

GENERAL DESCRIPTION

This document describes the specification for the IDTF1953 Digital Step Attenuator. The F1953 is part of a family of *Glitch-Free™* DSAs optimized for the demanding requirements of communications Infrastructure. These devices are offered in a compact 4x4 QFN package with 50 Ω impedances for ease of integration.

COMPETITIVE ADVANTAGE

Digital step attenuators are used in Receivers and Transmitters to provide gain control. The IDTF1953 is a 6-bit step attenuator optimized for these demanding applications. The silicon design has very low insertion loss and low distortion (> +60 dBm IP_{3I}.) The device has pinpoint accuracy and settles to final attenuation value within 400 nsec. Most importantly, the F1953 includes IDT's *Glitch-Free™* technology which results in less than 0.5 dB of overshoot ringing during MSB transitions. This is in stark contrast to competing DSAs that *glitch as much as 10 dB* (see p. 10.)

- ✓ Lowest insertion loss for best SNR
- ✓ *Glitch-Free™* when transitioning – won't damage PA or ADC
- ✓ Extremely accurate with low distortion



APPLICATIONS

- Base Station 2G, 3G, 4G, TDD radiocards
- Repeaters and E911 systems
- Digital Pre-Distortion
- Point to Point Infrastructure
- Public Safety Infrastructure
- WIMAX Receivers and Transmitters
- Military Systems, JTRS radios
- RFID handheld and portable readers
- Cable Infrastructure

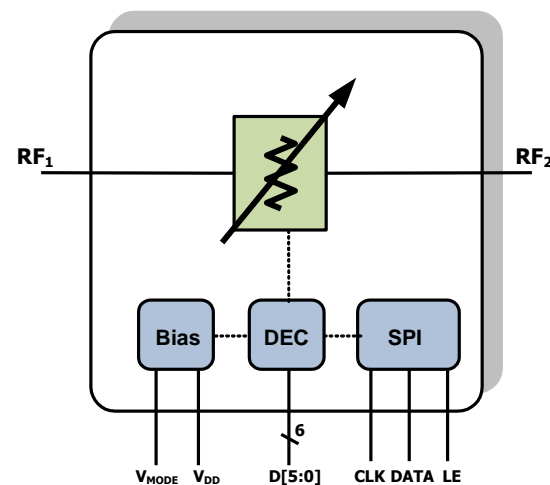
PART# MATRIX

Part#	Freq range	Resolution / Range	Control	IL	Pinout
F1950	150 - 4000	0.25 / 31.75	Parallel & Serial	-1.3	PE43702 PE43701
F1951	100 - 4000	0.50 / 31.5	Serial Only	-1.2	HMC305
F1952	100 - 4000	0.50 / 15.5	Serial Only	-0.9	HMC305
F1953	400 - 4000	0.50 / 31.5	Parallel & Serial	-1.3	PE4302 DAT-31R5

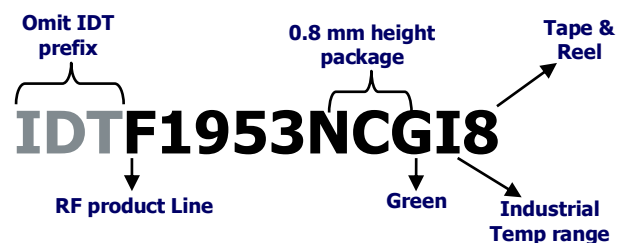
FEATURES

- *Glitch-Free™*, < 0.6 dB transient overshoot
- Spurious Free Design
- 2.7 to 3.3 V supply
- Attenuation Error < 0.5 dB @ 2 GHz
- Low Insertion Loss < 1.4 dB @ 2 GHz
- Excellent Linearity >+60 dBm IP_{3I}
- Fast settling time, < 400 nsec
- Serial or Parallel Interface 31.5 dB Range
- Stable Integral Non-Linearity over temperature
- Low Power Consumption < 200 uA
- Integrated DC blocking capacitors
- Drop-In replacement
- 4x4 mm Thin QFN 20 pin package

DEVICE BLOCK DIAGRAM



ORDERING INFORMATION



ABSOLUTE MAXIMUM RATINGS

V_{DD} to GND	-0.3V to +3.3V
D[5:0], DATA, CLK, LE, V_{MODE}	-0.3V to 3.6V
RF Input Power (RF1, RF2) calibration and testing	+29 dBm
RF Input Power (RF1, RF2) continuous RF operation	+23 dBm
θ_{JA} (Junction – Ambient)	+50°C/W
θ_{JC} (Junction – Case) The Case is defined as the exposed paddle	+3°C/W
Operating Temperature Range (Case Temperature)	$T_C = -40^{\circ}\text{C}$ to $+100^{\circ}\text{C}$
Maximum Junction Temperature	140°C
Storage Temperature Range	-65°C to +150°C
Lead Temperature (soldering, 10s) .	+260°C

Stresses above those listed above may cause permanent damage to the device. Functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ESD Caution

This product features proprietary protection circuitry. However, it may be damaged if subjected to high energy ESD. Please use proper ESD precautions when handling to avoid damage or loss of performance.



6-bit Digital Step Attenuator
400 to 4000 MHz IDTF1953
IDTF1953 SPECIFICATION (31.5 dB Range)

 Specifications apply at $V_{DD} = +3.0V$, $f_{RF} = 2000MHz$, $T_C = +25^{\circ}C$, $V_{MODE} > V_{IH}$ (Serial Mode) EVkit losses are de-embedded (see p. 17)

Parameter	Comment	Sym.	min	typ	max	units
Logic Input High	CLK, CSb, SDI, SDO, RSTb	V_{IH}	$0.7xV_{DD}$		V_{DD}	V
Logic Input Low	CLK, CSb, SDI, SDO, RSTb	V_{IL}			$0.3xV_{DD}$	V
Logic Current	V_{MODE} , D[5:0]	I_{IH}, I_{IL}	-5		+5	μA
Logic Current	LE	I_{IH}, I_{IL}	-35		+35	μA
Supply Voltage(s)	Main Supply	V_{DD}		2.7 to 3.3		V
Supply Current	Total $V_{DD} = 3V$	I_{DD}		0.16	0.25¹	mA
Temperature Range	Operating Range (Case)	T_C		-40 to +100		degC
Frequency Range	Operating Range	f_{RF}		400 to 4000		MHz
RF1,RF2 Return Loss	$20*\log(S_{11})$, $20*\log(S_{22})$	S_{11}, S_{22}		-23		dB
Minimum Attenuation	D[5:0] = [000000]	A_{MIN}		1.35	1.90	dB
Maximum Attenuation	D[5:0] = [111111]	A_{MAX}	32.0	32.4		dB
Minimum Gain Step	Least Significant Bit	LSB		0.50		dB
Phase Delta	Phase change A_{MIN} vs. A_{MAX}	Φ_{Δ}		39		deg
Differential ATTN Error	Between adjacent steps	DNL		0.09		dB
Integral ATTN Error	Error vs. line (A_{MIN} ref) to 13.5dB ATTN	INL_1		0.20	0.60	dB
Integral ATTN Error	Error vs. line (A_{MIN} ref) to 31.5dB ATTN	INL_2		0.47	0.75	dB
Input IP3	D[5:0] = [000000] = A_{MIN} D[5:0] = [011111] = $A_{15.5}$ D[5:0] = [111111] = A_{MAX} ▪ $P_{IN} = +10$ dBm per tone ▪ 50 MHz Tone Separation	$IP3I_1$ $IP3I_2$ $IP3I_3$	$+57^2$ $+53$ $+53$	$+66$ $+60$ $+60$		dBm
0.1 dB Compression <i>Please note ABS MAX P_{IN} on Page 2</i>	▪ D[5:0] = [000101] = $A_{2.5}$ ▪ Baseline $P_{IN} = 20$ dBm	$P_{0.1}$		28.5		dBm
Settling Time (parallel mode)	▪ Start LE rising edge $> V_{IH}$ ▪ End +/-0.10 dB Pout settling ▪ 15.5 – 16.0 transition	T_{LSB}		400		nsec
Serial Clock Speed	SPI 3 wire bus	F_{CLK}		10	50	MHz
Serial Setup Time	From rising edge of Vmode to rising edge of CLK for D5	A	20			ns
Clock width	Clock high pulse width	B	10			ns
LE setup time	From rising edge of CLK pulse for D0 to LE rising edge	C	10			ns
LE pulse	LE minimum pulse width	D	30			ns

SPECIFICATION NOTES:

- 1 – Items in min/max columns in **bold italics** are Guaranteed by Test
- 2 – All other Items in min/max columns are Guaranteed by Design Characterization

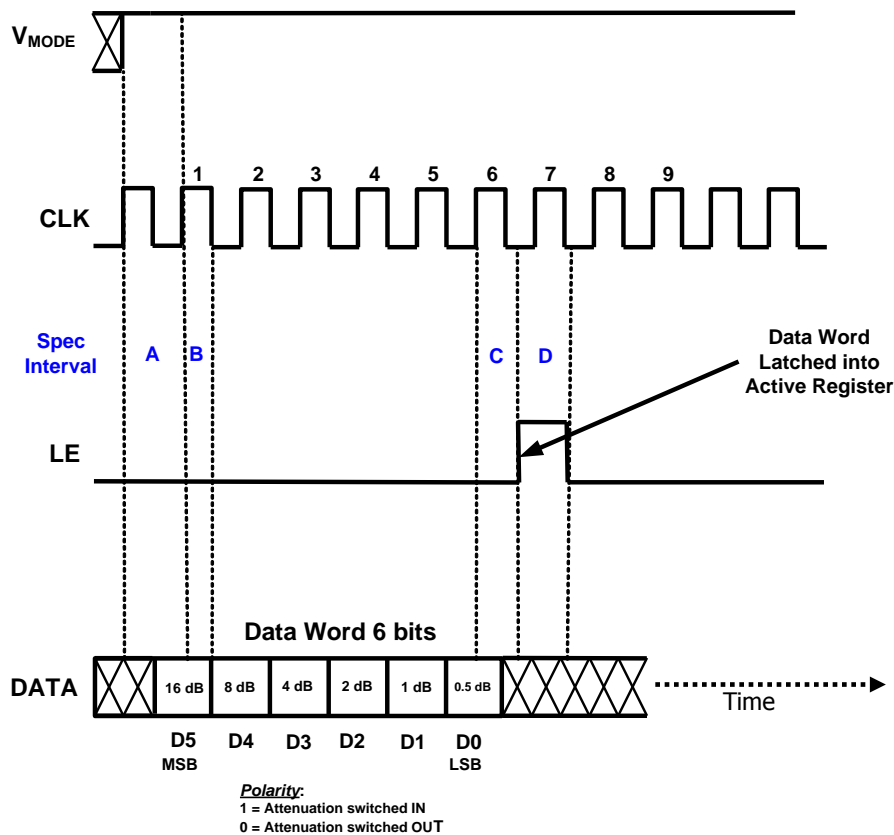
SERIAL CONTROL

Serial mode is selected when V_{MODE} is pulled high ($> V_{IH}$). In serial mode the F1953 attenuation setting is programmed via the 3 wire bus (LE, CLK, DATA). In serial mode data is clocked in MSB first. Note the timing diagram below.

Note – The IDTF1953 includes a CLK inhibit feature designed to minimize sensitivity to CLK bus noise when the device is not being programmed. When Latch enable is high ($> V_{IH}$), the CLK input is disabled and DATA will not be clocked into the shift register. It is recommended that LE be pulled high ($> V_{IH}$) when the device is not being programmed.

SERIAL REGISTER DEFAULT CONDITION

If the device is powered up in Serial Mode, the device will default to whatever attenuation state is defined by the six parallel data input pins D5,D4,D3,D2,D1,D0 thus allowing *any attenuation setting* to be specified as the power up state.

SERIAL REGISTER TIMING DIAGRAM: (Note the Timing Spec Intervals in **Blue**)

SERIAL REGISTER TIMING TABLE

Interval Symbol	Description	Min Spec	Max Spec	Units
A	From rising edge of Vmode to rising edge of CLK for D5	20		nsec
B	Clock high pulse width	10		nsec
C	From rising edge of CLK pulse for D0 to LE rising edge	10		nsec
D	LE minimum pulse width	30		nsec

PARALLEL CONTROL MODE

The user has the option of running in one of two parallel modes: *Direct Parallel Mode* or *Latched Parallel Mode*.

DIRECT-PARALLEL MODE:

Direct-parallel mode is selected when V_{MODE} (pin 13) is $< V_{IL}$ and LE (pin 5) is $> V_{IH}$. In this mode the device will immediately react to any voltage changes to the parallel control pins [pins 1, 15, 16, 17, 19, 20]. Use direct-parallel mode for the fastest settling time.

LATCHED-PARALLEL MODE:

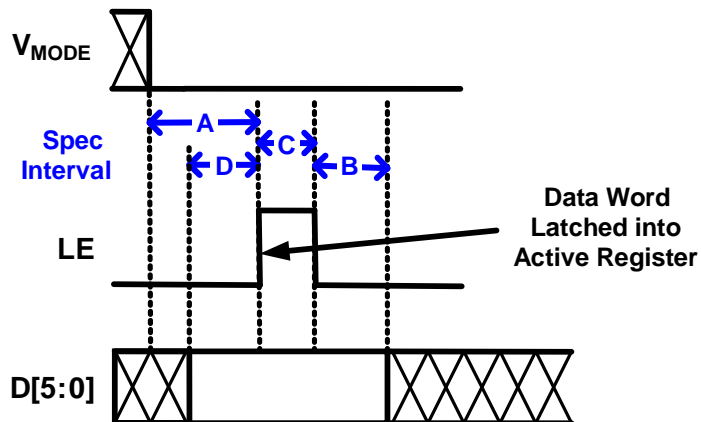
Latched-parallel mode is selected when V_{MODE} (pin 13) is $< V_{IL}$ and LE (pin 5) is toggled from $< V_{IL}$ to $> V_{IH}$

To utilize latched-parallel mode:

- Set $LE < V_{IL}$
- Adjust pins [1, 15, 16, 17, 19, 20] to the desired attenuation setting. (Note the device will not react to these pins while $LE < V_{IL}$.)
- Pull $LE > V_{IH}$. The device will then transition to the attenuation settings reflected by these pins.

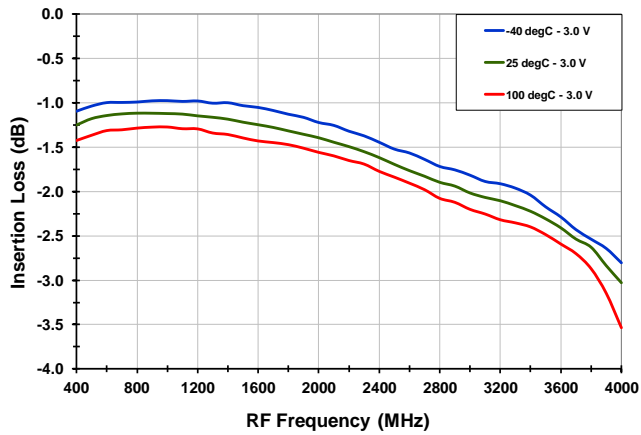
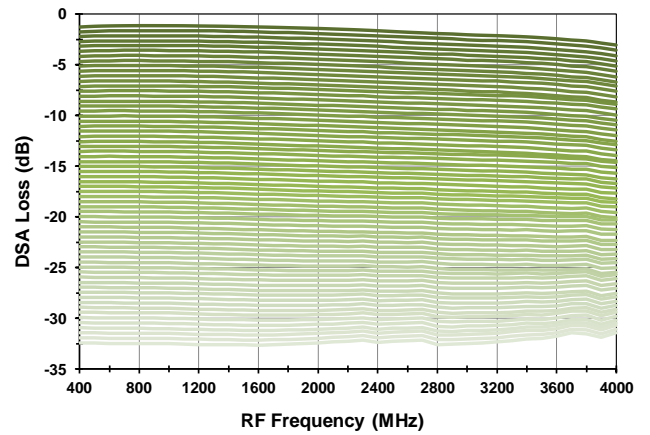
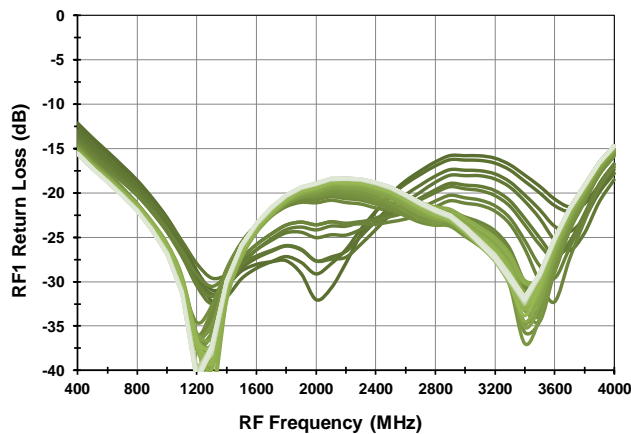
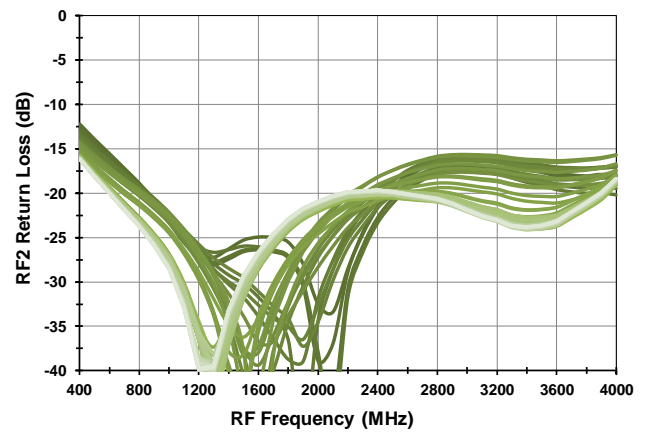
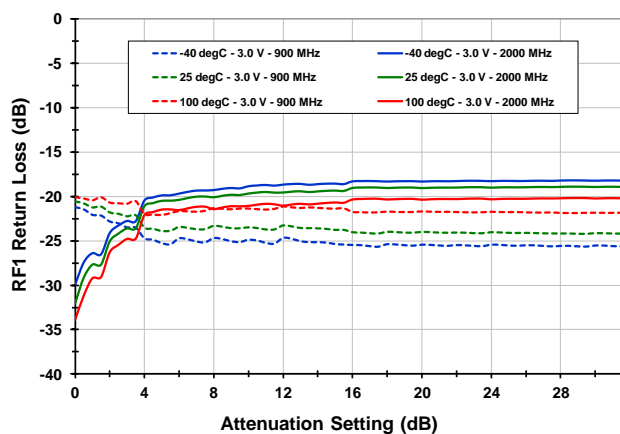
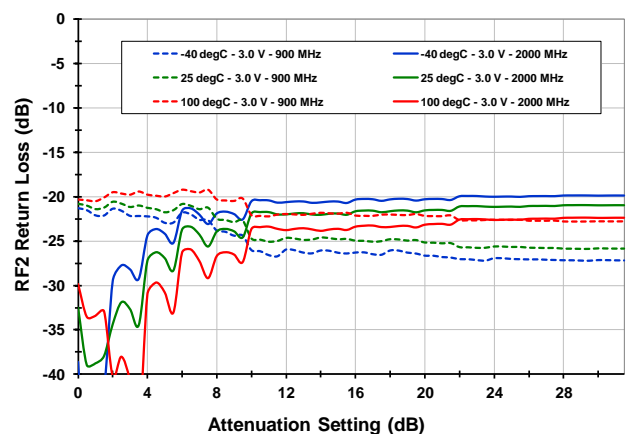
When the device is powered up In Latched Parallel Mode [$V_{MODE} < V_{IL}$ and $LE < V_{IL}$], the attenuation setting defaults to the state defined by the six parallel data pins [pins 1, 15, 16, 17, 19, 20]

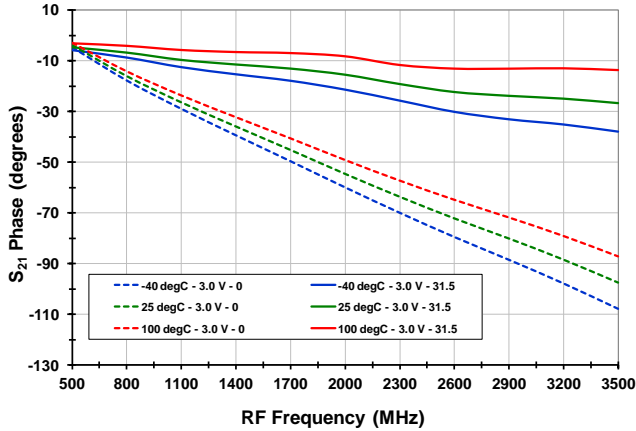
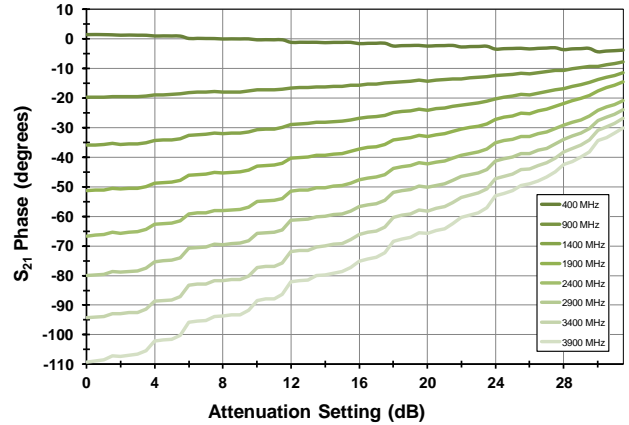
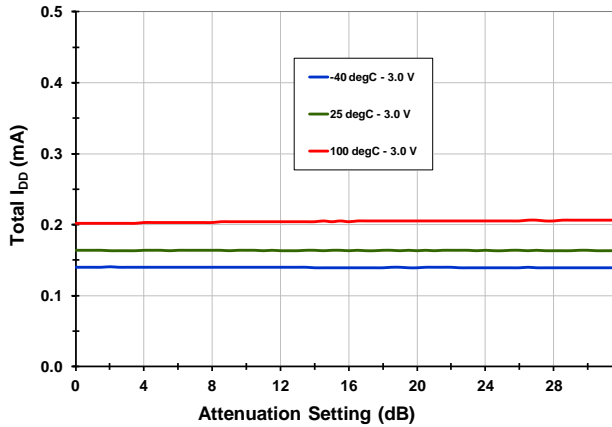
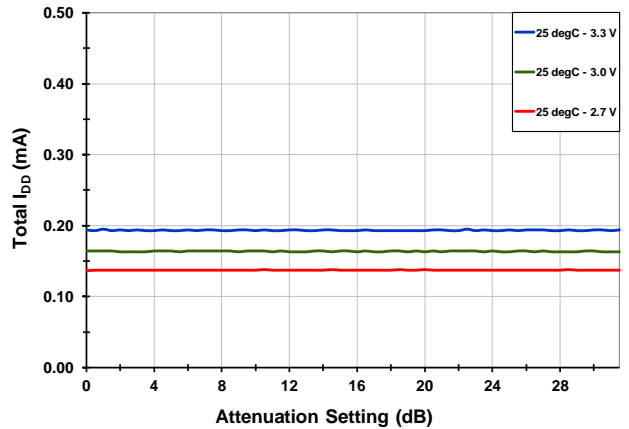
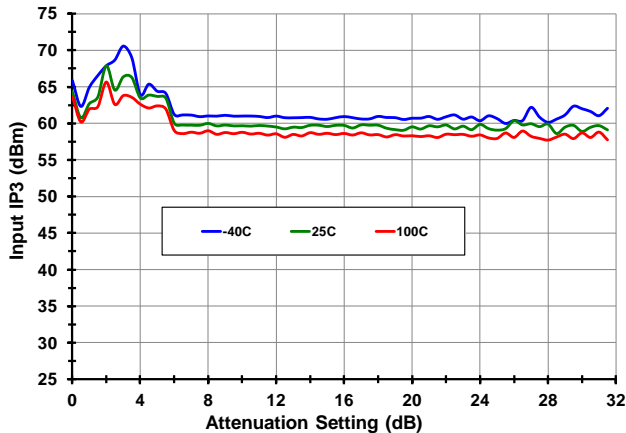
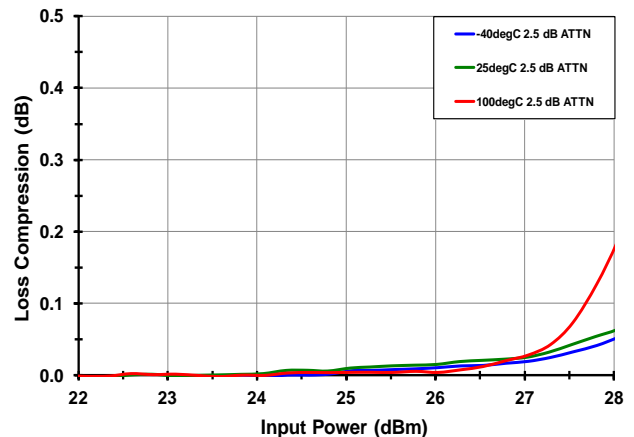
LATCHED PARALLEL MODE TIMING DIAGRAM: (Note the Timing Spec Intervals in Blue)

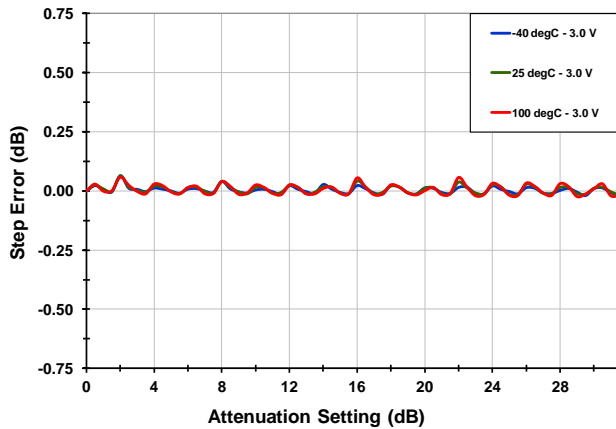
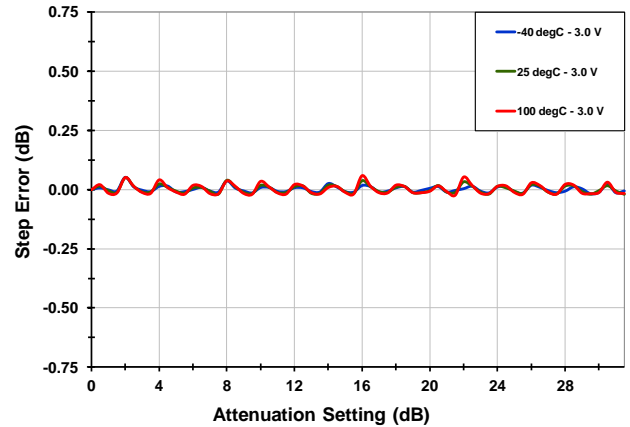
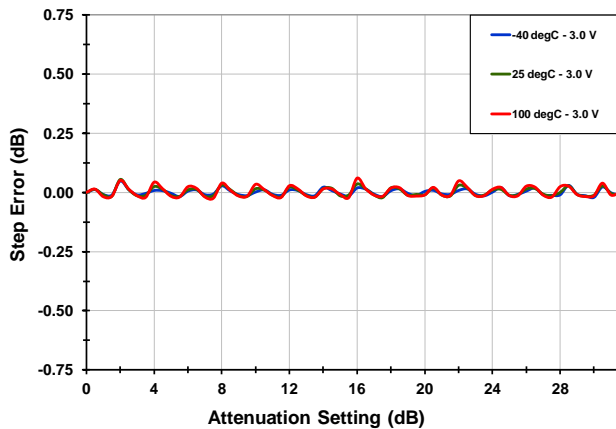
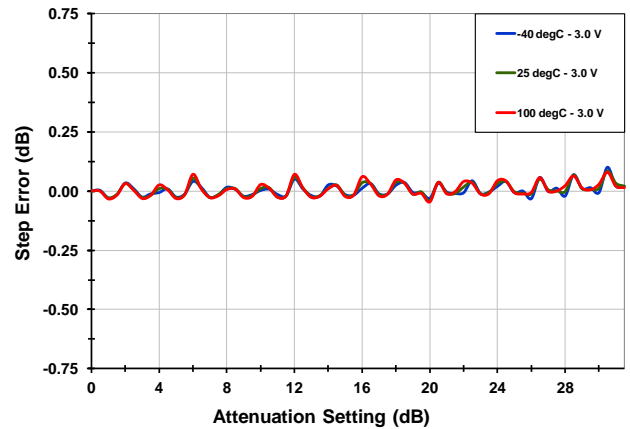
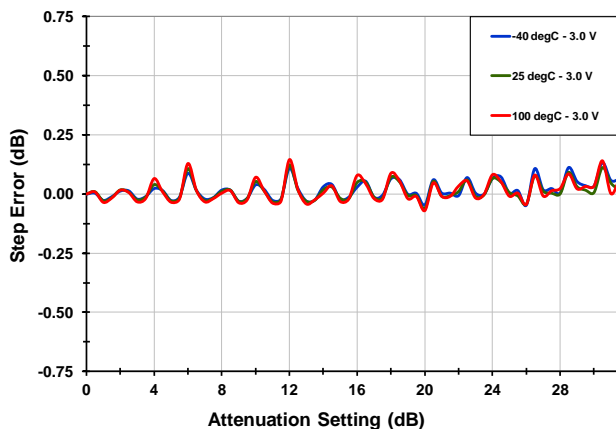
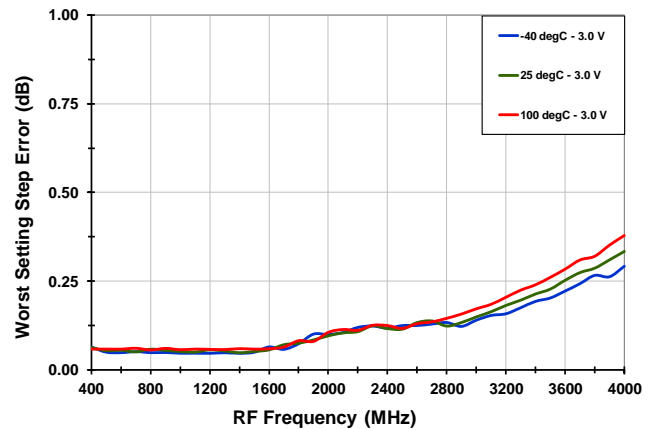


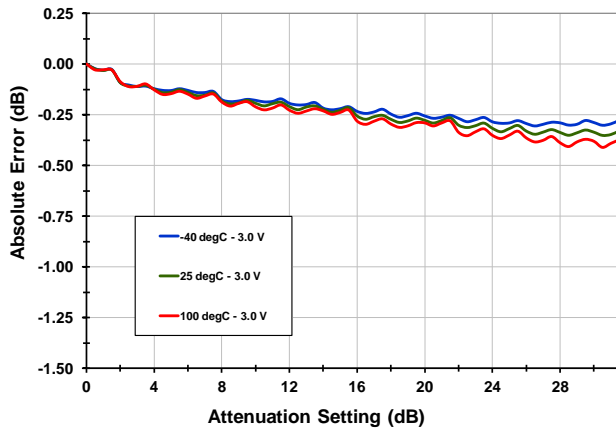
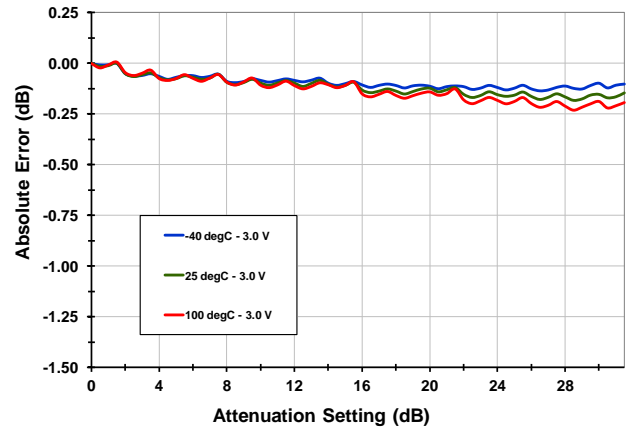
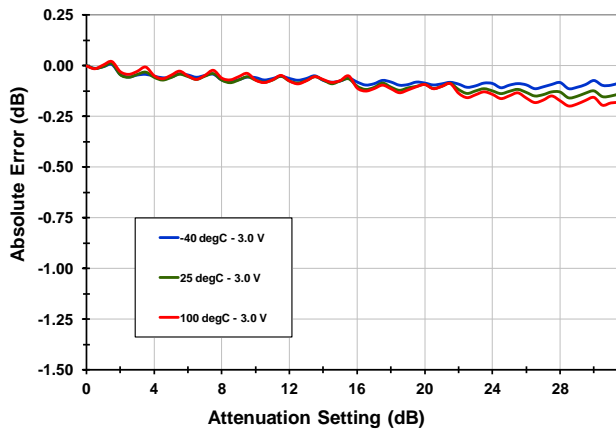
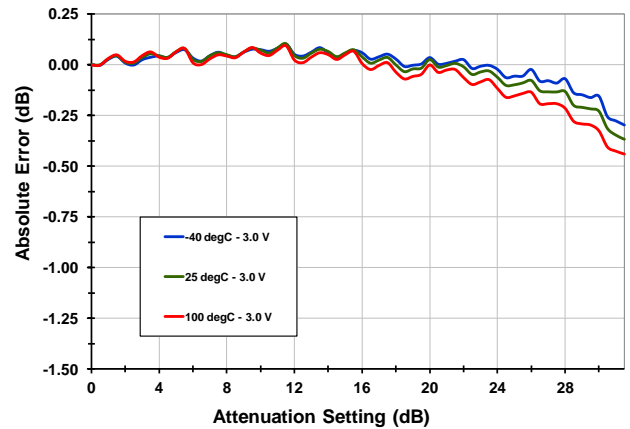
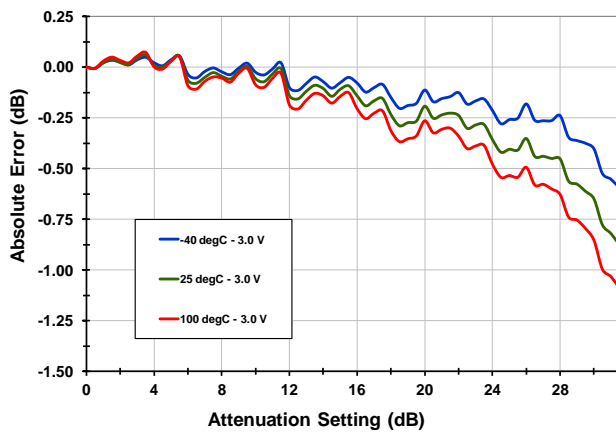
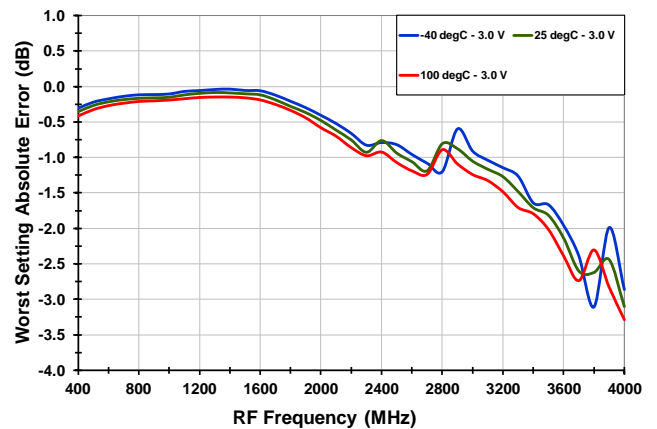
LATCHED PARALLEL MODE TIMING TABLE:

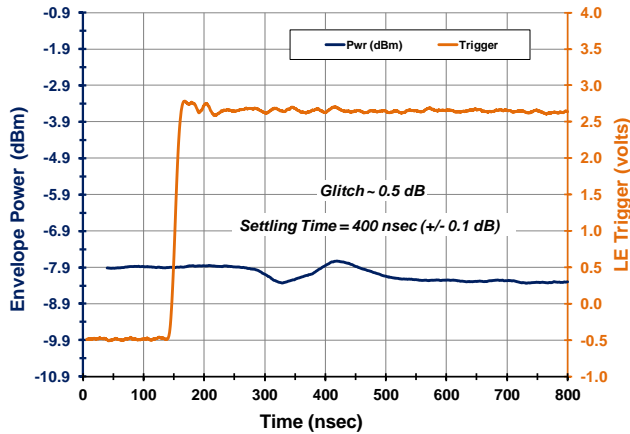
Interval Symbol	Description	Min Spec	Max Spec	Units
A	Serial to Parallel Mode Setup Time	100		nsec
B	Parallel Data Hold Time	10		nsec
C	LE minimum pulse width	10		nsec
D	Parallel Data Setup Time	10		nsec

TYPICAL OPERATING PARAMETRIC CURVES (EVKit loss de-embedded, 3.0V unless otherwise noted)
Insertion Loss vs. Frequency [A_{MIN}]

Attenuation vs. Freq [$T_{CASE} = +25C, 0.5$ dB steps]

 S_{11} vs. Frequency [$T_{CASE} = +25C, 0.5$ dB steps]

 S_{22} vs. Frequency [$T_{CASE} = +25C, 0.5$ dB steps]

 S_{11} vs. Attenuation State

 S_{22} vs. Attenuation State


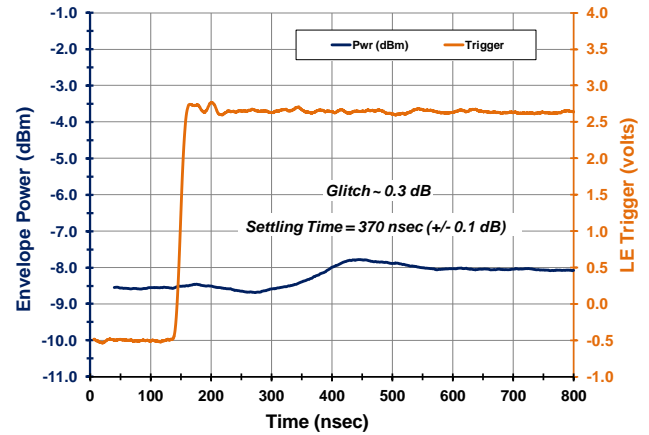
TOCS CONTINUED (-2-)
Phase vs. Frequency

Phase vs. Attenuation Setting

Supply Current I_{DD} [vs. Temp]

Supply Current I_{DD} [vs. V_{DD}]

Input IP₃ [f_{RF} = 1900 MHz, V_{DD} = 3.0 V]

Compression [f_{RF} = 2000 MHz, ATTN = 2.5 dB]


TOCS CONTINUED (-3-)
DNL [400 MHz]

DNL [700 MHz]

DNL [900 MHz]

DNL [1900 MHz]

DNL [2800 MHz]

Worst Setting DNL


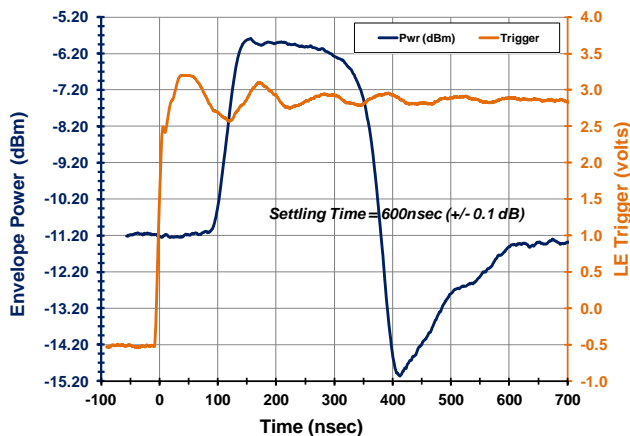
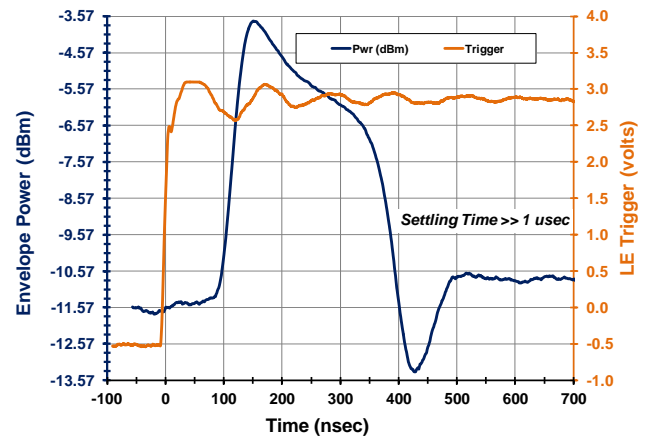
TOCS CONTINUED (-4-)
INL [400 MHz]

INL [700 MHz]

INL [900 MHz]

INL [1900 MHz]

INL [2900 MHz]

Worst Setting INL


TOCS CONTINUED (-5-) [$f_{RF} = 900 \text{ MHz}$]
Transient [15.5 to 16.0 (MSB+) 3.3V F1953]


The graphs **ABOVE** show the transient overshoot and settling time performance for both the MSB+ and MSB- cases for the F1953. The device settles very quickly (~ 400) nsec with benign (~ 0.5) dB overshoot.

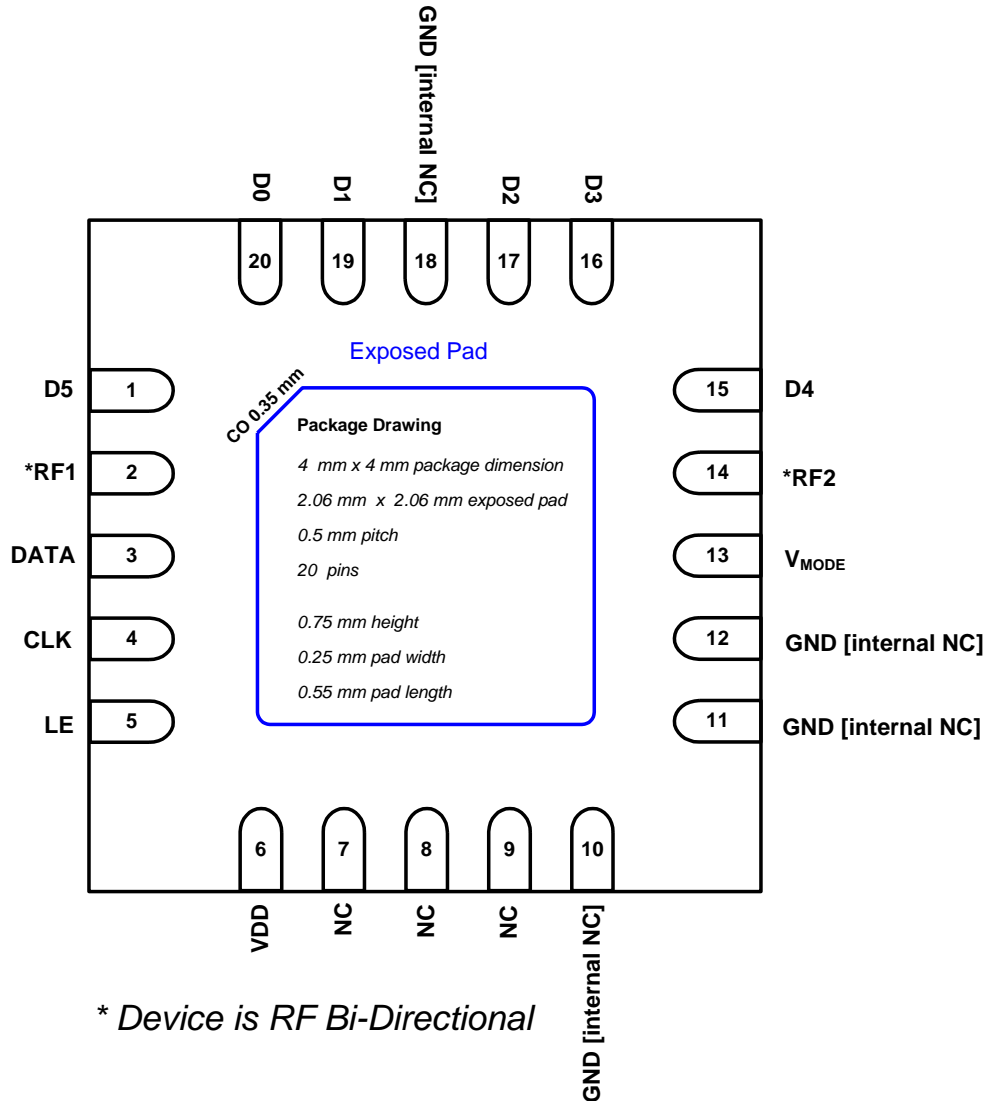
Transient [16.0 to 15.5 (MSB-) 5.0V F1953]


The graphs **BELOW** show the transient overshoot and settling time performance for a popular competing DSA. *Note the overshoot/undershoot excursion of almost 10 dB and the very long settling time.* For the MSB- case, the settling time is off the scale, ~ 3 usec.

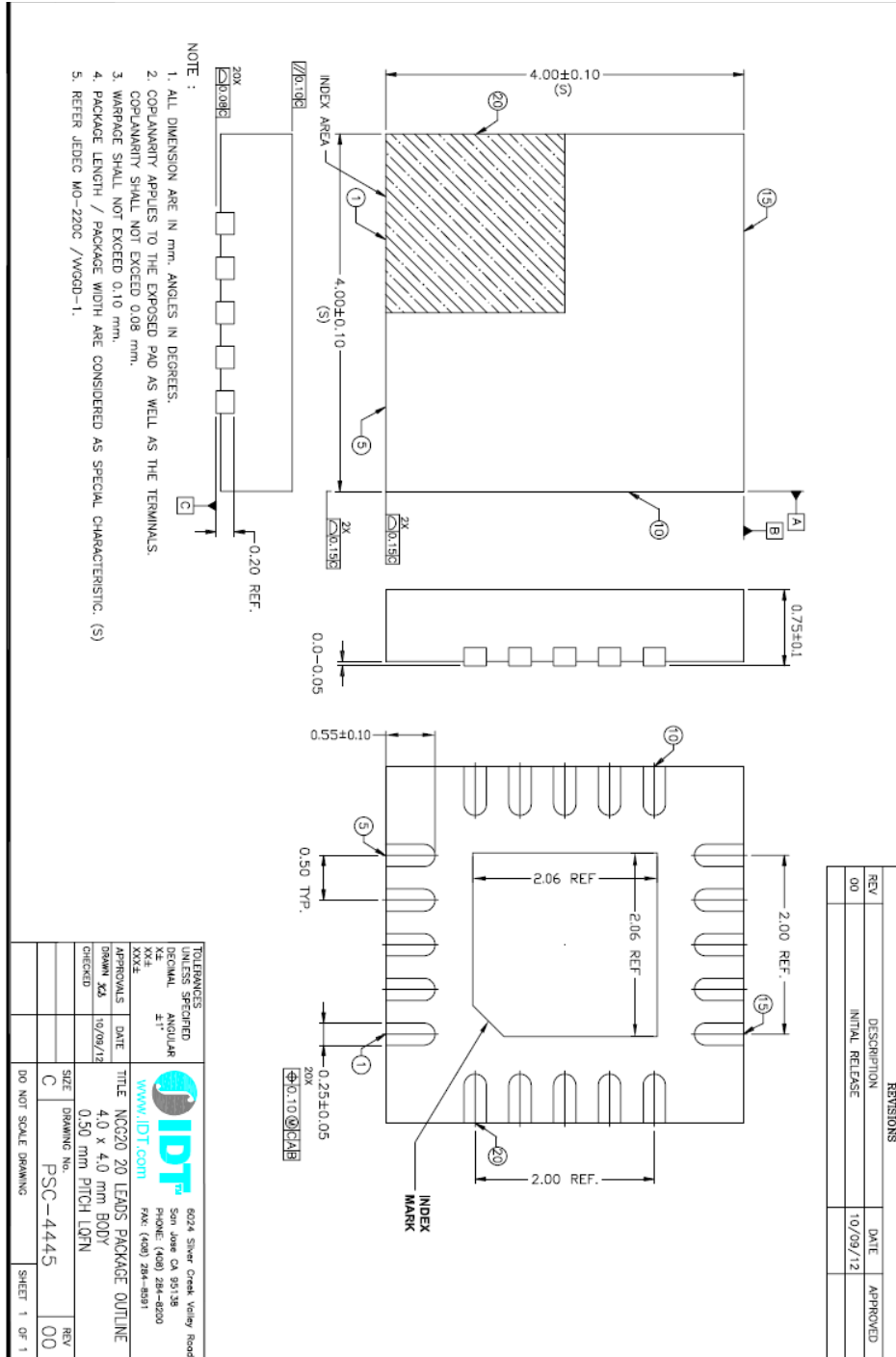
Transient [15.75 to 16.00 (MSB+) Standard DSA]

Transient [16.00 to 15.75 (MSB-) Standard DSA]


PIN DIAGRAM

TOP View
(looking through the top of the package)



PACKAGE DRAWING (NCG20 4x4 20 PIN)



PIN DESCRIPTIONS

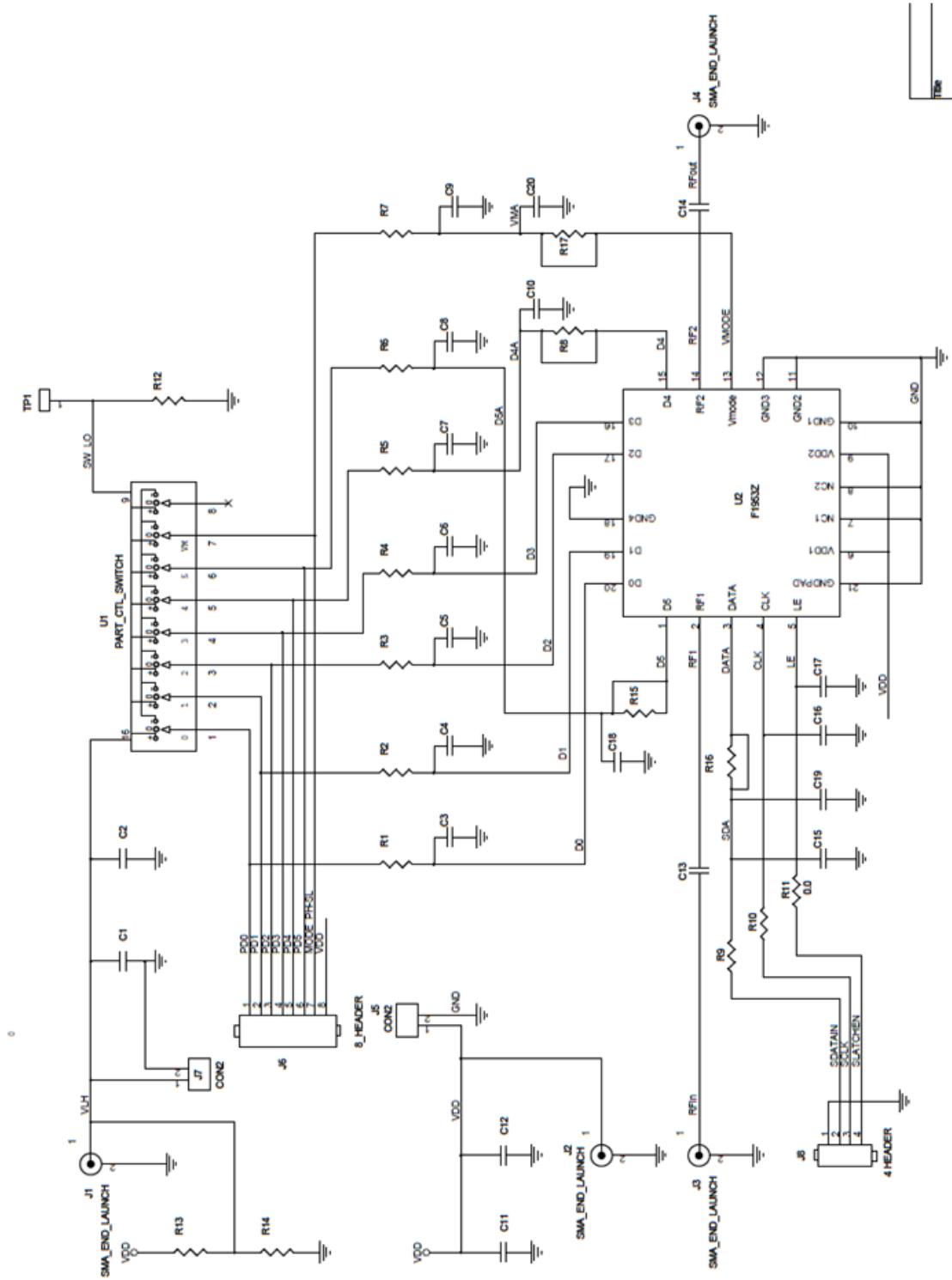
Pin #	Pin Name	Pin Function
1	D5	16 dB Attenuation Control Bit. Pull high for 16 dB ATTN.
2	RF1	Device RF input or output (bi-directional). Internally DC blocked.
3	DATA	Serial interface Data Input.
4	CLK	Serial interface Clock Input.
5	LE	Serial interface Latch Enable Input. Internal pullup (100K ohm).
6	VDD	Power supply pin.
7	NC	Internally unconnected.
8	NC	Internally unconnected.
9	NC	Internally unconnected.
10	GND	Connect to Ground. (this pin is internally unconnected)
11	GND	Connect to Ground. (this pin is internally unconnected)
12	GND	Connect to Ground. (this pin is internally unconnected)
13	VMODE	Pull high for serial mode. Ground for Parallel control mode.
14	RF2	Device RF input or output (bi-directional). Internally DC blocked.
15	D4	8 dB Attenuation Control Bit. Pull high for 8 dB ATTN.
16	D3	4 dB Attenuation Control Bit. Pull high for 4 dB ATTN.
17	D2	2 dB Attenuation Control Bit. Pull high for 2 dB ATTN.
18	GND	Connect to Ground. (this pin is internally unconnected)
19	D1	1 dB Attenuation Control Bit. Pull high for 1 dB ATTN.
20	D0	0.5 dB Attenuation Control Bit. Pull high for 0.5 dB ATTN.
EP	Exposed Paddle	Connect to Ground with multiple vias for good thermal relief.

6-bit Digital Step Attenuator

400 to 4000 MHz IDTF1953

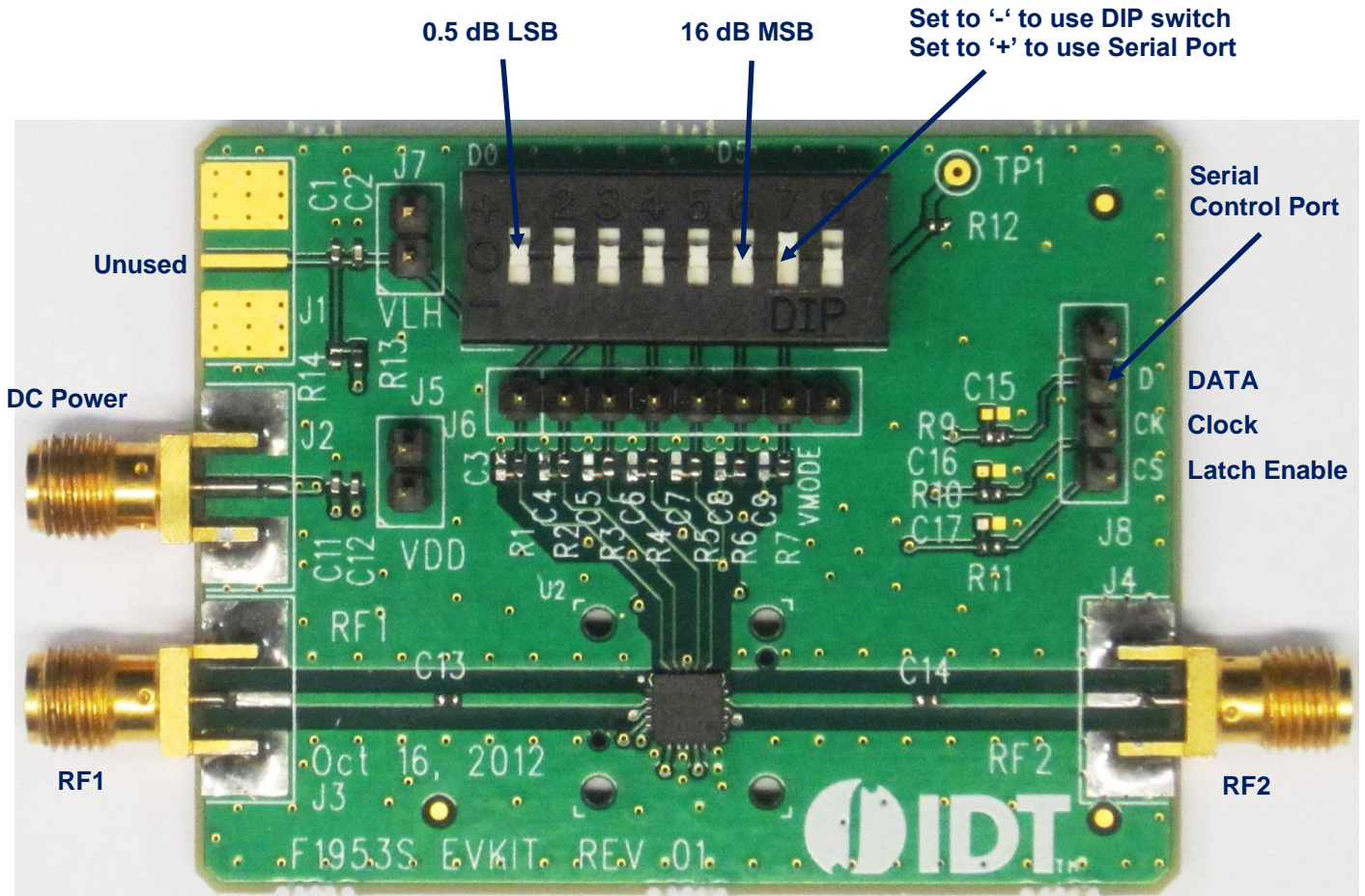
EVKIT SCHEMATIC

The diagram below describes the recommended applications / EVkit circuit:



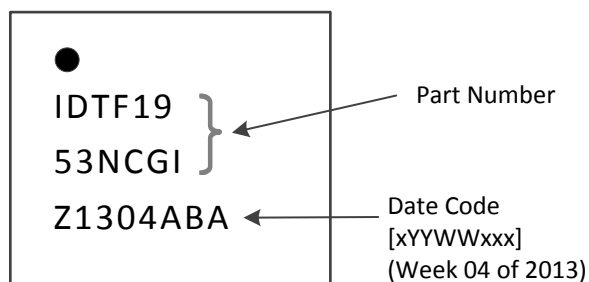
EVKIT OPERATION (Email: RFsupport@IDT.com to request an EVkit and Controller)

The picture and graphic below describe how to operate the EVkit



6-bit Digital Step Attenuator
400 to 4000 MHz IDTF1953
EVKIT BOM (F1953)
F1953 BOM Rev 01 PCB Rev 01 11/15/2012

Item #	Value	Size	Desc	Mfr. Part #	Mfr.	Part Reference	Qty
1	10nF	0402	CAP CER 10000PF 16V 10% X7R 0402	GRM155R71C103KA01D	MURATA	C2,12	2
2	0.1uF	0402	CAP CER 0.1UF 16V 10% X7R 0402	GRM155R71C104KA88D	MURATA	C1,11	2
3	Header 2 Pin	TH 2	CONN HEADER VERT SGL 2POS GOLD	961102-6404-AR	3M	J5,7	2
4	Header 4 Pin	TH 4	CONN HEADER VERT SGL 4POS GOLD	961104-6404-AR	3M	J8	1
5	Header 8 Pin	TH 8	CONN HEADER VERT SGL 8POS GOLD	961108-6404-AR	3M	J6	1
6	SMA_END_LAUNCH	.062	SMA_END_LAUNCH (Small)	142-0711-821	Emerson Johnson	J2,3,4	3
7	0	0402	RES 0.0 OHM 1/10W 0402 SMD	ERJ-2GE0R00X	Panasonic	R1-7,12,C13,C14	10
8	3K	0402	RES 3.00K OHM 1/10W 1% 0402 SMD	ERJ-2RKF3001X	Panasonic	R9-11	3
9	10K	0402	RES 10K OHM 1/10W 1% 0402 SMD	ERJ-2RKF1002X	Panasonic	R8,15-17	4
10	100K	0402	RES 100KOHM 1/10W 1% 0402 SMD	ERJ-2RKF104X	Panasonic	R13	1
11	267K	0402	RES 267K OHM 1/10W 1% 0402 SMD	ERJ-2RKF2673X	Panasonic	R14	1
12	DIPSwitch	TH 10	8 POSITION DIP SWITCH	KAT1108E	E-Switch	U1	1
13	Digital Step Attenuator		F1953Z	F1953Z	IDT	U2	1
14	PCB		PCB Rev 01	F1953S Evkit Rev 01	SBC		1
15	100pF	0402	CAP CER 100PF 16V 10% X7R 0402	GRM155R71C103KA01D	MURATA	C3-10,15-20	DNP
16	SMA_END_LAUNCH	.062	SMA_END_LAUNCH (Small)	142-0711-821	Emerson Johnson	J1	DNP

Total 33
TOPMARKINGS


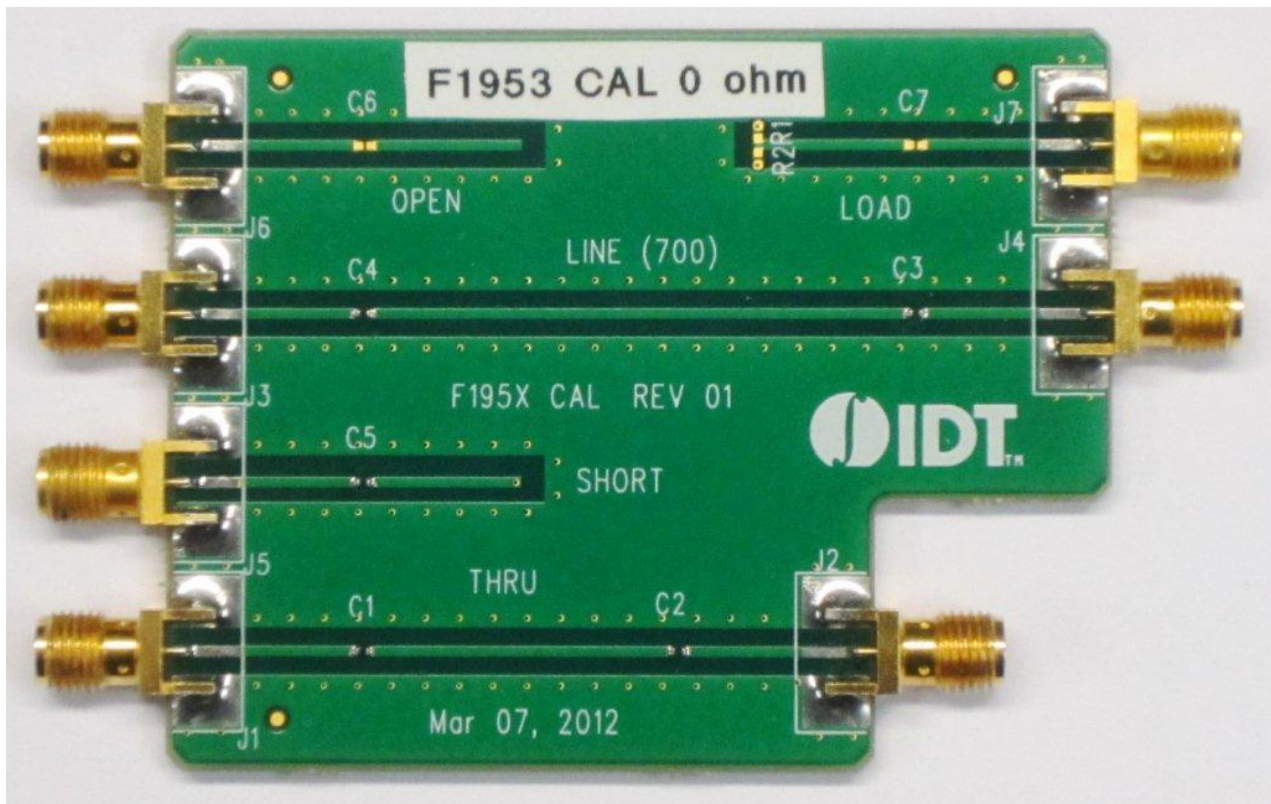
EVKIT THROUGH-REFLECT-LINE (TRL) CALIBRATION

The “Through-Reflect-Line” (TRL) method [1] is used to de-embed the evaluation board losses from the S-parameter measurements of the F1953. This method requires the use of three standards: a through, a reflection, and a line. The TRL method has the advantage over other calibration methods in that it requires only one of these three standards to be well defined.

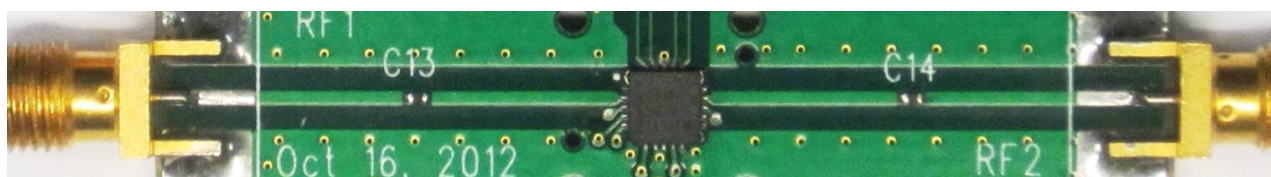
The TRL through which is used for the F1953 TRL calibration was constructed identically to the evaluation board, minus the DUT and its corresponding length. Therefore, the through corresponds to a precise zero length connection between the input and output reference planes of the DUT. This through satisfies the requirement of the TRL method that one of the three standards be precisely specified.

The TRL reflection standard used is constructed identically to the input and output lines of the evaluation board, with a short placed at the reference plane of the DUT. In accordance with the TRL method’s requirements, the actual magnitude and phase were not accurately specified, but the phase was known to within 90 degrees and the TRL reflection standard has a magnitude close to one.

The TRL line standard is identical to the TRL through, but with an additional length of 0.8 inches (2 cm). This satisfies the TRL method’s requirement that the TRL be a different length than the TRL through, that it have the same impedance and propagation constant as the through, and that the phase difference between the through and the line be between 20 degrees and 160 degrees. The difference in length yields a phase difference of approximately 20 degrees at 500 MHz, and a phase difference of 160 degrees at 4 GHz.



Standards used for F195x TRL calibration



F1953 evaluation circuit

Engen, G.F.; Hoer, C.A.; “Thru-Reflect-Line: An Improved Technique for Calibrating the Dual Six-Port Automatic Network Analyzer,” **IEEE Transactions on Microwave Theory and Techniques**, Volume: 27 Issue:12, pp. 987 – 993, Dec 1979.



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