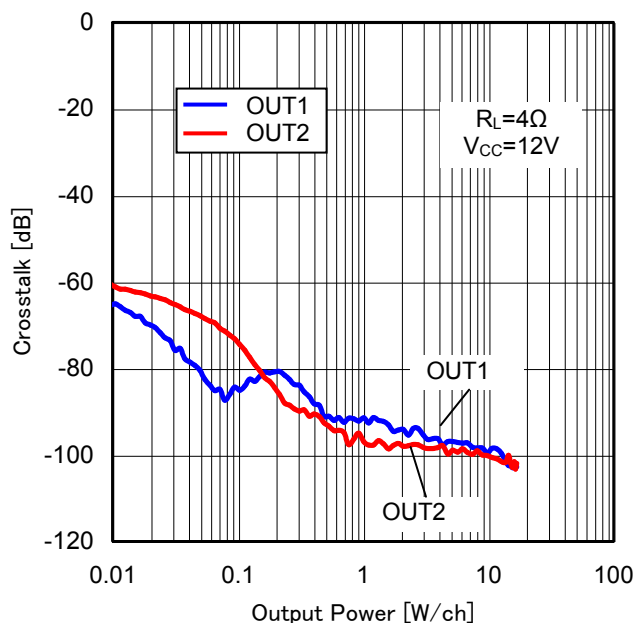


Figure 37. Crosstalk vs Output Power (4Ω)

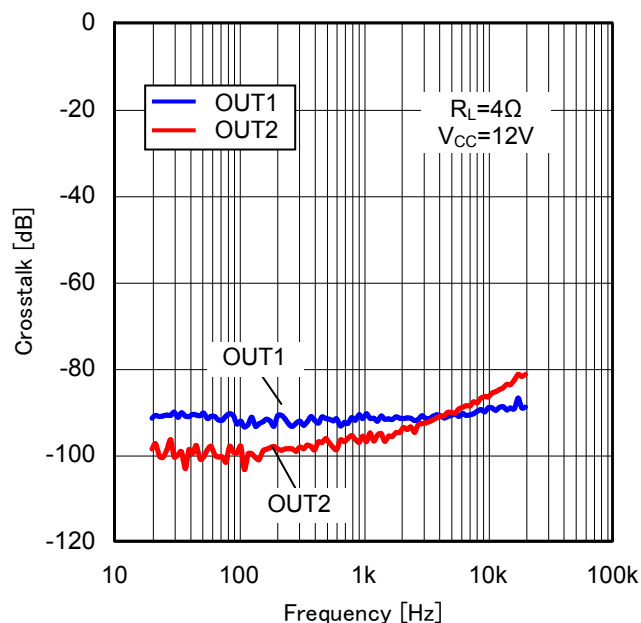


Figure 38. Crosstalk vs Frequency (4Ω)

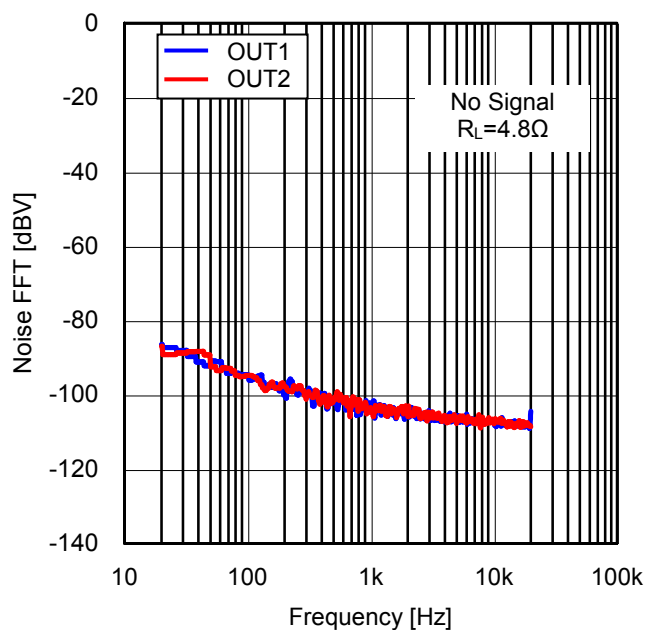


Figure 39. FFT of output noise voltage (4.8Ω)

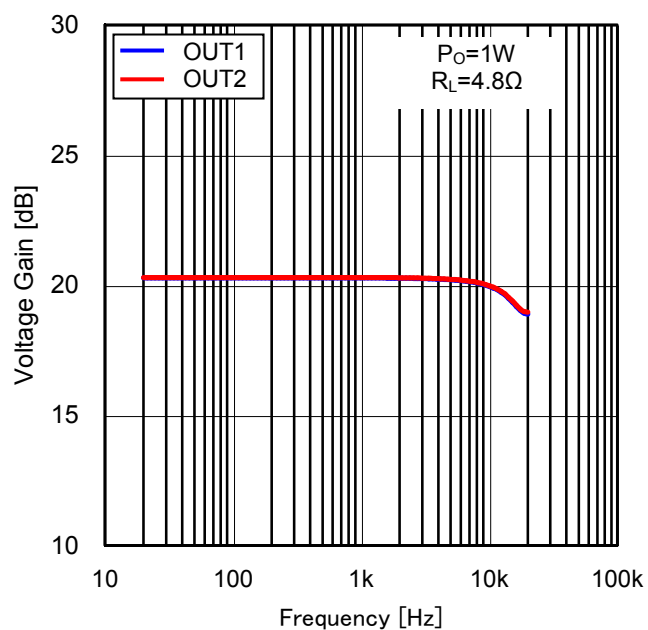


Figure 40. Voltage Gain vs Frequency (4.8Ω)

Typical Performance Curves – continued (10/11)

(Unless otherwise specified, $T_a=25^\circ\text{C}$, $V_{CC}=18\text{V}$, $f=1\text{kHz}$, $R_L=4.8\Omega$, $R_{STX}=3.3\text{V}$, $MUTEX=3.3\text{V}$, $f_s=48\text{kHz}$, $MCLK=256f_s$, $\text{Gain}=20\text{dB}$, ROHM 4-layer Board)

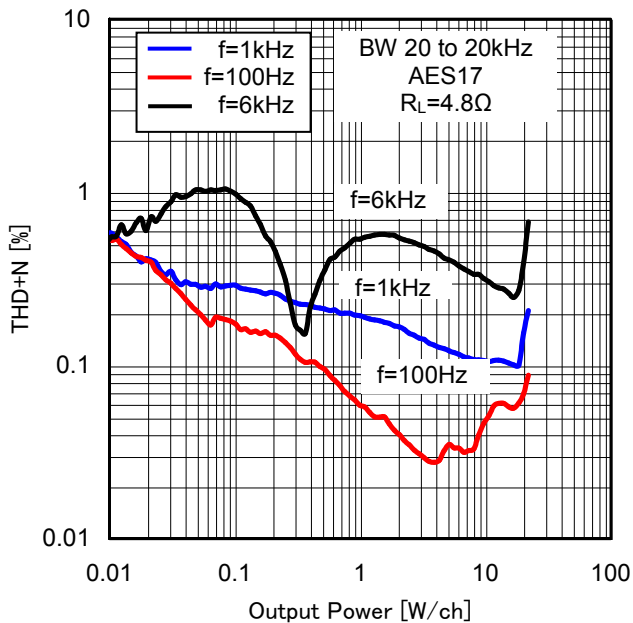


Figure 41. THD+N vs Output Power (4.8Ω)

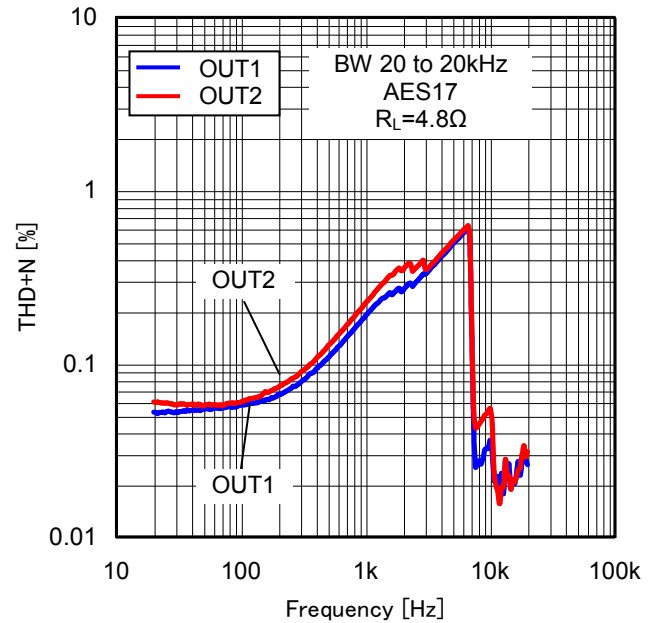


Figure 42. THD+N vs Frequency (4.8Ω)

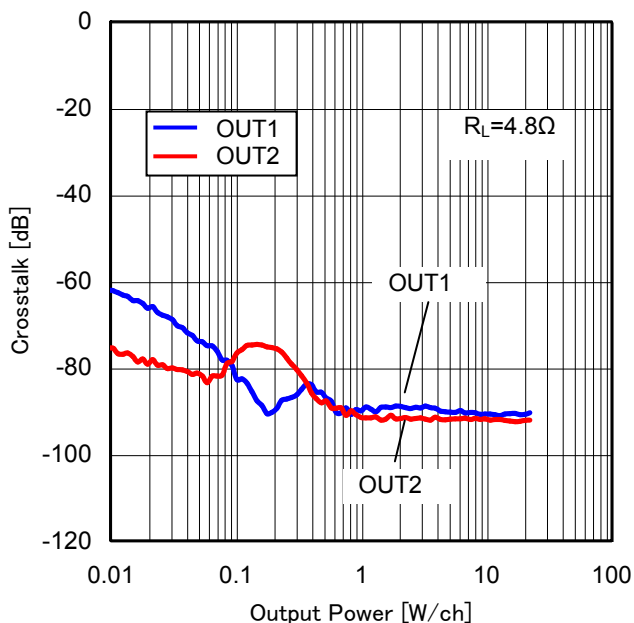


Figure 43. Crosstalk vs Output Power (4.8Ω)

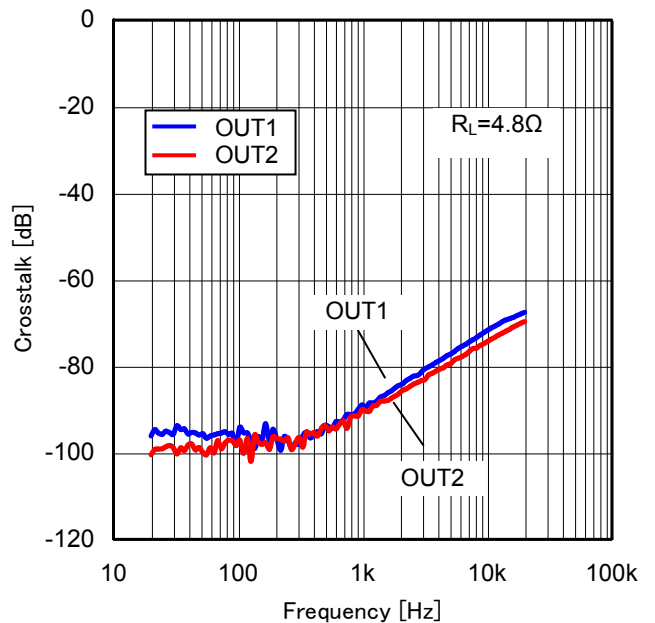


Figure 44. Crosstalk vs Frequency (4.8Ω)

Typical Performance Curves – continued (11/11)

(Unless otherwise specified, $T_a=25^\circ\text{C}$, $V_{CC}=18\text{V}$, $f=1\text{kHz}$, $R_L=8\Omega/6\Omega$, $R_{STX}=3.3\text{V}$, $MUTEX=3.3\text{V}$, $f_s=48\text{kHz}$, $MCLK=256f_s$, ROHM 4-layer Board)

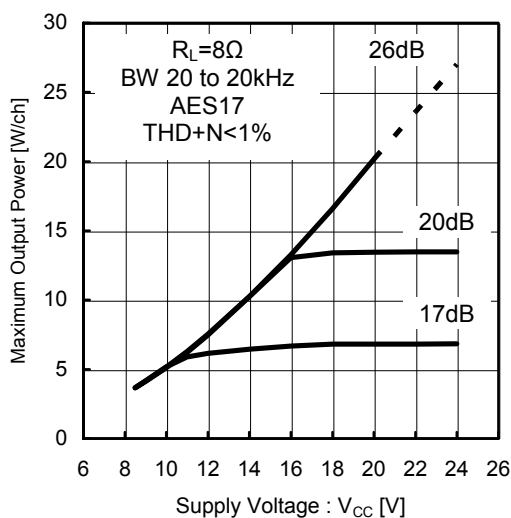


Figure 45. Supply Voltage vs Maximum Output Power (8Ω)

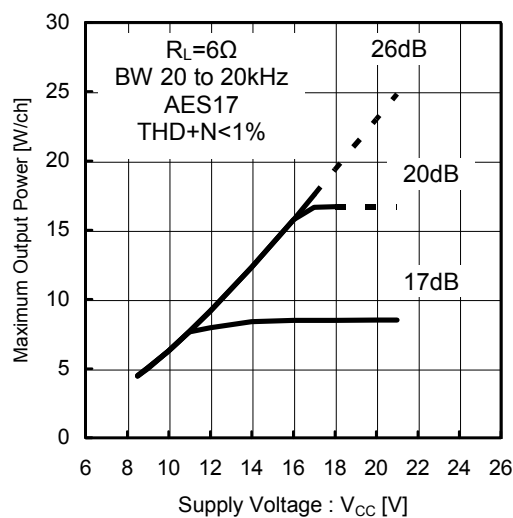


Figure 46. Supply Voltage vs Maximum Output Power (6Ω)

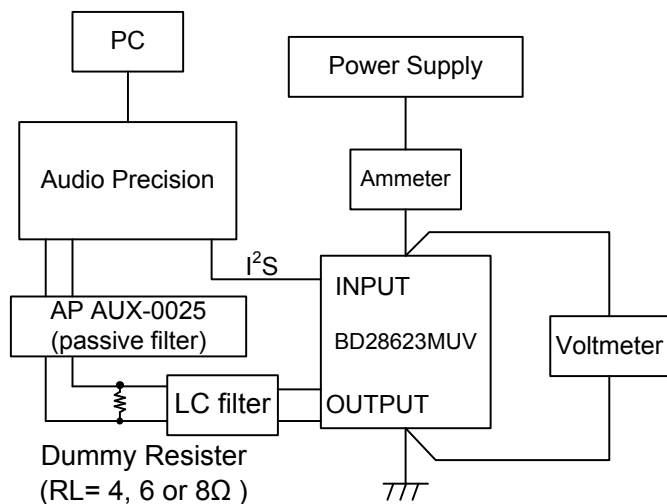


Figure 47. Audio Characteristics Measurement Environment

※ Dotted line means power dissipation is exceeded.

Timing Chart

1. Power Supply Start-up Sequence

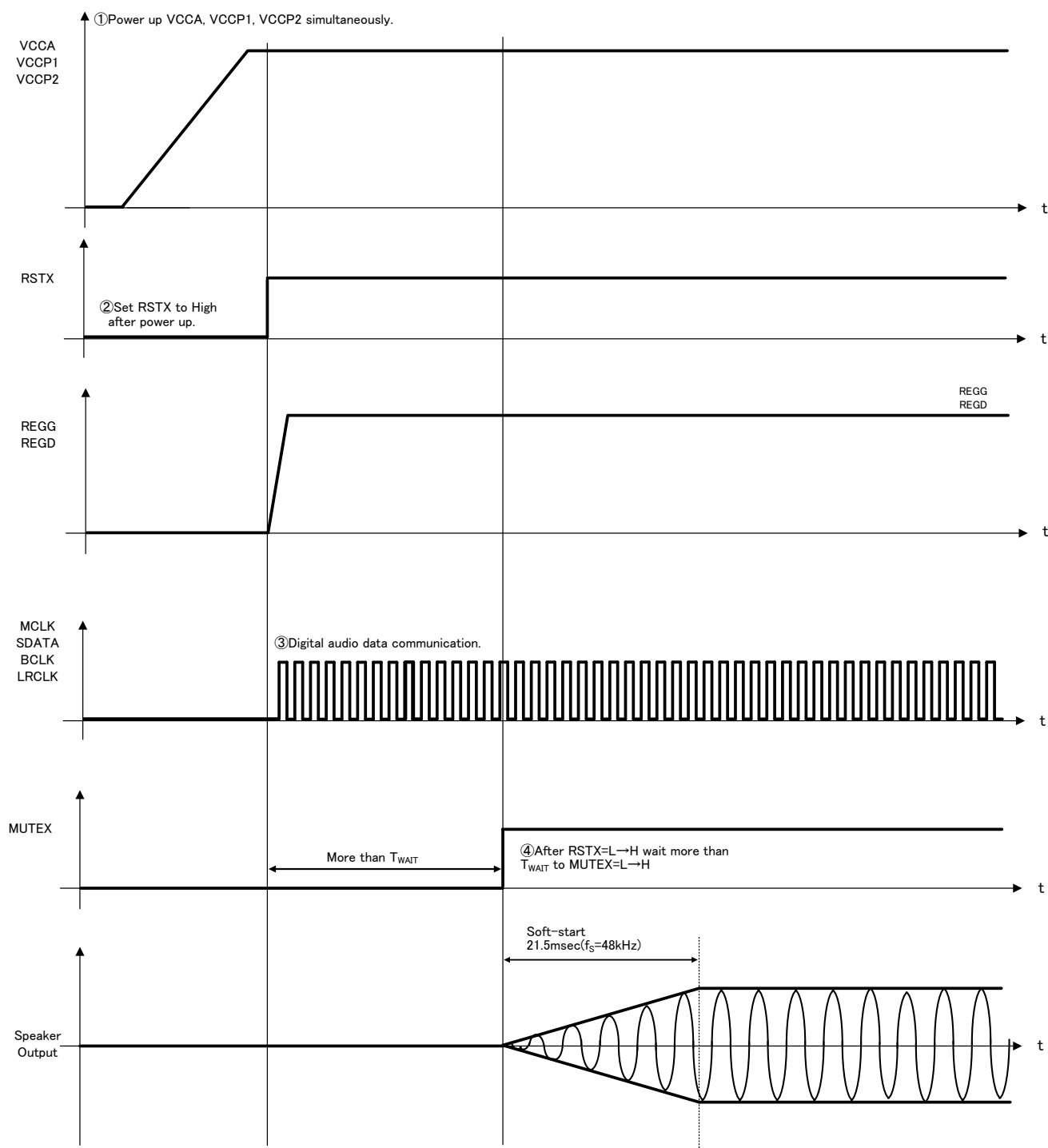


Figure 48. Power Supply Start-up Sequence

Caution: To eliminate pop noise when power supply is turned ON, RSTX and MUTEX should always be set Low. And also, all power supply terminals should start up together.
Order of ② and ③ can be interchange

BSP Capacitor Value (C9, C12, C19, C22)	Limit of T_{WAIT}			Unit
	Min	Typ	Max	
3.3 μ F	300	-	-	msec
4.7 μ F	400	-	-	msec

2. Power Supply Shutdown Sequence

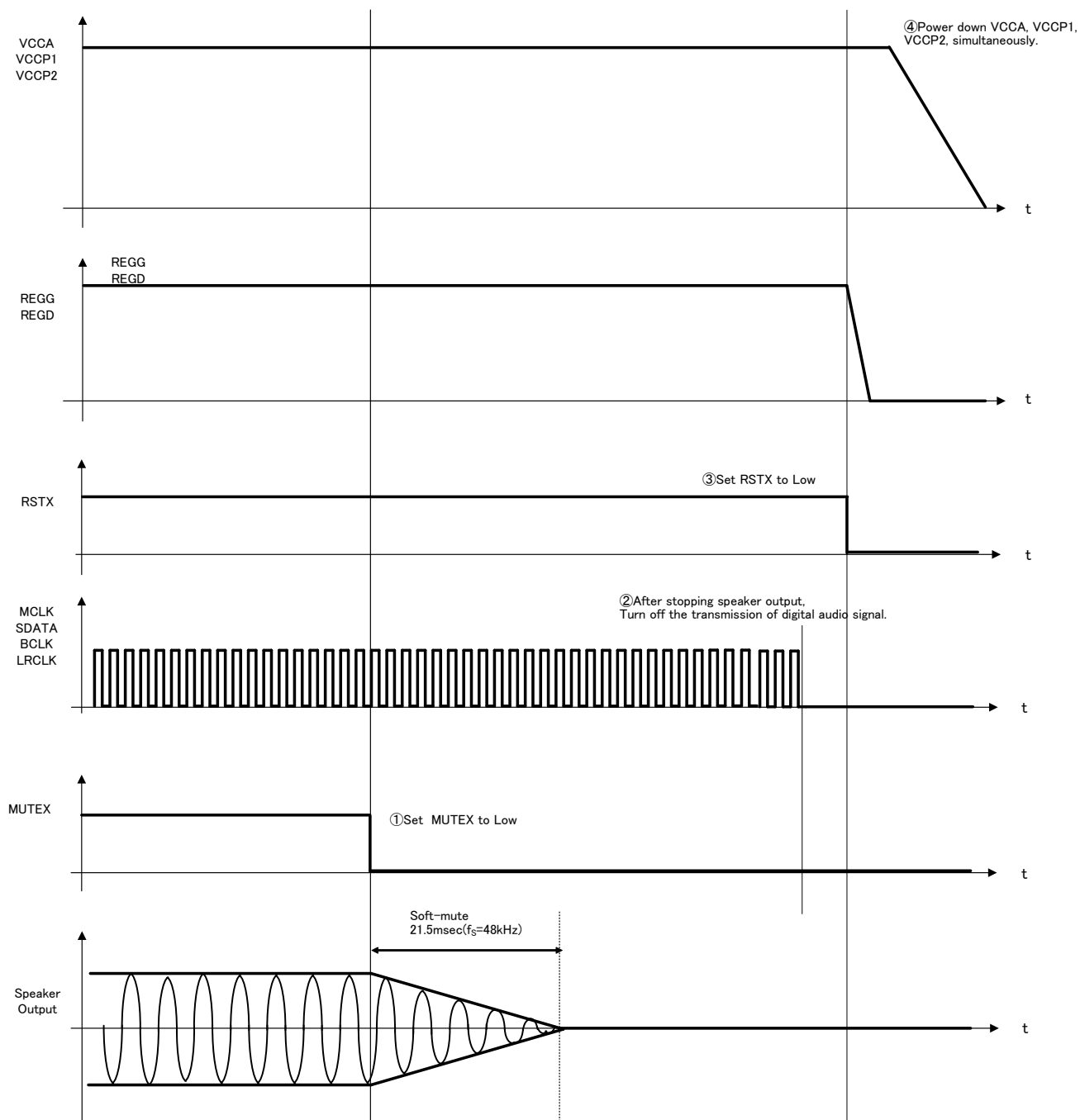


Figure 49. Power Supply Shutdown Sequence

Caution: To eliminate pop noise when power supply is turned OFF, RSTX and MUTE should always be set Low first. And also, all power supply terminals should shut down together.

Order of ② and ③ can be interchanged

3. About Changing Audio Signal

Output PWM frequency is sixteen times the sampling frequency " f_s ".

Therefore, output PWM frequency will also become unstable if MCLK becomes unstable when switching channel or switching input. During unstable period, LC resonance may occur and short protection function may work.

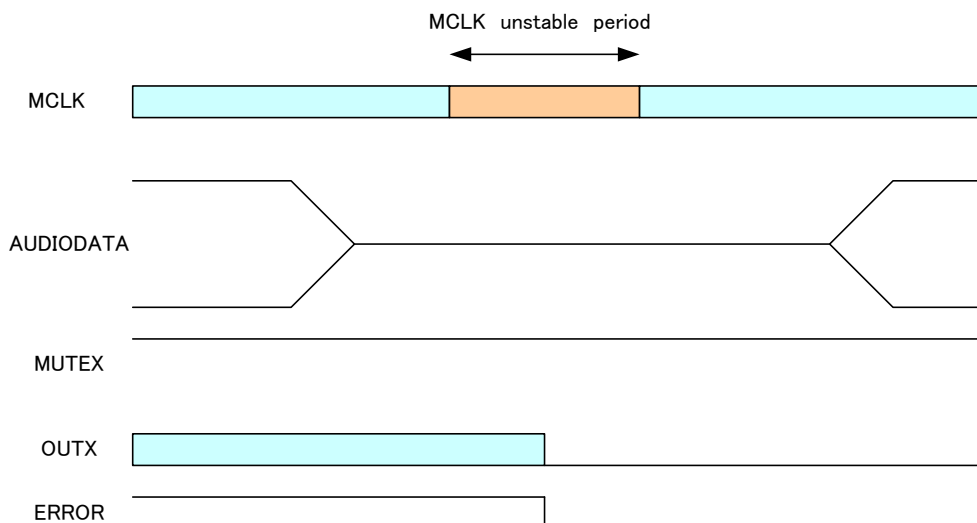
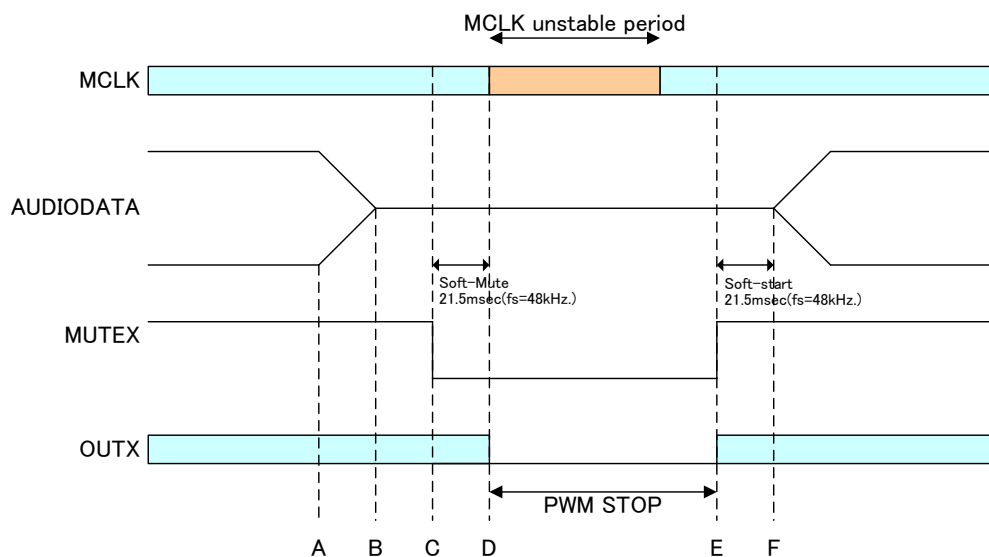


Figure 50. Action at MCLK Unstable 1

To prevent "MCLK unstable condition", please obey the following process.

- (1) Mute "AUDIODATA" from scaler IC. (A)
- (2) After muting "AUDIODATA" (B), set MUTEX=L (C).
- (3) After MCLK goes to stable state, set MUTEX=H (D).
- (4) Release mute "AUDIODATA" (E).



Order of E and F can be interchanged

Figure 51. Action at MCLK Unstable 2

Especially, if the “twice and more frequency compared with normality” is entered, for some timing, the incorrect data is set to the IC’s internal resistor and it generates noises continuously.

In case the “twice and more frequency compared with normality” is entered, please follow the timing chart bellow and add a reset sequence.

(Please release reset after MCLK (BCLK) becomes stable, then release mute of BD28623MUV.)

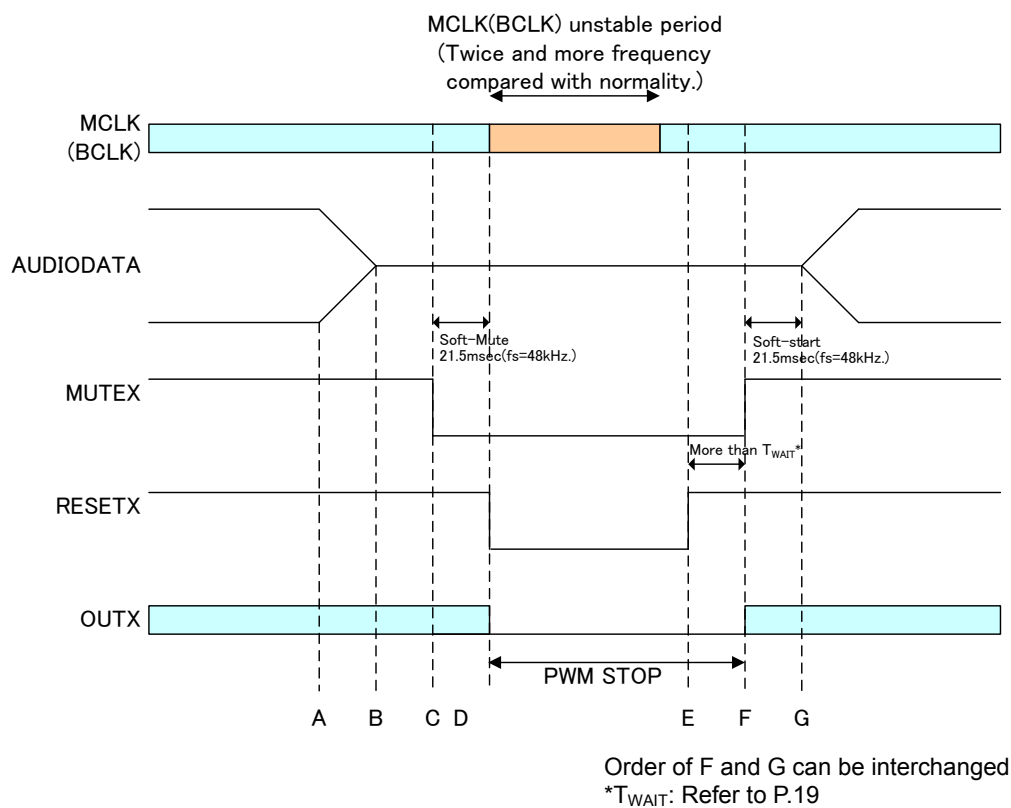


Figure 52. Action at MCLK Unstable 3

4. Recovery Sequence from the Instantaneous Power Supply Interruption

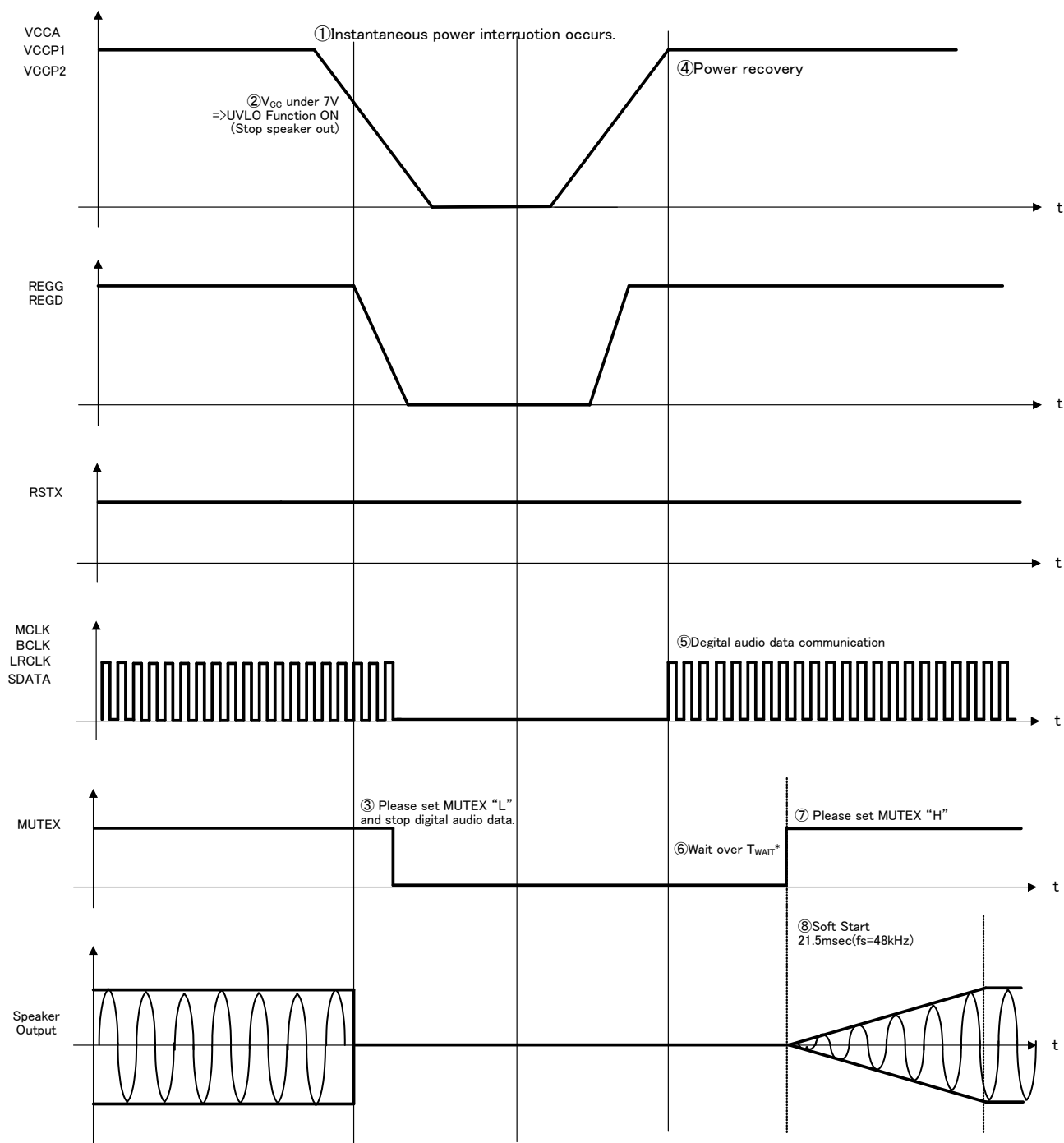


Figure 53. Instantaneous Power Interruption Recovery Sequence

* T_{WAIT} : Refer to P.19

Application Information

1. About digital audio input

(1) Input digital audio signal sampling frequency (f_s)

PWM frequency, Soft-start time, Soft-mute time, and the detection time of the DC voltage protection in the speaker depend on the sampling frequency (f_s) of the digital audio input.

Sampling Frequency of the Digital Audio Input (f_s)	PWM Frequency (f_{PWM})	Soft-start / Soft-mute Time	DC Voltage Protection in the Speaker Detection Time
32kHz	512kHz	32msec	1.02sec
44.1kHz	705.6kHz	23msec	0.74sec
48kHz	768kHz	21.5msec	0.68sec

(2) Format of digital audio input

MCLK: System Clock input signal

It will input LRCLK, BCLK, SDATA that synchronizes with this clock. MCLK frequency is 256 times the sampling frequency ($256f_s$) or 512 times the sampling frequency ($512f_s$).

LRCLK: L/R Clock input signal

It corresponds to 32kHz/44.1kHz/48kHz clock (f_s) which are same to the sampling frequency (f_s). The audio data of left and right channel for one sample is input to this section.

BCLK: Bit Clock input signal

It is used to latch data per bit using 64 times the sampling frequency ($64f_s$).

SDATA: Data input signal

It is amplitude data. The data length is different according to the resolution of the input digital audio data. It corresponds to 16/ 20/ 24 bits.

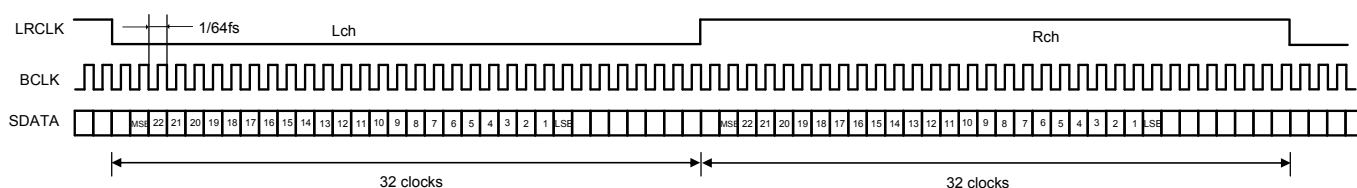
(3) I²S Data Format

Figure 54. I2S Data Format 64fs, 24bit Data

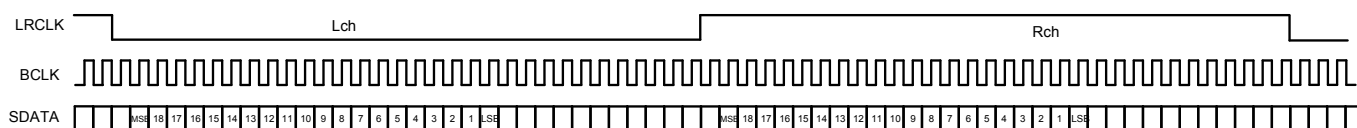


Figure 55. I2S Data Format 64fs, 20bit Data

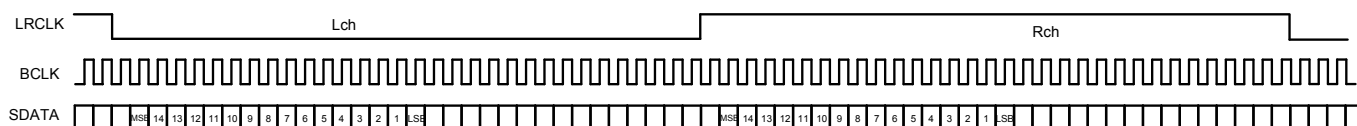


Figure 56. I2S Data Format 64fs, 16bit Data

The Low section of LRCLK becomes Lch and the High section of LRCLK becomes Rch. After changing LRCLK, second bit becomes MSB.

(4) Audio Interface Format and Timing

Recommended timing and operating condition (MCLK, BCLK, LRCLK and SDATA)

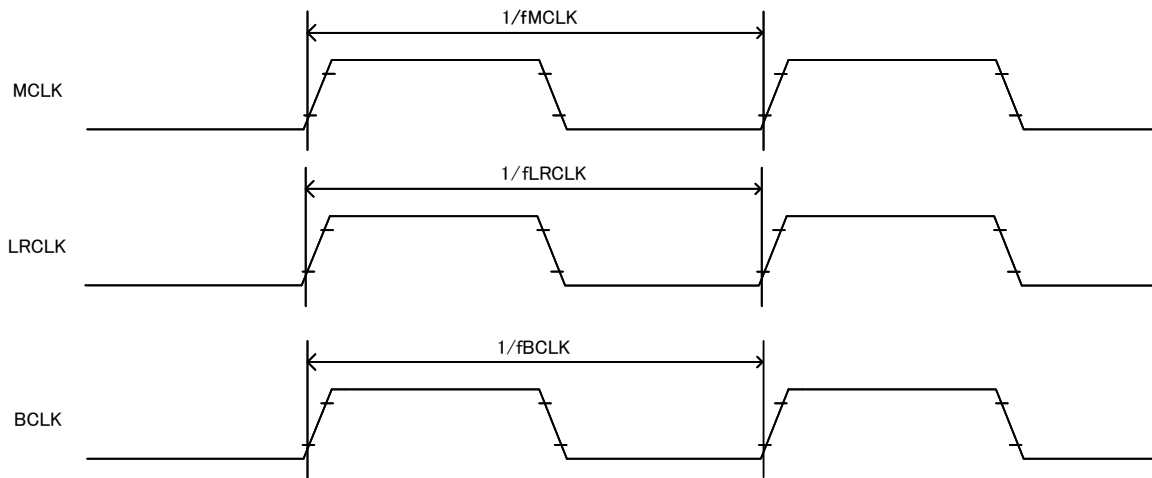


Figure 57. Clock Timing

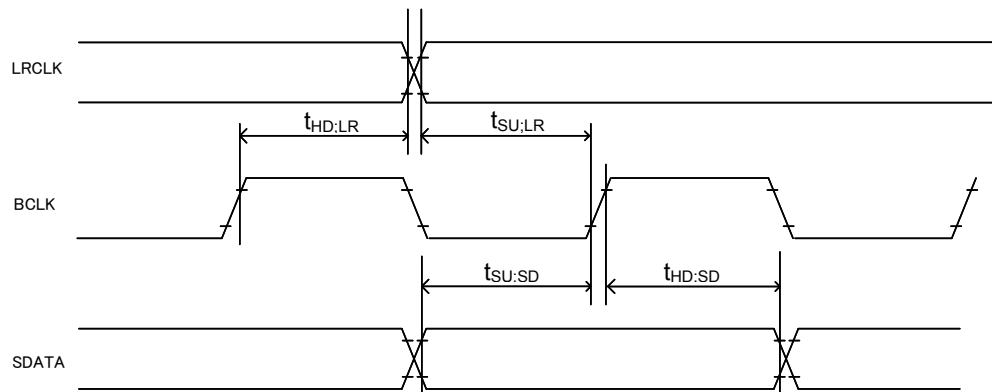


Figure 58. Audio Interface Timing

No.	Parameter	Symbol	Limit				Unit
			MCLK=256f _s		MCLK=512f _s		
			Min	Max	Min	Max	
1	MCLK Frequency ^(Note 8-1)	f _{MCLK}	8.192 ±10%	12.288 ±10%	16.384 ±10%	24.576 ±10%	MHz
2	LRCLK Frequency ^(Note 8-1)	f _{LRCLK}	32 ±10%	48 ±10%	32 ±10%	48 ±10%	kHz
3	BCLK Frequency ^(Note 8-1)	f _{BCLK}	2.048 ±10%	3.072 ±10%	2.048 ±10%	3.072 ±10%	MHz
4	Setup Time, LRCLK ^(Note 8-2)	t _{SU;LR}	20	—	20	—	ns
5	Hold Time, LRCLK ^(Note 8-2)	t _{HD;LR}	20	—	20	—	ns
6	Setup Time, SDATA	t _{SU;SD}	20	—	20	—	ns
7	Hold Time, SDATA	t _{HD;SD}	20	—	20	—	ns
8	MCLK, DUTY	d _{MCLK}	40	60	40	60	%
9	LRCLK, DUTY	d _{LRCLK}	40	60	40	60	%
10	BCLK, DUTY	d _{BCLK}	40	60	40	60	%

(Note 8-1) Must be synchronized with BCLK, LRCLK

(Note 8-2) This regulation is to keep rising edge of LRCLK and rising edge of BCLK from overlapping.

1. Terminal Setting

1) RSTX Pin, MUTEX Pin Function

Condition	RSTX	MUTEX	Normal		Error Detection	
			PWM Outputs (OUT1P, 1N, 2P, 2N)	ERROR	PWM Outputs (OUT1P, 1N, 2P, 2N)	ERROR
“RESET”(Note 9)	L	L/H	High-Z_Low ^(Note 10) (Reset mode)	H	High-Z_Low (Reset mode)	H
“MUTE”	H	L	High-Z_Low (MUTE_ON)	H	High-Z_Low (MUTE_ON)	L
“ACTIVE”	H	H	Active (MUTE_OFF)	H	High-Z_Low (MUTE_ON)	L

(Note 9) If RSTX is set Low, internal registers (I²S / I/F part, ×8 over sampling digital filter part, latch circuit when detecting ERROR) are initialized.

(Note 10) This means that all power transistors are OFF and output terminals are pulled down by 40kΩ (Typ).

2) GAIN Pin Function

GAIN terminal sets the gain. Gain setting limits maximum output power.

GAIN setting depends on the value of speaker load, because maximum output power depends on speaker load.

Please set GAIN after setting MUTE to L. Pop noise may occur if GAIN is set while MUTE=H.

GAIN	Gain Setting (BTL)	Output Power
L	17dB	Min 5 W (at 8Ω)
Pull-up (3.3V) to 47kΩ (1/16W, J (±5%))	20dB	Min 10 W (at 8Ω)
H	26dB	-

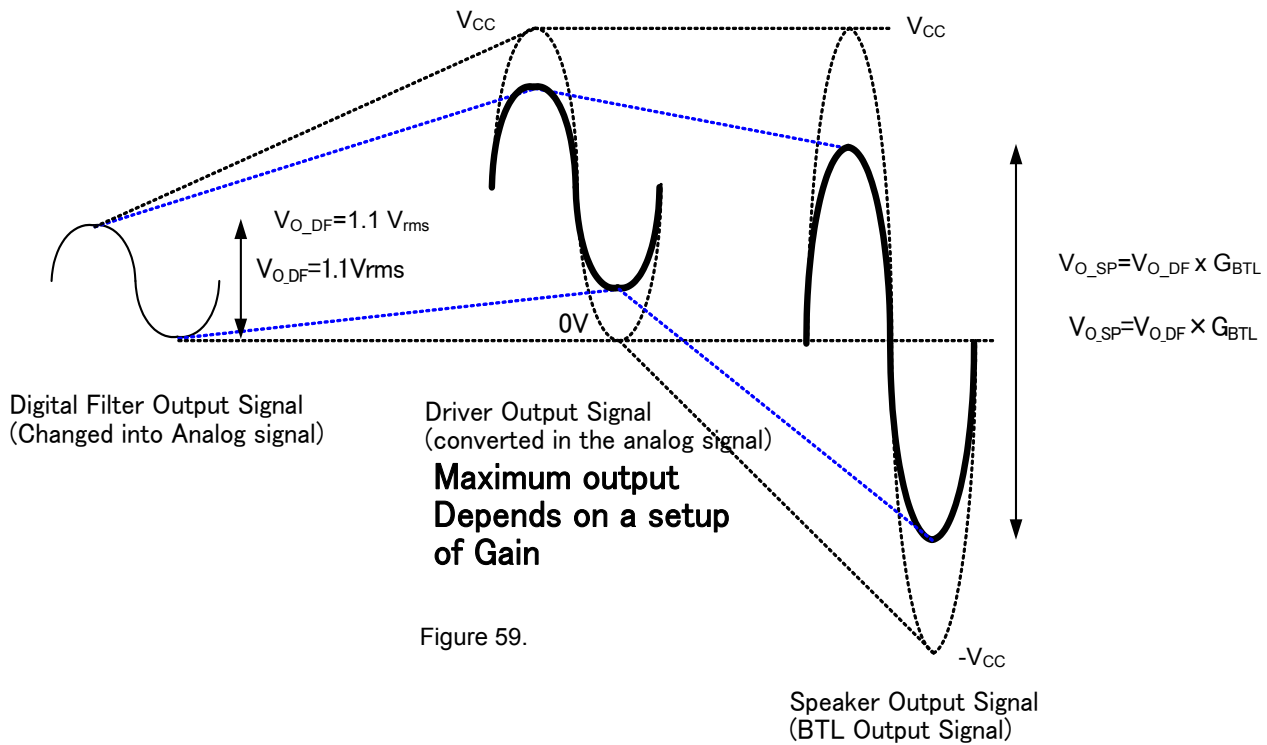


Figure 59.

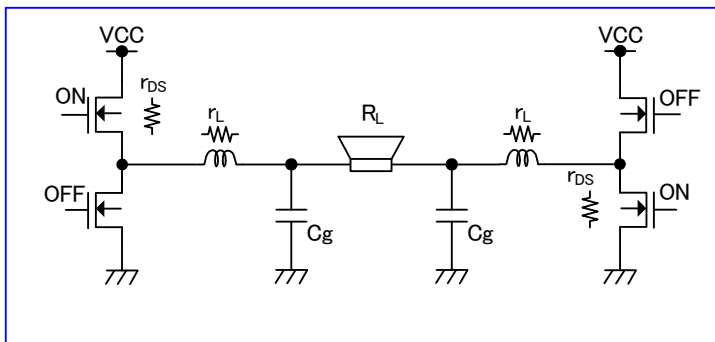


Figure 60. Schematic of Output Equivalent

$$V_{O_SP} = \left(10^{\frac{V_{IN}}{20}} \right) \times 10^{\left(\frac{G_{BTL}}{20} \right)} \times \frac{R_L}{(r_{DS} + 2r_{DC}) + R_L} \quad [V_{rms}]$$

$$P_{O(THD \leq 1\%)} = \frac{\left[\left(10^{\frac{V_{IN}}{20}} \right) \times 10^{\left(\frac{G_{BTL}}{20} \right)} \times \frac{R_L}{(r_{DS} + 2r_{DC}) + R_L} \right]^2}{R_L} \quad [W]$$

where:

V_{IN} is the I^2S input level [dBFS]

G_{BTL} is the gain setting [dB]

R_L is the load resistance [Ω]

r_{DS} is the resistance of FET [Ω]

(Typ=0.23 Ω)

r_{DC} is the DC resistance of inductor [Ω]

2. About the Protection Function

Protection Function	Detecting & Releasing Condition		PWM Output OUT1P, 1N, 2P, 2N	ERROR
Output Short Protection	Detecting condition	Detecting current = 8A (Typ) /5A (Min. Tj=85°C)	High-Z_Low (Latch)	L (Latch)
DC Voltage Protection in the Speaker	Detecting condition	At speaker output, impressed DC voltage over 0.68sec (f _S =48kHz) Over 3.5V (Gain=26dB) Over 1.75V (Gain=20dB) Over 1.225V (Gain=17dB)	High-Z_Low (Latch)	L (Latch)
Overheat Protection	Detecting condition	Chip temperature above 150°C(Min.)	High-Z_Low	L
	Releasing condition	Chip temperature below 120°C(Min.)	Normal operation	
Under Voltage Protection	Detecting condition	Power supply voltage below 7V (Typ)	High-Z_Low	H
	Releasing condition	Power supply voltage above 7.5V (Typ)	Normal operation	
Clock Stop Protection	Detecting condition	No change in MCLK for more than 1μsec (Typ) or	High-Z_Low	H
		No change in BCLK for more than 1μsec (Typ) or		
		No change in LRCLK for more than 21μsec (at f _S =48kHz.).		
	Releasing condition	Normal input to MCLK, BCLK and LRCLK.	Normal operation	

(Note) The ERROR pin is Nch open-drain output. ERROR pin is pulled up by 100kΩ resistor.

(Note) Once an IC is latched, the circuit is not released automatically even after the detecting status is removed.

Procedure ① or ② is needed for recovery.

①MUTEX terminal is turned Low (holding time at Low = 10msec(Min)) then turned back to High again.

②Power supply is turned on again after dropping to V_{CC}<3V(10msec (Min) holding) in which the internal power ON reset circuit activates.

(Note) Please remove the DC component in SCALER IC of the preceding paragraph of this IC so that DC voltage protection feature not to aim at does not operate.

The High pass filter function for the DC component removal is not to BD28623MUV.

(1) Output Short Protection (Short to Power Supply)

This IC has PWM output short protection circuit that stops the PWM output when the Speaker output (after LC-filter) is short-circuited to the power supply due to wrong condition.

Detecting condition - It will detect when MUTE \bar{X} pin is set High and the current that flows in the PWM output pin becomes 8A(Typ) or more. The PWM output instantaneously enters the state of High-Z $_{Low}$ if detected, and the IC is latched.

Releasing method - ① After MUTE \bar{X} terminal is turned Low (holding time at Low = 10msec(Min)) then turned back to High again.

② Power supply is turned on again after dropping to $V_{CC} < 3V$ (10msec (Min) holding) in which the internal power ON reset circuit activates.

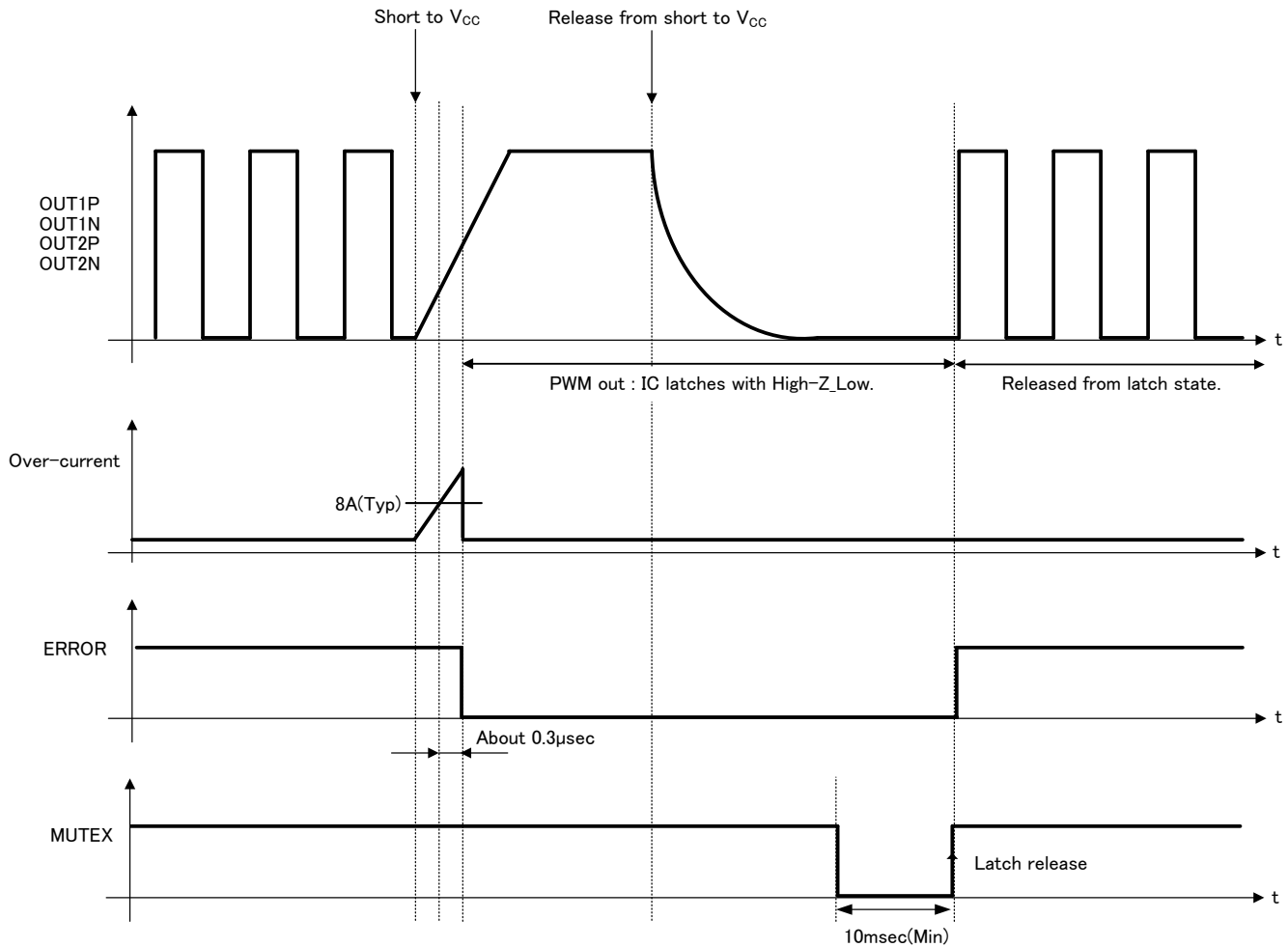


Figure 61. Output Short Protection (Short to Power Supply) Sequence

(2) Output Short Protection6 (Short to GND)

This IC has PWM output short protection circuit that stops the PWM output when the Speaker output (after LC-filter) is short-circuited to GND due to wrong condition.

Detecting condition - It will detect when MUTEX pin is set High and the current that flows in the PWM output terminal becomes 8A(Typ) or more. The PWM output instantaneously enters the state of High-Z_Low if detected, and the IC is latched.

Releasing method - ① After MUTEX terminal is turned Low (holding time at Low = 10msec(Min)) then turned back to High again.

② Power supply is turned on again after dropping to $V_{CC} < 3V$ (10msec (Min) holding) in which the internal power ON reset circuit activates.

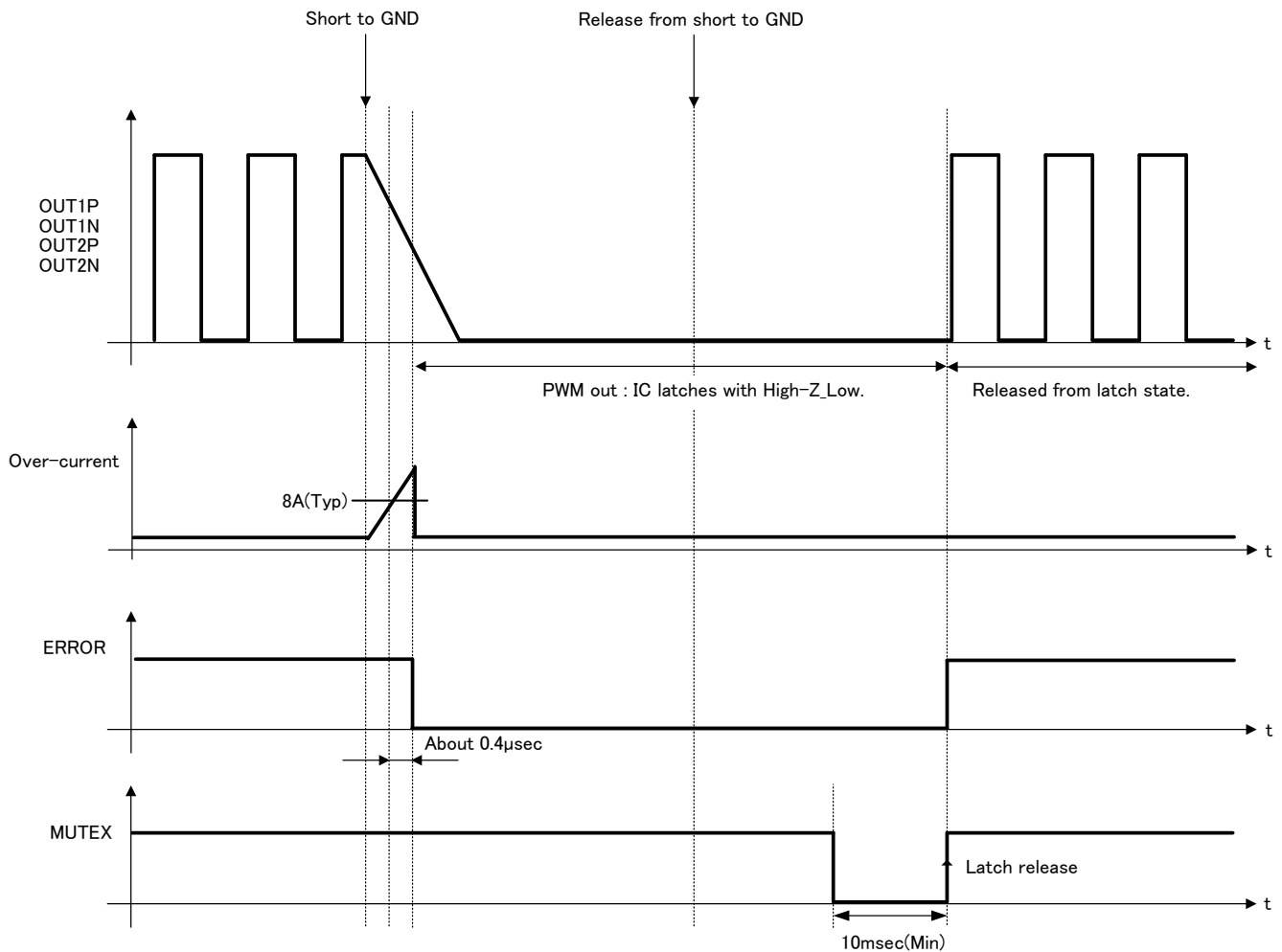


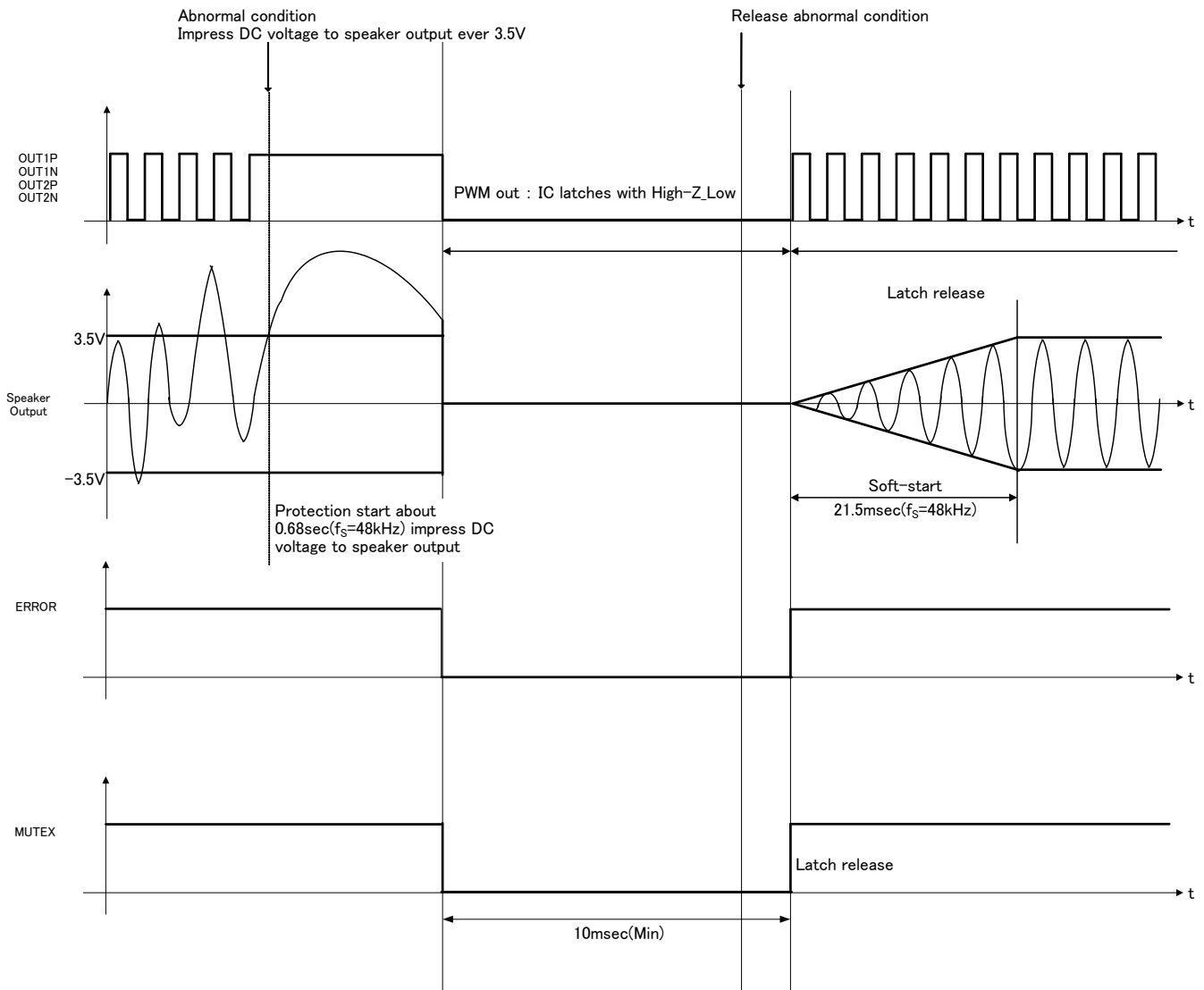
Figure 62. Output Short Protection (Short to GND) Sequence

(3) DC Voltage Protection

When DC voltage is applied to the speaker due to wrong condition, this IC has protection circuit where the speaker is protected from destruction.

Detecting condition - It will detect when MUTE pin is set High and speaker output is more than 3.5V (TYP, Gain=26dB setting), 1.75V (TYP, Gain=20dB setting), 1.225V (TYP, Gain=17dB setting), 0.68sec ($f_s=48\text{kHz}$) or above. Once detected, the PWM output instantaneously enters the state of High-Z_Low, and the IC is latched.

Releasing method -
 ① After MUTE pin is turned Low (holding time at Low = 10msec(Min)) then turned back to High again.
 ② Power supply is turned on again after dropping to $V_{CC}<3\text{V}$ (10msec (Min) holding) in which the internal power ON reset circuit activates.



(GAIN=26dB settings)

Figure 63. DC Voltage Protection Sequence

(Note) Please remove the DC component in SCALER IC of the preceding paragraph of this IC so that DC voltage protection feature not to aim at does not operate.

The High pass filter function for the DC component removal is not to BD28623MUV.

(4) Overheat Protection

This IC has the overheat protection circuit that prevents thermal runaway when the temperature of the chip exceeds $T_{jmax}=150^{\circ}\text{C}$.

Detecting condition - It will detect when MUTEX pin is set High and the temperature of the chip becomes 150°C (Min) or more. Speaker output turns MUTE immediately when high temperature protection is detected.

Releasing condition - It will release when MUTEX pin is set High and the temperature of the chip becomes 120°C (Min) or less. The speaker output is outputted through a soft-start when released. (Auto recovery)

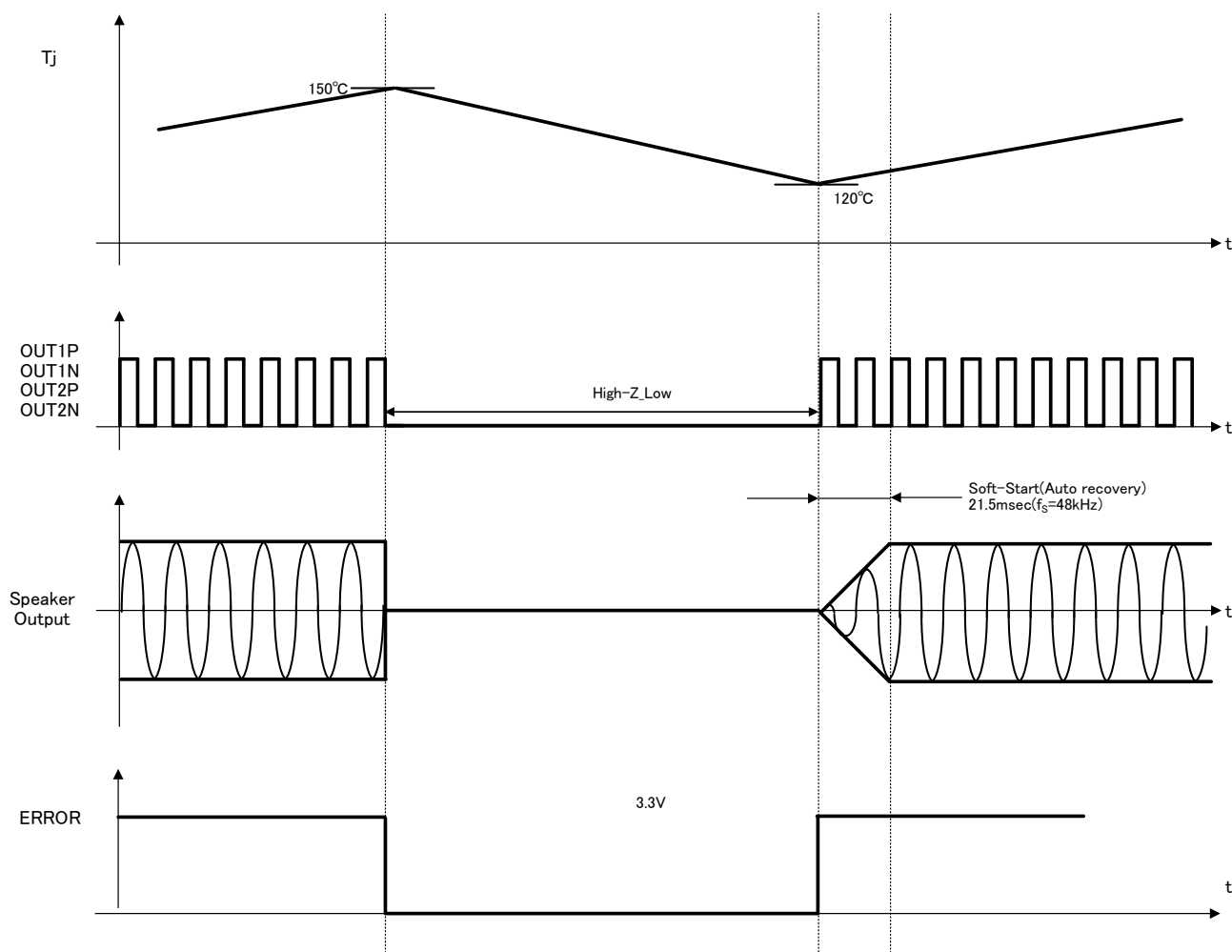


Figure 64. Overheat Protection Sequence

(5) Under Voltage Protection

This IC has under voltage protection circuit that mutes the speaker output mute once it detects extreme drop of the power supply voltage.

Detecting condition - It will detect when MUTE pin is set High and the power supply voltage becomes lower than 7V (Typ). Speaker output turns MUTE immediately when under voltage protection is detected.

Releasing condition - It will release when MUTE pin is set High and the power supply voltage becomes more than 7.5V (Typ). The speaker output is outputted through a soft-start when released. (Auto recovery)

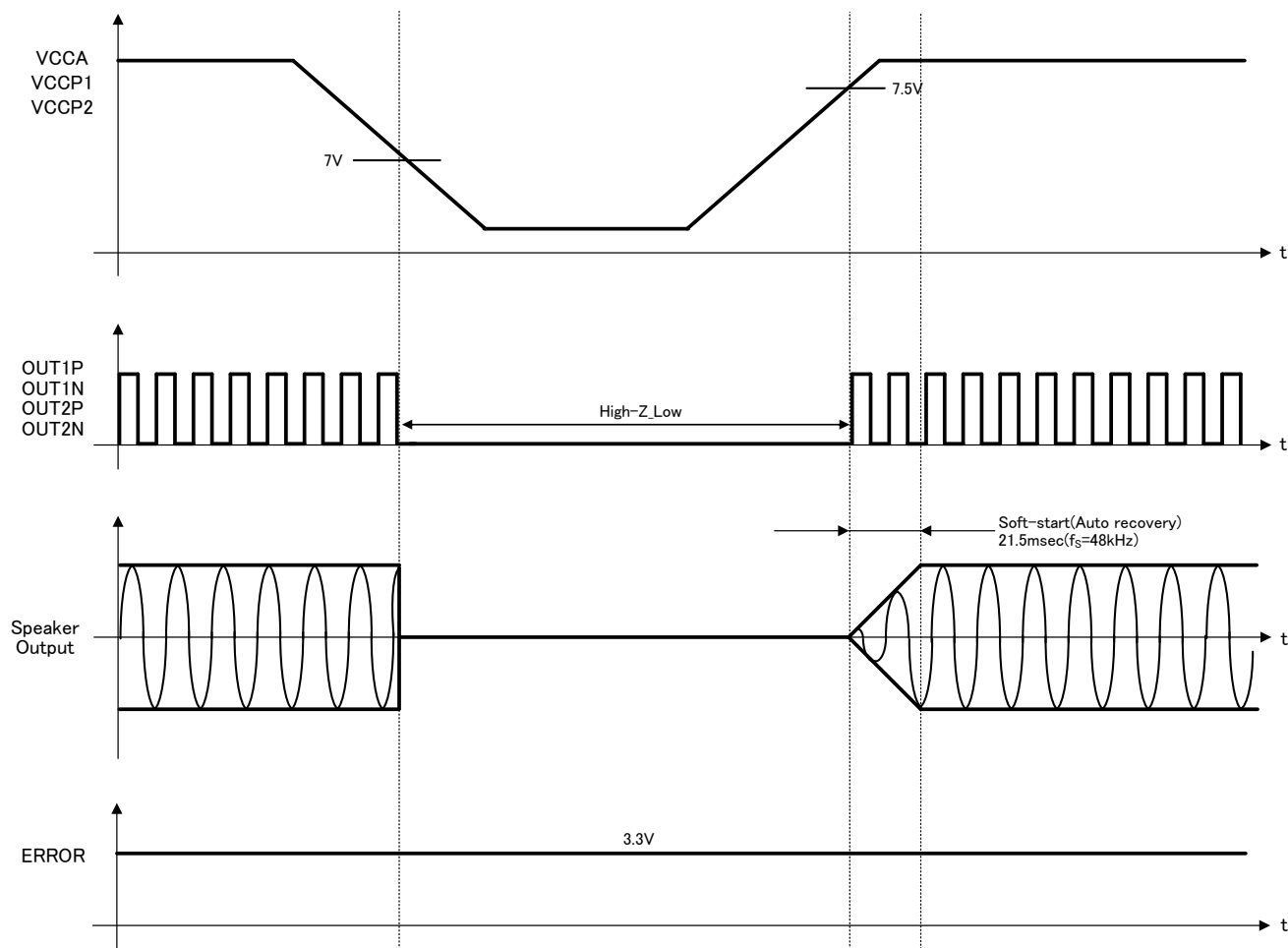


Figure 65. Under Voltage Protection Sequence

(6) Clock Stop Protection (MCLK)

This IC has clock stop protection circuit that mutes the speaker output when the MCLK signal of the digital audio input stops.

Detecting condition - It will detect when MUTE pin is set High and the MCLK signal stops for about 1 μ sec or more. Speaker output turns MUTE immediately when clock stop protection is detected.

Releasing condition - It will release when MUTE pin is set High and the MCLK signal returns to the normal clock operation. The speaker output is outputted through a soft-start when released. (Auto recovery)

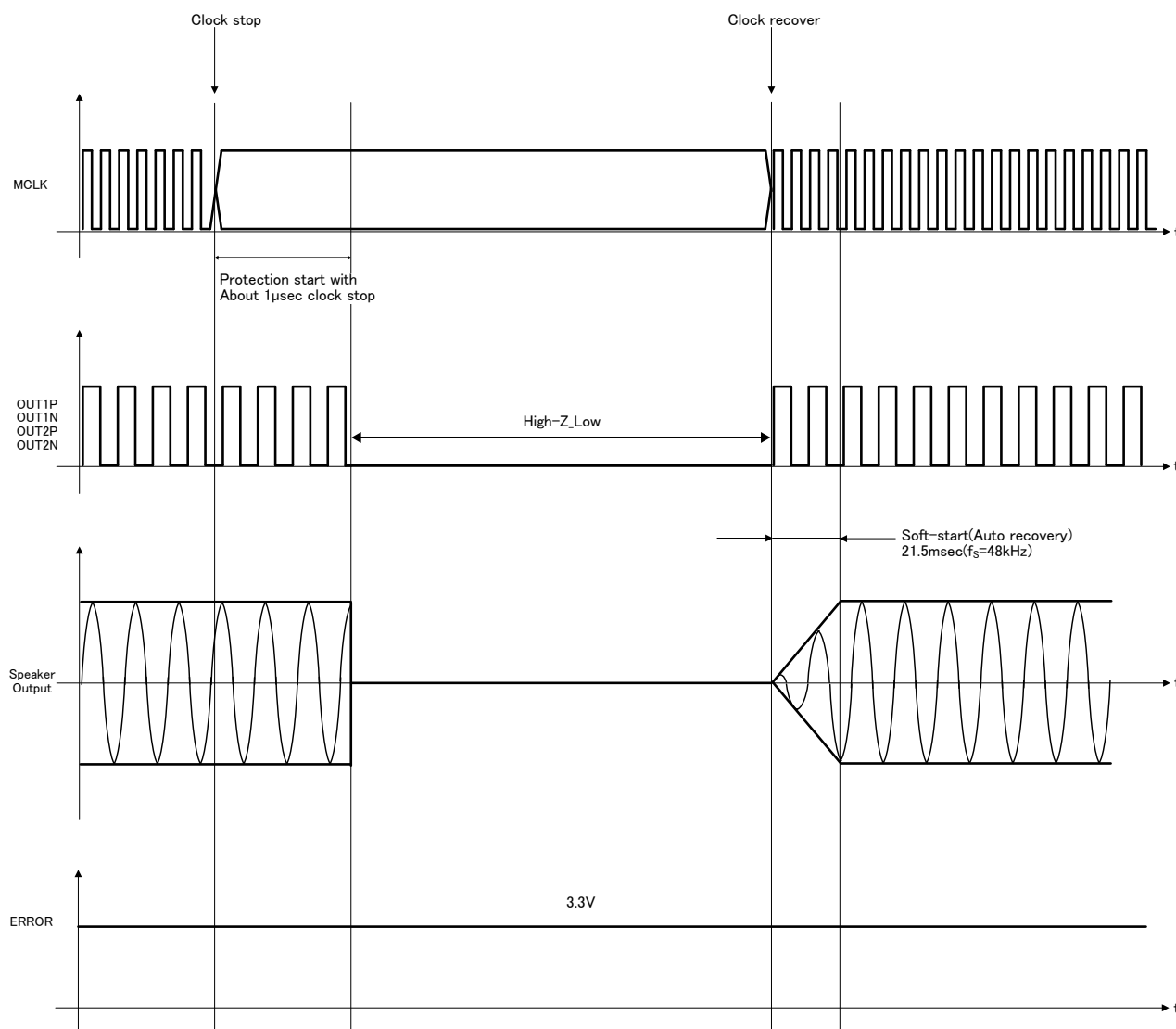


Figure 66. Clock Stop Protection (MCLK) Sequence

(7) Clock Stop Protection (BCLK)

This IC has clock stop protection circuit that mutes the speaker output when the BCLK signal of the digital audio input stops.

Detecting condition - It will detect when MUTE pin is set High and the BCLK signal stops for about 1 μ sec or more. Speaker output turns MUTE immediately when clock stop protection is detected.

Releasing condition - It will release when MUTE pin is set High and the BCLK signal returns to the normal clock operation. The speaker output is outputted through a soft-start when released. (Auto recovery)

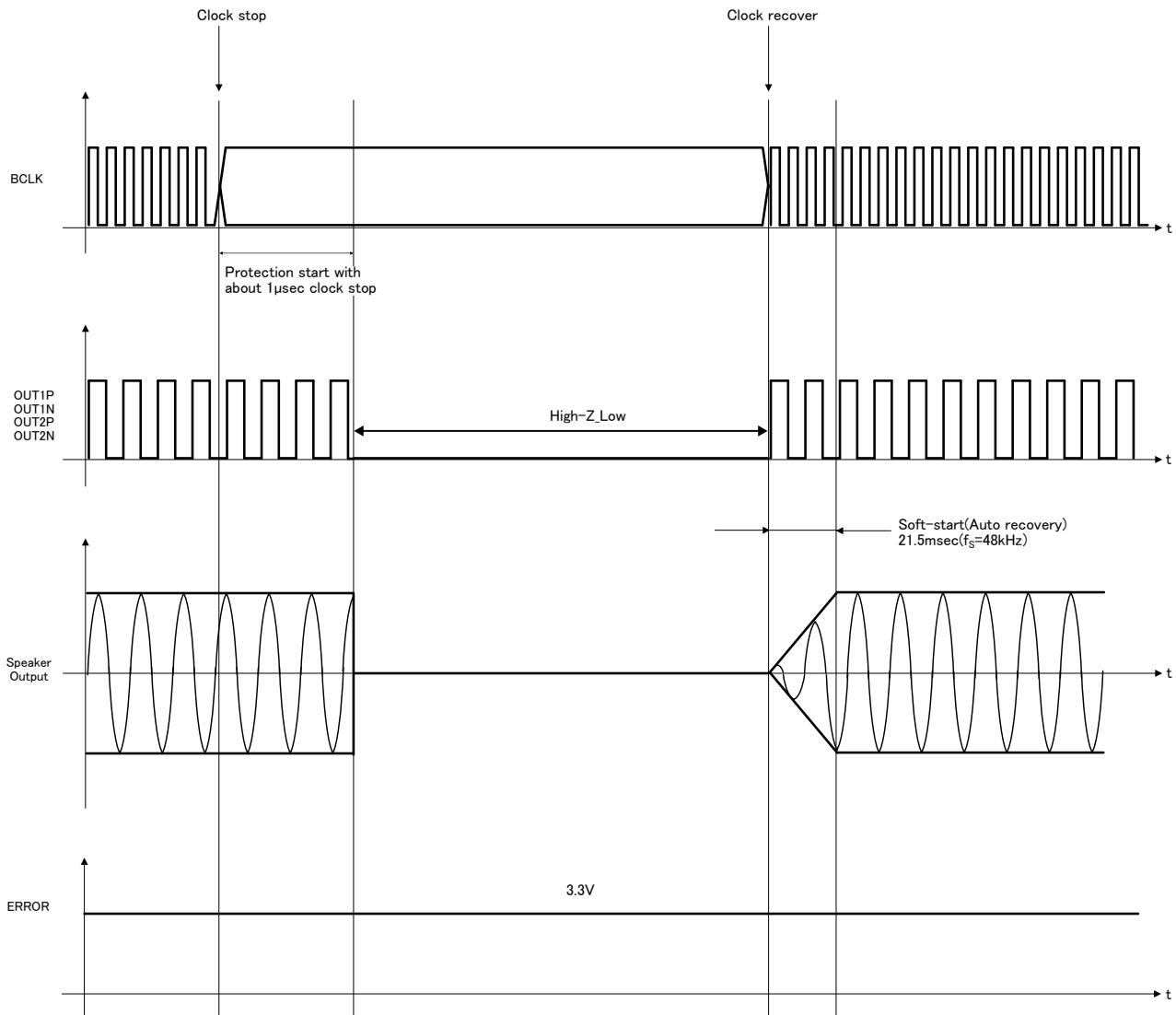


Figure 67. Clock Stop Protection (BCLK) Sequence

(8) Clock Stop Protection (LRCLK)

This IC has clock stop protection circuit that mutes the speaker output when the LRCLK signal of the digital audio input stops.

Detecting condition - It will detect when MUTE pin is set High and the LRCLK signal stops for about 21 μ sec (at $f_s=48$ kHz) or more. Speaker output turns MUTE immediately when clock stop protection is detected.

Releasing condition - It will release when MUTE pin is set High and the LRCLK signal returns to the normal clock operation. The speaker output is outputted through a soft-start when released. (Auto recovery)

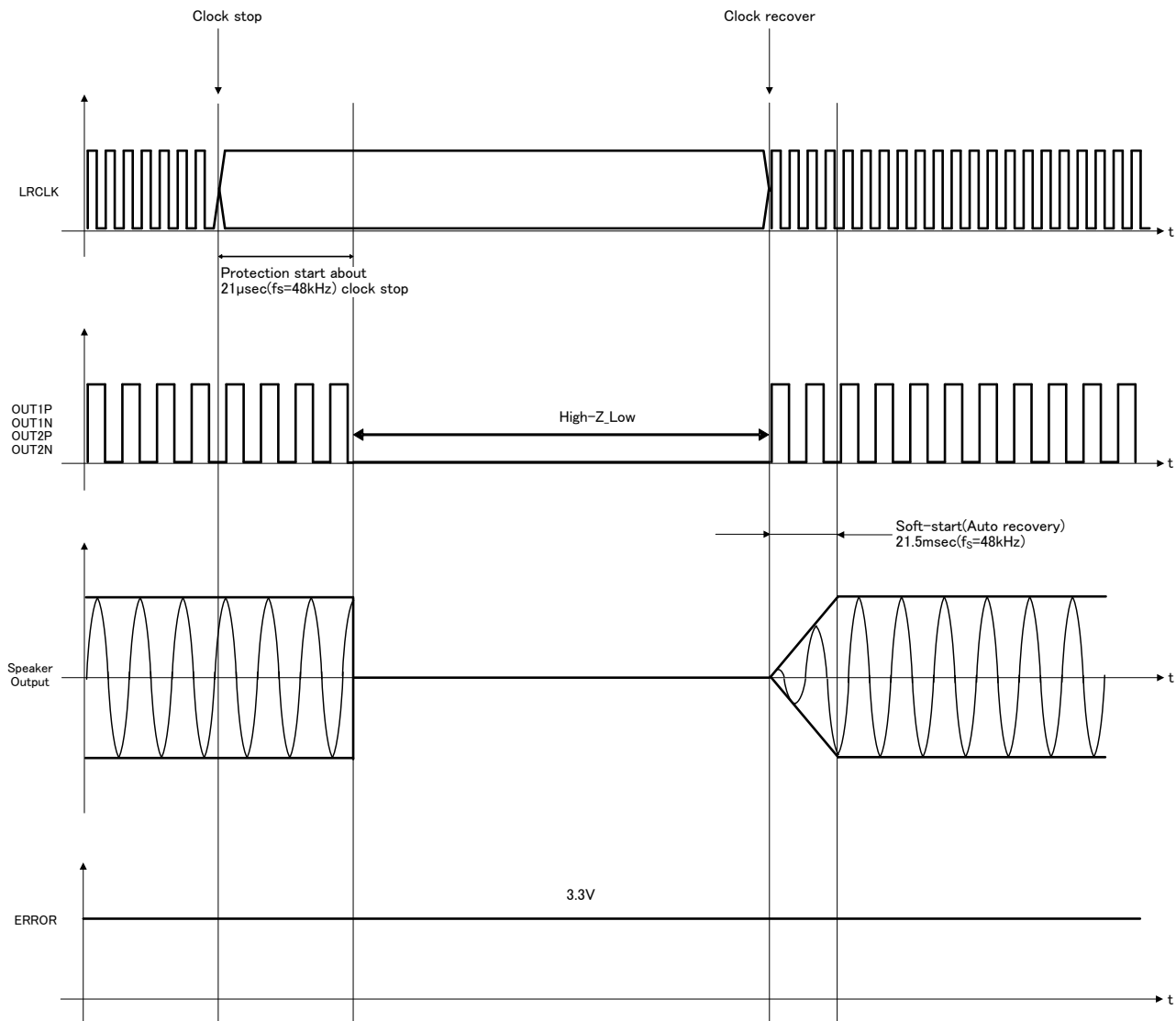


Figure 68. Clock Stop Protection (LRCLK) Sequence

3. Application Circuit Example

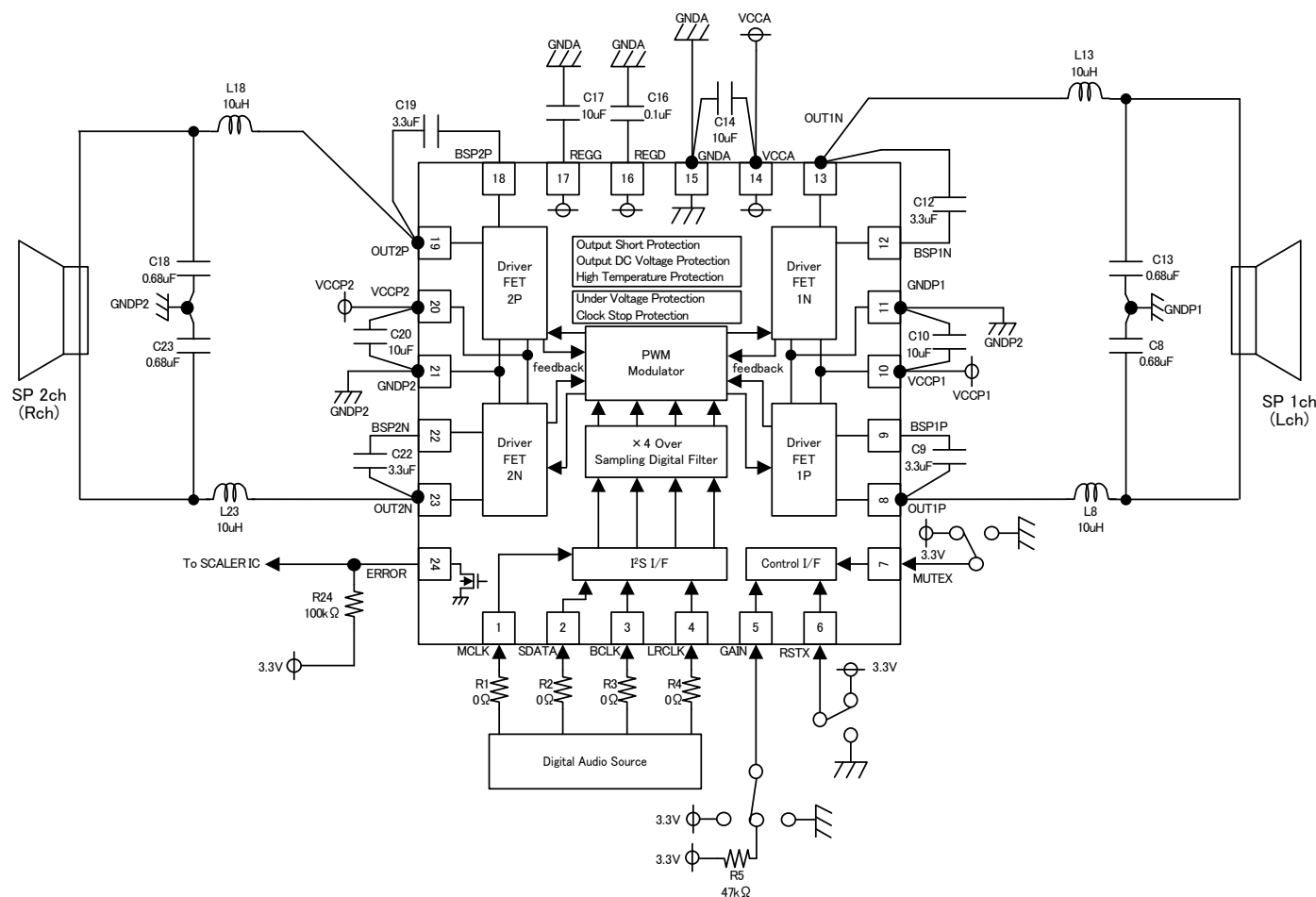
Stereo BTL Output, $R_L=8\Omega/6\Omega$, $V_{CC}\leq 22V$ 

Figure 69. Application Circuit

Parts	Qty	Parts No.	Description	Company	Product No.
Inductor	4	L8, L13, L18, L23	10μH / 3.8A / (±20%) / 7.6mm×7.6mm	TOKO	B1047DS-100M
			10μH / 3.1A / (±20%) / 6.0mm×6.0mm	Taiyo Yuden	NRS6045T-100MMGK
Resistor	1	R5	47kΩ / 1/16W / J(±5%) / 1.0mm×0.5mm	ROHM	MCR01MZPJ473
	4	R1, R2, R3, R4	0Ω / 1/10W / J(±5%) / 1.6mm×0.8mm		MCR03EZPJ000
	1	R24	100kΩ / 1/16W / J(±5%) / 1.0mm×0.5mm		MCR01MZPJ104
Capacitor	4	C8, C13, C18, C23	0.68μF / 50V / B(±10%) / 2.0mm×1.25mm	MURATA	GRM21BB31H684KAC4
	4	C9, C12, C19, C22	3.3μF / 16V / B(±10%) / 1.6mm×0.8mm		GRM21BB31E335KA75
	1	C16	0.1μF / 16V / B(±10%) / 1.6mm×0.8mm		GRM188B11C104KA01
	1	C17	10μF / 16V / B(±10%) / 2.0mm×1.25mm		GRM21BB31C106KE15
	3	C10, C14, C20	10μF / 35V / B(±10%) / 3.2mm×2.5mm	MURATA	GRM32EB3YA106KA12

Caution1: If the impedance characteristics of the speakers at high-frequency range increase rapidly, the IC might not have stable operation in the resonance frequency range of the LC filter. Therefore, consider adding damping-circuit, etc., depending on the impedance of the speaker.

Caution2: Though this IC has a short protection function, when short to VCC or GND after the LC filter, over current occurs during short protection function operation. Be careful about over/undershoot which exceeds the maximum standard ratings because back electromotive force of the inductor will occur which sometimes leads to IC destruction.

The Inductor must be use to the coil with large margin of rated DC current (saturation current). When the short-circuit of the speaker output (After the LC filter) to VCC or GND occurs when the coil with small rated DC current is used, IC destruction might be caused. Because the coil cause the magnetic saturation behavior, it instantaneously pass the heavy-current to IC.

Caution3: Overshoot of output PWM differs according to the board or coupling capacitor of Vcc, and etc. Please check to ensure that it is lower than absolute maximum ratings.

If it exceeds the absolute maximum ratings, snubber circuit must need to be added, the circuit example is shown on the P41 page.

Caution4: When it is used over $V_{CC}=22V$, snubber circuit must need to be added, the circuit example is shown on the P.42 page, and must change LC filter value to suppress the influence of the LRC resonance..

Caution5: This circuit constant is value with ROHM evaluation board, and adjustment of the constant may be necessary for the application board. Please carry out enough evaluations.

Application Circuit Example

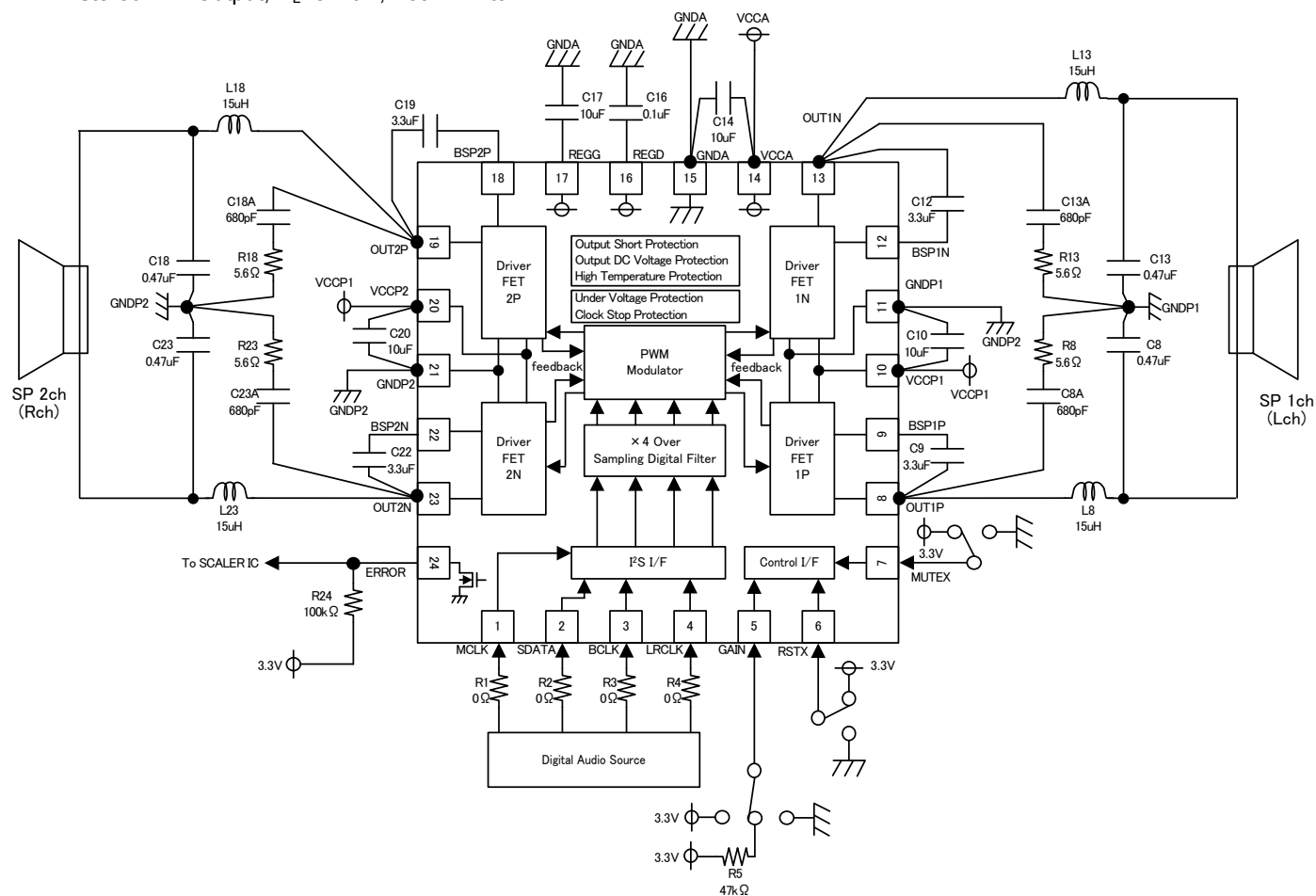
Stereo BTL Output, $R_L=8\Omega/6\Omega$, $V_{CC}=22V$ to $24V$ 

Figure 70. Application Circuit

Parts	Qty	Parts No.	Description	Company	Product No.
Inductor	4	L8, L13, L18, L23	15μH / 2.9A / (±20%) / 7.6mm×7.6mm	TOKO	B1047DS-150M
			15μH / 2.5A / (±20%) / 6.0mm×6.0mm	Taiyo Yuden	NRS6045T-150MMGK
Resistor	1	R5	47kΩ / 1/16W / J(±5%) / 1.0mm×0.5mm	ROHM	MCR01MZPJ473
	4	R1, R2, R3, R4	0Ω / 1/10W / J(±5%) / 1.6mm×0.8mm		MCR03EZPJ000
	4	R8, R13, R18, R23	5.6Ω / 1/4W / J(±5%) / 1.6mm×0.8mm		ESR03EZPJ5R6
	1	R24	100kΩ / 1/16W / J(±5%) / 1.0mm×0.5mm		MCR01MZPJ104
Capacitor	4	C8A, C13A, C18A, C23A	680pF / 50V / CH(±5%) / 1.0mm×0.5mm	MURATA	GRM1552C1H681JA01
	4	C8, C13, C18, C23	0.47μF / 50V / B(±10%) / 2.0mm×1.25mm		GRM21BB31H474KA87
	4	C9, C12, C19, C22	3.3μF / 16V / B(±10%) / 1.6mm×0.8mm		GRM21BB31E335KA75
	1	C16	0.1μF / 16V / B(±10%) / 1.6mm×0.8mm		GRM188B11C104KA01
	1	C17	10μF / 16V / B(±10%) / 2.0mm×1.25mm		GRM21BB31C106KE15
	3	C10, C14, C20	10μF / 35V / B(±10%) / 3.2mm×2.5mm	MURATA	GRM32EB3YA106KA12

Caution1: If the impedance characteristics of the speakers at high-frequency range increase rapidly, the IC might not have stable operation in the resonance frequency range of the LC filter. Therefore, consider adding damping-circuit, etc., depending on the impedance of the speaker.

Caution2: Though this IC has a short protection function, when short to VCC or GND after the LC filter, over current occurs during short protection function operation. Be careful about over/undershoot which exceeds the maximum standard ratings because back electromotive force of the inductor will occur which sometimes leads to IC destruction.

The Inductor must be use to the coil with large margin of rated DC current (saturation current). When the short-circuit of the speaker output (After the LC filter) to VCC or GND occurs when the coil with small rated DC current is used, IC destruction might be caused. Because the coil cause the magnetic saturation behavior, it instantaneously pass the heavy-current to IC.

Caution3: Overshoot of output PWM differs according to the board or coupling capacitor of Vcc, and etc. Please check to ensure that it is lower than absolute maximum ratings.

If it exceeds the absolute maximum ratings, snubber circuit must need to be added, the circuit example is shown on the P41 page.

Caution4: When it is used over $V_{CC}=22V$, snubber circuit must need to be added, the circuit example is shown on the P42 page, and must change LC filter value to suppress the influence of the LRC resonance..

Caution5: This circuit constant is value with ROHM evaluation board, and adjustment of the constant may be necessary for the application board. Please carry out enough evaluations.

Application Circuit Example

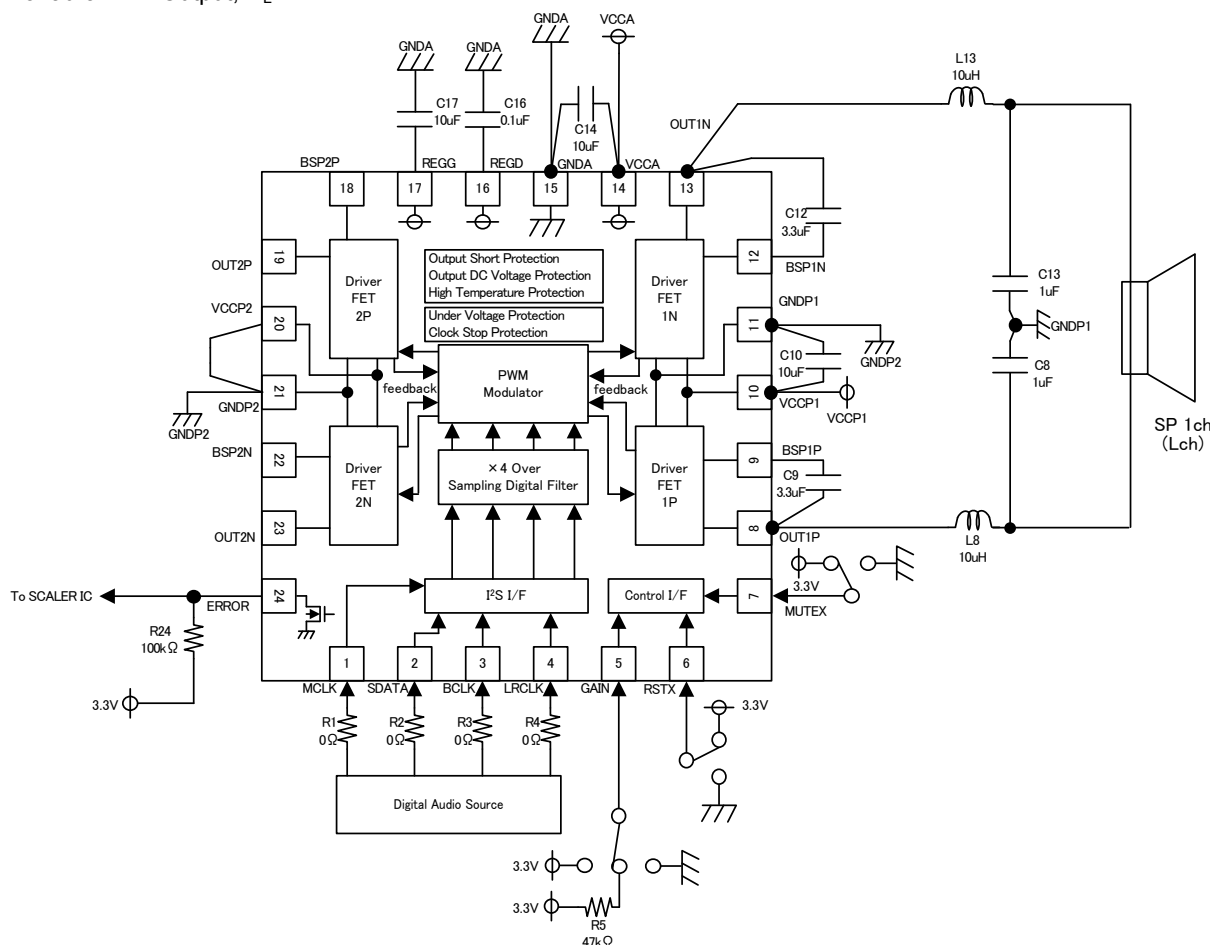
Monaural BTL Output, $R_L=4\Omega$ 

Figure 71. Application Circuit

Parts	Qty	Parts No.	Description	Company	Product No.
Inductor	4	L8, L13, L18, L23	10μH / 3.8A / (±20%) / 7.6mm×7.6mm	TOKO	B1047DS-100M
			10μH / 3.1A / (±20%) / 6.0mm×6.0mm	Taiyo Yuden	NRS6045T-100MMGK
Resistor	1	R5	47kΩ / 1/16W / J(±5%) / 1.0mm×0.5mm	ROHM	MCR01MZPJ473
	4	R1, R2, R3, R4	0Ω / 1/10W / J(±5%) / 1.6mm×0.8mm		MCR03EZPJ000
	1	R24	100kΩ / 1/16W / J(±5%) / 1.0mm×0.5mm		MCR01MZPJ104
Capacitor	4	C8, C13, C18, C23	1μF / 50V / B(±10%) / 2.0mm×1.25mm	MURATA	GRM21BB31H105KA12
	4	C9, C12, C19, C22	3.3μF / 16V / B(±10%) / 1.6mm×0.8mm		GRM21BB31E335KA75
	1	C16	0.1μF / 16V / B(±10%) / 1.6mm×0.8mm		GRM188B11C104KA01
	1	C17	10μF / 16V / B(±10%) / 2.0mm×1.25mm		GRM21BB31C106KE15
	3	C10, C14, C20	10μF / 35V / B(±10%) / 3.2mm×2.5mm	MURATA	GRM32EB3YA106KA12

Caution1: If the impedance characteristics of the speakers at high-frequency range increase rapidly, the IC might not have stable operation in the resonance frequency range of the LC filter. Therefore, consider adding damping-circuit, etc., depending on the impedance of the speaker.

Caution2: Though this IC has a short protection function, when short to VCC or GND after the LC filter, over current occurs during short protection function operation. Be careful about over/undershoot which exceeds the maximum standard ratings because back electromotive force of the inductor will occur which sometimes leads to IC destruction.

The Inductor must be used to the coil with large margin of rated DC current (saturation current). When the short-circuit of the speaker output (After the LC filter) to VCC or GND occurs when the coil with small rated DC current is used, IC destruction might be caused. Because the coil cause the magnetic saturation behavior, it instantaneously pass the heavy-current to IC.

Caution3: Overshoot of output PWM differs according to the board or coupling capacitor of Vcc, and etc. Please check to ensure that it is lower than absolute maximum ratings.

If it exceeds the absolute maximum ratings, snubber circuit must need to be added, the circuit example is shown on the P41 page.

Caution4: When it is used over Vcc=22V, snubber circuit must need to be added, the circuit example is shown on the P42 page, and must change LC filter value to suppress the influence of the LRC resonance..

Caution5: This circuit constant is value with ROHM evaluation board, and adjustment of the constant may be necessary for the application board. Please carry out enough evaluations.

1. About Using BD28623MUV ICs for 2.1ch or 2.2ch audio

Be careful when using two BD28623MUVs at the same time for 2.1ch or 2.2ch audio.

BD28623MUV doesn't have the function that synchronizes both PWM frequencies of the two BD28623MUVs. Beat noise may occur due to the difference between PWM frequencies.

Switching current flows to the GND of LC-Filter and only a small part to the speaker which lowers emission noise. When you have two BD28623MUVs used at the same time with synchronized PWM output, there is common impedance in the GND of the filter. The GND electric potential becomes higher which also causes noise to become higher. The GND of the filter is shorted at one point when you use two BD28623MUVs at the same time. (Figure 73.)

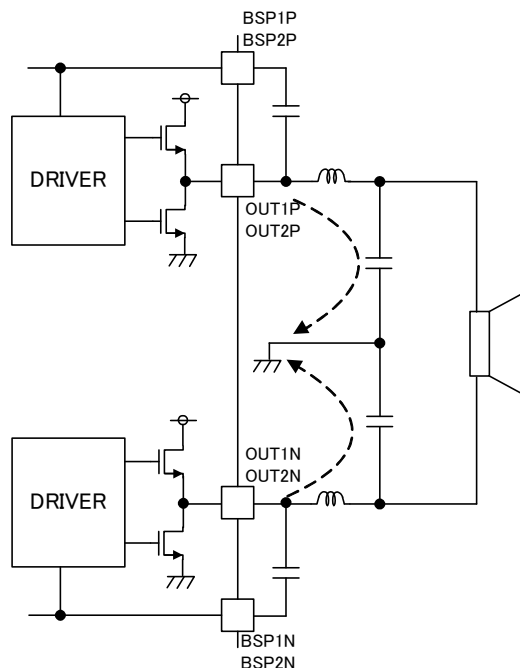


Figure 72. Output LC Filter

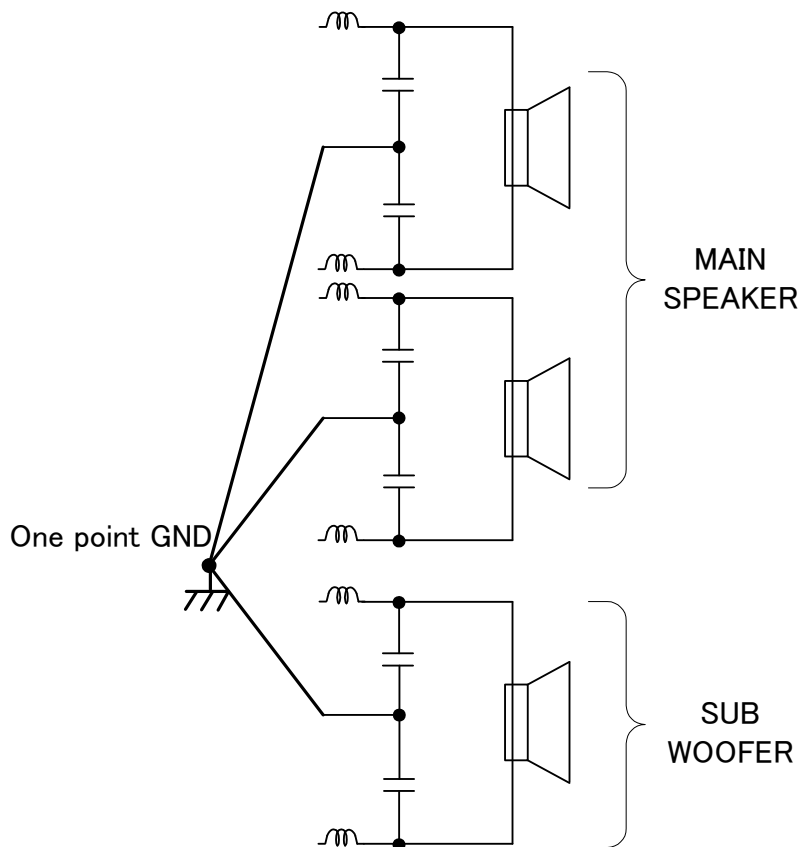


Figure 73. Circuit Using Two ICs to 2.1ch audio

2. Selecting External Components

(1) Output LC Filter Circuit

An output filter is required to eliminate radio-frequency components exceeding the audio-frequency region supplied to a load (speaker). Because this IC uses sampling clock frequencies from 512kHz ($f_s=32\text{kHz}$) to 768kHz ($f_s=48\text{kHz}$) in the output PWM signals, the high-frequency components must be appropriately removed.

This section takes an example of an LC type LPF shown below, in which coil L and capacitor C compose a differential filter with an attenuation property of -12dB/oct. A large part of switching currents of PWM signal flow to capacitor C, and only a small part of the currents flow to speaker R_L . This filter reduces unwanted emission this way. In addition, coil L and capacitor C compose a filter against in-phase components, reducing unwanted emission further.

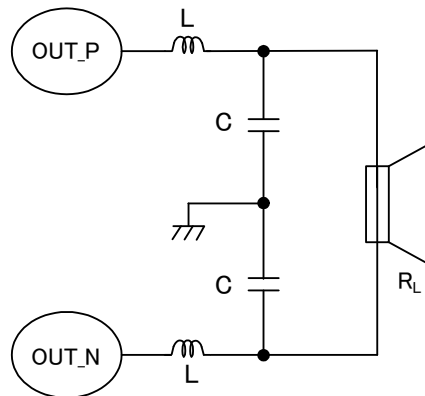


Figure 74. Output LC Filter

The following shows output LC filter constants with typical load impedances.

R_L	L	C
4Ω	10μH	1μF
6Ω, 8Ω ($V_{CC} \leq 22\text{V}$)	10μH	0.68μF
6Ω, 8Ω ($V_{CC} > 22\text{V}$)	15μH	0.47μF

The inductors must be use with low ESR and with sufficient margin of rated DC current (saturation current).

Power loss will increase if inductors with high ESR are used.

When the short-circuit of the speaker output (After the LC filter) to VCC or GND occurs when the coil with small rated DC current is used, IC destruction might be caused. Because the coil cause the magnetic saturation behavior, it instantaneously pass the heavy-current to IC. (The coil of the rated DC current: 7.2A or more will be recommended when using it by 22V or more.)

Select a closed magnetic circuit type product in normal cases to prevent emission noise.

Use capacitors with low equivalent series resistance, and good impedance characteristics at high frequency ranges (100kHz or higher). Also, select an item with sufficient voltage rating because massive amount of high-frequency current flow is expected.

(2) Snubber circuit constant

When overshoot of PWM Output exceeds absolute maximum rating, or when overshoot of PWM output negatively affects EMC, or when ringing deteriorates the audio characteristic of the PWM output, snubber circuit is used as shown below.

(a) Measure the spike resonance frequency " f_1 " of PWM output waveform (when rising) by using Low capacitance Probe (e.g. FET probe) at the OUT terminal. (Figure 75)
Shorten GND lead of FET probe and monitor as near as possible to output pin.

(b) Measure the resonance frequency " f_2 " of the spike as the snubber-circuit R value equals 0Ω
(capacitor "C" is connected to GND)

Adjust the value of the capacitor "C" until it becomes ($2 \times f_2 = f_1$)

The value of "C" that becomes ($2 \times f_2 = f_1$) is 3 times of the parasitic capacity " C_p " that a spike is formed. ($C=3C_p$)

(c) Parasitic inductance " L_p " is calculated using the next formula.

$$L_p = \frac{1}{(2\pi f_1)^2 C_p}$$

(d) The character impedance Z of resonance is calculated from the parasitic capacity " C_p " and the parasitic inductance L_p using the next formula.

$$Z = \sqrt{\frac{L_p}{C_p}}$$

(e) Set snubber circuit "R" same as the character impedance "Z".

Set snubber circuit "C" 4 to 10 times of the parasitic capacity " C_p ".

If "C" is set larger than $10C_p$, switching current will possibly increase.

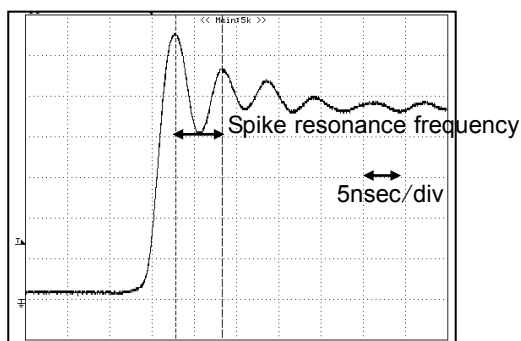


Figure 75. PWM Output Waveform
(Measure of Spike Resonance Frequency)

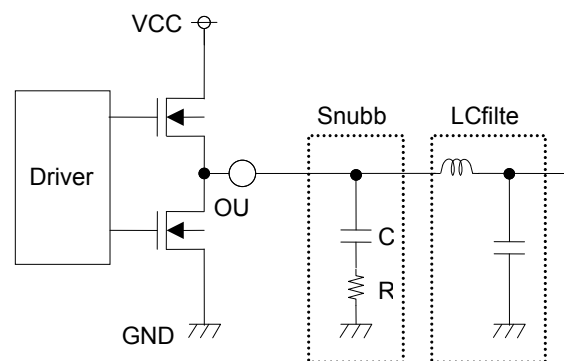


Figure76. Snubber Schematic

The following table shows ROHM recommended value of "Snubber filter constants" when using ROHM 4 layer board.
($V_{cc}=22V$ to $24V$, $R_L=8\Omega$, $P_o=10W+10W$)

C	R
470pF to 820pF, 50V CH(±5%) Murata GRM1552C1H Series	5.6Ω, 1/4W J(±5%) ROHM ESR03EZPJ5R6

3) Operating condition with the application component

Parameter	Parts No.	Limit			Unit	Conditions
		Min	Typ	Max		
Tolerance of Coupling capacitor for Power supply	C10, C14, C20	1 ^(Note 11)	10	-	μF	B characteristics Ceramic type capacitor recommended
Tolerance of Capacitor for REGG	C17	1 ^(Note 11)	10	-	μF	B characteristics, 16V Ceramic type capacitor recommended
Tolerance of Capacitor for REGD	C16	0.05 ^(Note 11)	0.1	-	μF	B characteristics, 16V Ceramic type capacitor recommended
Tolerance of Capacitor for BSP	C9, C12, C19, C22	2.0 ^(Note 11)	3.3	4.5 ^(Note 12)	μF	B characteristics, 16V Ceramic type capacitor recommended
		2.0 ^(Note 11)	4.7	6.3 ^(Note 12)	μF	
Tolerance of GAIN Terminal Pull up resistor	R5	43	47	51	kΩ	1/16W J(±5%) recommended

(Note 11) Should use the capacity of the capacitor not to be less than a minimum in consideration of temperature characteristics and dc-bias characteristics.

(Note 12) It is value in consideration of +/-10% of capacity unevenness, capacity rate of change 22%. Please use the capacitor within this limit.

Power Dissipation

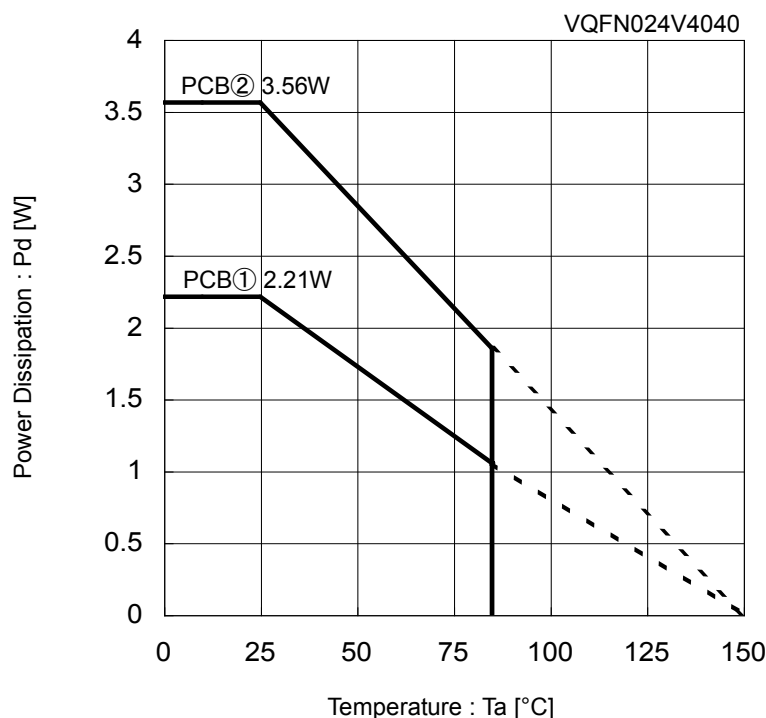


Figure 77. Power Dissipation Curve

Measuring instrument : TH-156 (Shibukawa Kuwano Electric Instruments Co., Ltd.)

Measuring conditions : Installation on ROHM's board

Board size : 74.2mm x 74.2mm x 1.6mm (with thermal via on board)

Material : FR4

- The board and exposed heat sink on the back of package are connected by soldering.

PCB① : 4-layer board (Top and bottom layer back copper foil size: 10.29mm², 2nd and 3rd layer back copper foil size: 5505mm²), $\theta_{ja} = 56.6^{\circ}\text{C/W}$

PCB② : 4-layer board (back copper foil size: 5505mm²), $\theta_{ja} = 35.1^{\circ}\text{C/W}$

Use a thermal design that has sufficient margin so as not to exceed allowable power dissipation (Pd) in actual operating conditions. This IC exposes its frame of the backside of package. Note that this part is used to provide heat dissipation treatment to improve heat dissipation efficiency. Try to occupy as wide as possible heat dissipation pattern not only on the board surface but also the backside.

Class D speaker amplifier has high efficiency and low heat generation in comparison with conventional analog power amplifier. However, in case it is operated continuously by maximum output power, power dissipation (Pdiss) may exceed package dissipation. Please consider heat design that power dissipation (Pdiss) does not exceed package dissipation (Pd) in average power (Poav).

$$\begin{aligned} \text{Package dissipation : } Pd(W) &= (T_{j\max} - T_a) / \theta_{ja} \\ \text{Power dissipation : } Pdiss(W) &= Poav \times (1/\eta - 1) \end{aligned}$$

where:

$T_{j\max}$ is the maximum junction temperature=150°C

T_a is the peripheral temperature[°C],

θ_{ja} is the thermal resistance of package[°C/W],

Poav is the average power [W],

η is the efficiency

Operational Notes

1. **Reverse Connection of Power Supply**
Connecting the power supply in reverse polarity can damage the IC. Take precautions against reverse polarity when connecting the power supply, such as mounting an external diode between the power supply and the IC's power supply pins.
2. **Power Supply Lines**
Design the PCB layout pattern to provide low impedance supply lines. Separate the ground and supply lines of the digital and analog blocks to prevent noise in the ground and supply lines of the digital block from affecting the analog block. Furthermore, connect a capacitor to ground at all power supply pins. Consider the effect of temperature and aging on the capacitance value when using electrolytic capacitors.
3. **Ground Voltage**
Ensure that no pins are at a voltage below that of the ground pin at any time, even during transient condition.
4. **Ground Wiring Pattern**
When using both small-signal and large-current ground traces, the two ground traces should be routed separately but connected to a single ground at the reference point of the application board to avoid fluctuations in the small-signal ground caused by large currents. Also ensure that the ground traces of external components do not cause variations on the ground voltage. The ground lines must be as short and thick as possible to reduce line impedance.
5. **Thermal Consideration**
Should by any chance the power dissipation rating be exceeded the rise in temperature of the chip may result in deterioration of the properties of the chip. The absolute maximum rating of the Pd stated in this specification is when the IC is mounted on a 70mm x 70mm x 1.6mm glass epoxy board. In case of exceeding this absolute maximum rating, increase the board size and copper area to prevent exceeding the Pd rating.
6. **Recommended Operating Conditions**
These conditions represent a range within which the expected characteristics of the IC can be approximately obtained. The electrical characteristics are guaranteed under the conditions of each parameter.
7. **Inrush Current**
When power is first supplied to the IC, it is possible that the internal logic may be unstable and inrush current may flow instantaneously due to the internal powering sequence and delays, especially if the IC has more than one power supply. Therefore, give special consideration to power coupling capacitance, power wiring, width of ground wiring, and routing of connections.
8. **Operation Under Strong Electromagnetic Field**
Operating the IC in the presence of a strong electromagnetic field may cause the IC to malfunction.
9. **Testing on Application Boards**
When testing the IC on an application board, connecting a capacitor directly to a low-impedance output pin may subject the IC to stress. Always discharge capacitors completely after each process or step. The IC's power supply should always be turned OFF completely before connecting or removing it from the test setup during the inspection process. To prevent damage from static discharge, ground the IC during assembly and use similar precautions during transport and storage.
10. **Inter-pin Short and Mounting Errors**
Ensure that the direction and position are correct when mounting the IC on the PCB. Incorrect mounting may result in damaging the IC. Avoid nearby pins being shorted to each other especially to ground, power supply and output pin. Inter-pin shorts could be due to many reasons such as metal particles, water droplets (in very humid environment) and unintentional solder bridge deposited in between pins during assembly to name a few.
11. **Unused Input Pins**
Input pins of an IC are often connected to the gate of a MOS transistor. The gate has extremely high impedance and extremely low capacitance. If left unconnected, the electric field from the outside can easily charge it. The small charge acquired in this way is enough to produce a significant effect on the conduction through the transistor and cause unexpected operation of the IC. So unless otherwise specified, unused input pins should be connected to the power supply or ground line.

Operational Notes – continued

12. Regarding the Input Pin of the IC

This monolithic IC contains P+ isolation and P substrate layers between adjacent elements in order to keep them isolated. P-N junctions are formed at the intersection of the P layers with the N layers of other elements, creating a parasitic diode or transistor. For example (refer to figure below):

When $GND > Pin\ A$ and $GND > Pin\ B$, the P-N junction operates as a parasitic diode.

When $GND > Pin\ B$, the P-N junction operates as a parasitic transistor.

Parasitic diodes inevitably occur in the structure of the IC. The operation of parasitic diodes can result in mutual interference among circuits, operational faults, or physical damage. Therefore, conditions that cause these diodes to operate, such as applying a voltage lower than the GND voltage to an input pin (and thus to the P substrate) should be avoided.

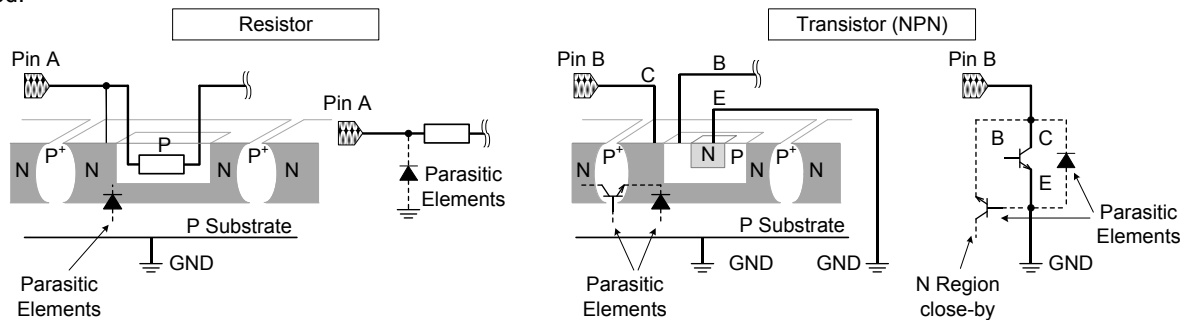


Figure 78. Example of Monolithic IC Structure

13. Ceramic Capacitor

When using a ceramic capacitor, determine the dielectric constant considering the change of capacitance with temperature and the decrease in nominal capacitance due to DC bias and others.

14. Area of Safe Operation (ASO)

Operate the IC such that the output voltage, output current, and power dissipation are all within the Area of Safe Operation (ASO).

15. Thermal Shutdown Circuit(TSD)

This IC has a built-in thermal shutdown circuit that prevents heat damage to the IC. Normal operation should always be within the IC's power dissipation rating. If however the rating is exceeded for a continued period, the junction temperature (T_j) will rise which will activate the TSD circuit that will turn OFF all output pins. When the T_j falls below the TSD threshold, the circuits are automatically restored to normal operation.

Note that the TSD circuit operates in a situation that exceeds the absolute maximum ratings and therefore, under no circumstances, should the TSD circuit be used in a set design or for any purpose other than protecting the IC from heat damage.

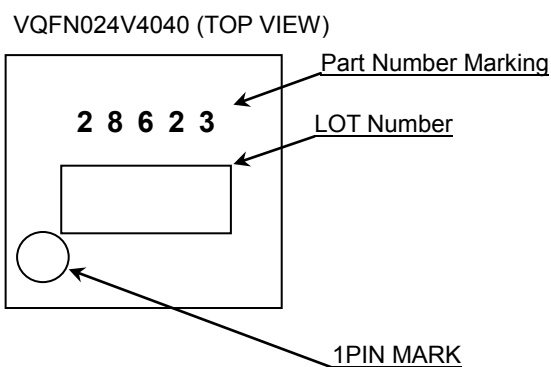
16. Over-Current Protection Circuit (OCP)

This IC incorporates an integrated over current protection circuit that is activated when the load is shorted. This protection circuit is effective in preventing damage due to sudden and unexpected incidents. However, the IC should not be used in applications characterized by continuous operation or transitioning of the protection circuit.

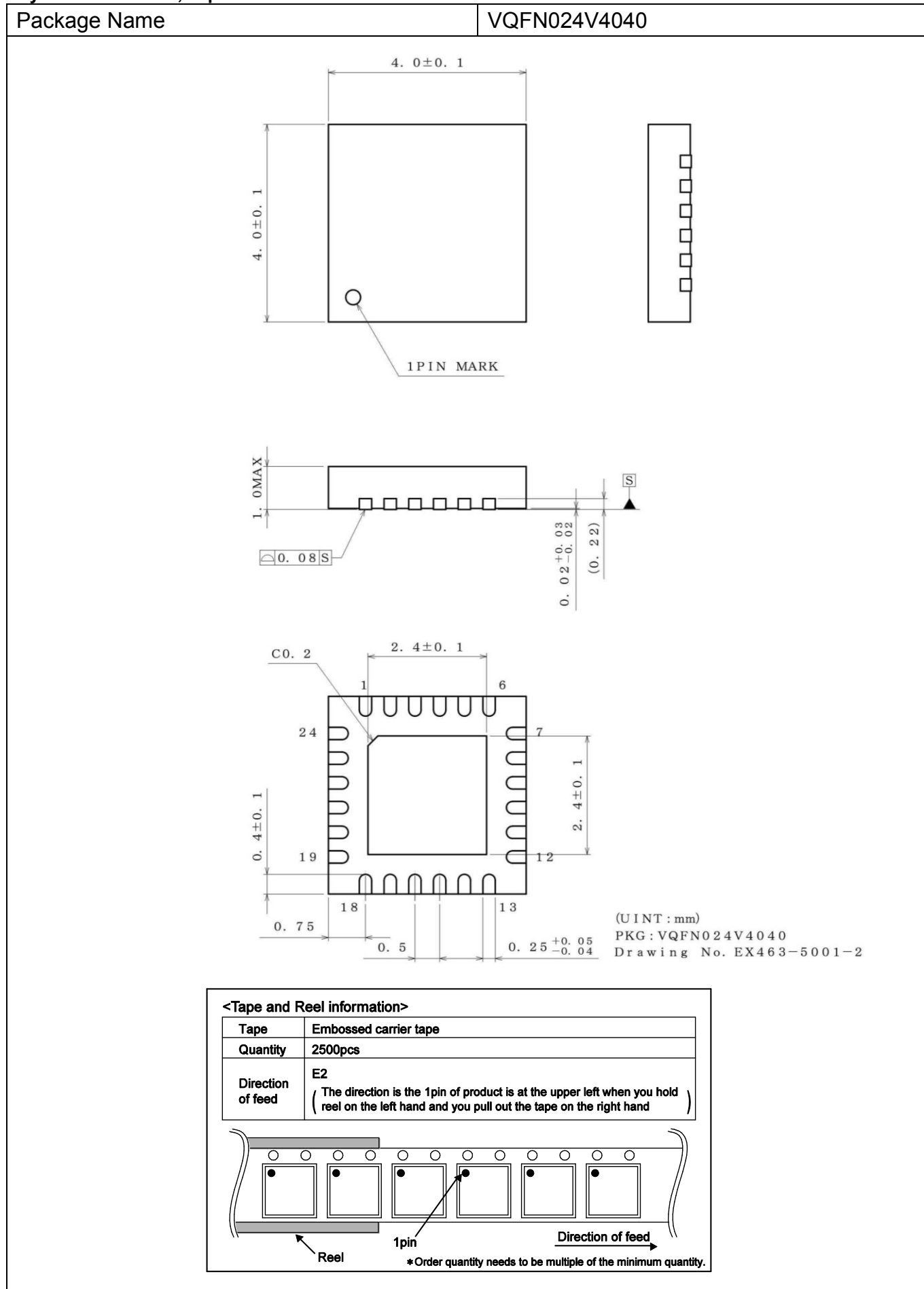
Ordering Information

B D 2 8 6 2 3 M U V										-	E 2	
Part Number										Package MUV: VQFN024V4040		Packaging and forming specification E2: Embossed tape and reel

Marking Diagram



Physical Dimension, Tape and Reel Information



Revision History

Date	Revision	Changes
20.Aug.2015	001	First revision
20.May.2016	002	P.3 Pin Descriptions, I/O Equivalent Circuits Pin No. 5 P.4 Pin Descriptions, I/O Equivalent Circuits Pin No. 17 P.7 Electrical Characteristics High level Input Voltage 1, Low level Input Voltage 1 Conditions P.19 Timing Chart/ Power Supply Start-up Sequence P.22 Timing Chart/ Action at MCLK Unstable 3 P.23 Timing Chart/ Instantaneous Power Interruption Recovery Sequence P.38 Application Circuit Example/Product No. P.39 Application Circuit Example /Description. P.40 Output LC Filter P.43 Operating condition with the application component

Notice

Precaution on using ROHM Products

- Our Products are designed and manufactured for application in ordinary electronic equipments (such as AV equipment, OA equipment, telecommunication equipment, home electronic appliances, amusement equipment, etc.). If you intend to use our Products in devices requiring extremely high reliability (such as medical equipment ^(Note 1), transport equipment, traffic equipment, aircraft/spacecraft, nuclear power controllers, fuel controllers, car equipment including car accessories, safety devices, etc.) and whose malfunction or failure may cause loss of human life, bodily injury or serious damage to property ("Specific Applications"), please consult with the ROHM sales representative in advance. Unless otherwise agreed in writing by ROHM in advance, ROHM shall not be in any way responsible or liable for any damages, expenses or losses incurred by you or third parties arising from the use of any ROHM's Products for Specific Applications.

(Note1) Medical Equipment Classification of the Specific Applications

JAPAN	USA	EU	CHINA
CLASS III	CLASS III	CLASS II b	CLASS III
CLASS IV		CLASS III	

- ROHM designs and manufactures its Products subject to strict quality control system. However, semiconductor products can fail or malfunction at a certain rate. Please be sure to implement, at your own responsibilities, adequate safety measures including but not limited to fail-safe design against the physical injury, damage to any property, which a failure or malfunction of our Products may cause. The following are examples of safety measures:
 - Installation of protection circuits or other protective devices to improve system safety
 - Installation of redundant circuits to reduce the impact of single or multiple circuit failure
- Our Products are designed and manufactured for use under standard conditions and not under any special or extraordinary environments or conditions, as exemplified below. Accordingly, ROHM shall not be in any way responsible or liable for any damages, expenses or losses arising from the use of any ROHM's Products under any special or extraordinary environments or conditions. If you intend to use our Products under any special or extraordinary environments or conditions (as exemplified below), your independent verification and confirmation of product performance, reliability, etc. prior to use, must be necessary:
 - Use of our Products in any types of liquid, including water, oils, chemicals, and organic solvents
 - Use of our Products outdoors or in places where the Products are exposed to direct sunlight or dust
 - Use of our Products in places where the Products are exposed to sea wind or corrosive gases, including Cl₂, H₂S, NH₃, SO₂, and NO₂
 - Use of our Products in places where the Products are exposed to static electricity or electromagnetic waves
 - Use of our Products in proximity to heat-producing components, plastic cords, or other flammable items
 - Sealing or coating our Products with resin or other coating materials
 - Use of our Products without cleaning residue of flux (even if you use no-clean type fluxes, cleaning residue of flux is recommended); or Washing our Products by using water or water-soluble cleaning agents for cleaning residue after soldering
 - Use of the Products in places subject to dew condensation
- The Products are not subject to radiation-proof design.
- Please verify and confirm characteristics of the final or mounted products in using the Products.
- In particular, if a transient load (a large amount of load applied in a short period of time, such as pulse. is applied, confirmation of performance characteristics after on-board mounting is strongly recommended. Avoid applying power exceeding normal rated power; exceeding the power rating under steady-state loading condition may negatively affect product performance and reliability.
- De-rate Power Dissipation depending on ambient temperature. When used in sealed area, confirm that it is the use in the range that does not exceed the maximum junction temperature.
- Confirm that operation temperature is within the specified range described in the product specification.
- ROHM shall not be in any way responsible or liable for failure induced under deviant condition from what is defined in this document.

Precaution for Mounting / Circuit board design

- When a highly active halogenous (chlorine, bromine, etc.) flux is used, the residue of flux may negatively affect product performance and reliability.
- In principle, the reflow soldering method must be used on a surface-mount products, the flow soldering method must be used on a through hole mount products. If the flow soldering method is preferred on a surface-mount products, please consult with the ROHM representative in advance.

For details, please refer to ROHM Mounting specification

Precautions Regarding Application Examples and External Circuits

1. If change is made to the constant of an external circuit, please allow a sufficient margin considering variations of the characteristics of the Products and external components, including transient characteristics, as well as static characteristics.
2. You agree that application notes, reference designs, and associated data and information contained in this document are presented only as guidance for Products use. Therefore, in case you use such information, you are solely responsible for it and you must exercise your own independent verification and judgment in the use of such information contained in this document. ROHM shall not be in any way responsible or liable for any damages, expenses or losses incurred by you or third parties arising from the use of such information.

Precaution for Electrostatic

This Product is electrostatic sensitive product, which may be damaged due to electrostatic discharge. Please take proper caution in your manufacturing process and storage so that voltage exceeding the Products maximum rating will not be applied to Products. Please take special care under dry condition (e.g. Grounding of human body / equipment / solder iron, isolation from charged objects, setting of ionizer, friction prevention and temperature / humidity control).

Precaution for Storage / Transportation

1. Product performance and soldered connections may deteriorate if the Products are stored in the places where:
 - [a] the Products are exposed to sea winds or corrosive gases, including Cl₂, H₂S, NH₃, SO₂, and NO₂
 - [b] the temperature or humidity exceeds those recommended by ROHM
 - [c] the Products are exposed to direct sunshine or condensation
 - [d] the Products are exposed to high Electrostatic
2. Even under ROHM recommended storage condition, solderability of products out of recommended storage time period may be degraded. It is strongly recommended to confirm solderability before using Products of which storage time is exceeding the recommended storage time period.
3. Store / transport cartons in the correct direction, which is indicated on a carton with a symbol. Otherwise bent leads may occur due to excessive stress applied when dropping of a carton.
4. Use Products within the specified time after opening a humidity barrier bag. Baking is required before using Products of which storage time is exceeding the recommended storage time period.

Precaution for Product Label

A two-dimensional barcode printed on ROHM Products label is for ROHM's internal use only.

Precaution for Disposition

When disposing Products please dispose them properly using an authorized industry waste company.

Precaution for Foreign Exchange and Foreign Trade act

Since concerned goods might be fallen under listed items of export control prescribed by Foreign exchange and Foreign trade act, please consult with ROHM in case of export.

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Электрон
Связь**

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Благодаря сотрудничеству с мировыми поставщиками мы осуществляем комплексные и плановые поставки широчайшего спектра электронных компонентов.

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Наши контакты:

Телефон: +7 812 627 14 35

Электронная почта: sales@st-electron.ru

Адрес: 198099, Санкт-Петербург,
Промышленная ул, дом № 19, литера Н,
помещение 100-Н Офис 331